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THE UNIVERSITY OF ALBERTA

THE ENHANCEMENT OF SIGNAL TO NOISE  
IN MAGNETOTELLURIC DATA

by

(C)

DOMINIQUE WEN KAO

A THESIS

SUBMITTED TO THE FACULTY OF GRADUATE STUDIES AND RESEARCH  
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## ABSTRACT

The magnetotelluric (MT) method of sounding is reviewed and the interpretative techniques including both numerical modelling and tensor impedance concepts are presented in a formulation suitable for the study of the problem of noise in magnetotelluric signals.

A statistical treatment of the noise problem has resulted in the development of a cyclical process which produces a pronounced enhancement of the signal to noise ratio over the complete spectral range in the majority of processed data samples. In those cases where the data was not so improved a definite confidence limit could be placed on that portion of the spectral range which was acceptable.

This technique was applied to magnetotelluric soundings made at 14 sites around the Black Hills of South Dakota in the frequency band  $10^{-3}$  to 10 Hz. The soundings cover an area of approximately 30,000 square Km and correspond to a depth range of 0.1 to 50 Km.

A three dimensional earth model of the estimated resistivity distribution is derived from the sounding results, using the generalized impedance concepts. An anomaly of high conductivity at depths of 3 to 10 Km is found in the central region of the Hills, under the cover of highly resistive granitic material.

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## TABLE OF CONTENTS

	<u>Page</u>
<b>CHAPTER I      INTRODUCTION</b>	1
A. Historical Review	1
B. Outline of the Thesis	3
<b>CHAPTER II    THEORETICAL DISCUSSION</b>	5
A. Maxwell's Equations for Plane Waves in a Uniform Medium	5
B. Consideration of Skin Depth Effect	7
C. Plane Wave Incident on an Uniform Earth	8
D. Apparent Resistivity of Layered Earth	9
1) Single-layer earth	9
2) n-layer earth	11
E. Tensor Impedance Analysis	14
F. Examples of Impedance Functions	18
1) Uniform earth	18
2) Layered earth	19
3) Vertical fault - two-dimensional model	21
G. Reduction of a Three-Dimensional Problem to a Two-Dimensional Formulation	23
H. Orientational Behavior	28

	<u>Page</u>
CHAPTER II (cont'd)	
I. Special Cases and the Indication of Dimensionality	33
1) One-dimensional case	33
2) Two-dimensional case	34
3) Three-dimensional case	37
CHAPTER III INSTRUMENTATION AND DATA ANALYSIS	42
A. Instrumentation	42
1) Magnetic system	42
2) Electric system	45
3) Digital recording	45
B. Data Analysis	45
1) Power spectrum computation	45
a) Truncation	46
b) Removing mean and trend	46
c) Tapering	46
d) Fourier transform	47
e) Hamming window	48
f) System response correction	48
g) Power spectrum computation	48
2) Polarization Discussion	50
a) Degree of Polarization	50
b) Polarization Angle	51
c) Ellipticity	51

	<u>Page</u>
CHAPTER III (cont'd)	
B. 3) Impedance computation	52
4) Criteria of data reliability	55
a) Phasor Coherency (CP)	55
b) Predicted Coherency (PCH)	56
CHAPTER IV NOISE DISCUSSION	58
A. General Discussion	59
1) Covariance and coherence	59
2) Power spectrum of signal and noise	62
a) Power spectrum of random noise	62
b) Power spectrum of signal	64
c) Spectral densities with mixed signal and noise	65
3) Power estimates and averaging effect	68
4) Experiment	71
B. Investigation of Noise Effect in Impedance Elements	72
1) General solution	72
2) Impedance solutions for noise-free cross-power estimates	78
3) Impedance solutions as noise in both auto- and cross-power estimates	83

	<u>Page</u>
CHAPTER V      MT SOUNDING	90
A.    Geological Background	90
B.    Data Analysis and Result	98
C.    Interpretation and Discussion	112
1) General discussion	112
2) Earth model	113
3) Summary and comment	115
APPENDIX A    MODELLING CONSIDERATION	130
I.    One-Dimensional Model	130
II.   Two-Dimensional Model	131
APPENDIX B   TABULATED ANALYSIS RESULTS	141
BIBLIOGRAPHY	125

## CHAPTER I

### INTRODUCTION

#### 1-A. Historical Review

The process of determining the subsurface resistivity structure of the earth, by measurements made at the surface, using the naturally occurring electromagnetic fields as a source, was originally proposed by Tikhonov (1950). Measurements were carried out both by Tikhonov (1950), Kato and Kikuchi (1950), and Rikitake (1950, 1951). In a definitive work Cagniard (1953) presented a graphical method for the interpretation of results in the case of a horizontally layered earth and introduced the expression "Magnetotelluric Method" to designate the technique.

As originally envisaged by Cagniard, the source was infinitely extended and thus a plane wave impinged on the horizontally stratified earth. Wait (1954), and later Price (1964), investigated the effect of finite source dimension on the measured impedances, and argued that if the source is finite in extent or possesses horizontal gradients, complex angles of incidence must be considered. However, Madden and Nelson (1964) showed that for reasonable earth conductivity models, the plane wave assumption appears to

be valid in the frequency range of interest for the MT method.

A more important consideration than that of source dimensions, may be that a one-dimensional model is inadequate for a real earth where lateral inhomogeneity and anisotropy in the conductivity are present, and the electric and magnetic fields may, in general, no longer be orthogonal. d'Erceville and Kunetz (1962) and Rankin (1962) gave analytic solutions for a vertical fault and vertical dyke respectively. Chetaev (1960), Cantwell (1960), Bostick and Smith (1962), Mann (1965), O'Brien and Morrison (1967), Praus and Petr (1969), Rankin and Reddy (1969, 1970), and others have studied various types of anisotropy in conductivity by means of impedance tensor technique, and have proposed methods to find the principal directions of conductivity anisotropy. Madden and Nelson (1964), Swift (1967), Morrison et al. (1968), Sims and Bostick (1969), Word et al. (1970), Vozoff (1972), and others have presented technique to compute tensor impedance elements; this tensor matrix relates the electro-magnetic fields for two and three dimensional structures. Analog model studies have been carried out by Rankin et al. (1965), Dosso (1966), Takaes (1969), and others.

A considerable number of experimental results have been reported from various locations in North America, Europe, and Russia (Srivastava et al. 1963, Whitham and Anderson 1966, Berdichevsky 1966, Vozoff 1969, 1972, Peeples and Rankin 1973, and others). While the results have been significant, many of the questions and problems involved, particularly with respect to interpretive techniques, still remain.

#### I-B. Outline of the Thesis

Chapter two will present the theory of the MT methods.

Chapter three will describe the recording system and the techniques of data analysis.

Chapter four will discuss the noise effect, and present a technique for improving the signal to noise ratio.

Chapter five will present the sounding results, geological background, and model interpretation for the Black Hills in South Dakota.

## CHAPTER II

### THEORETICAL DISCUSSION

#### III-A. Maxwell's Equation for Plane Waves in a Uniform (homogeneous and isotropic) Medium

Consider the model in the cartesian coordinate system represented by xyz-axes as shown in Fig. (2-1).

The z-axis is vertically downward and the earth of conductivity  $\sigma(x, y, z)$  occupies the half space  $z \geq 0$  with the surface at  $z = 0$ . The source is in free space  $z < 0$ . Maxwell's Equations in the MKS system for  $e^{j\omega t}$  time dependence are

$$\nabla \times \vec{E} = - \frac{\partial \vec{B}}{\partial t} = -j\omega\mu\vec{H} \quad (2-1a)$$

$$\nabla \times \vec{H} = \vec{J} = (\sigma + j\omega\epsilon)\vec{E} \quad (2-1b)$$

$$\nabla \cdot \vec{E} = q/\epsilon \quad (2-1c)$$

$$\nabla \cdot \vec{H} = 0$$

where  $\omega = 2\pi f$  is the angular frequency;  $q$  is free charge density; and  $\sigma, \mu, \epsilon$  are the conductivity, permeability, and dielectric constant respectively, of the media.

Except for anomalies which will not be discussed in this thesis,  $\mu = \mu_0$ , and  $\epsilon = \epsilon_0$ . For a uniform half space, the wave equation can be obtained from (2-1a,b) by taking the Curl.

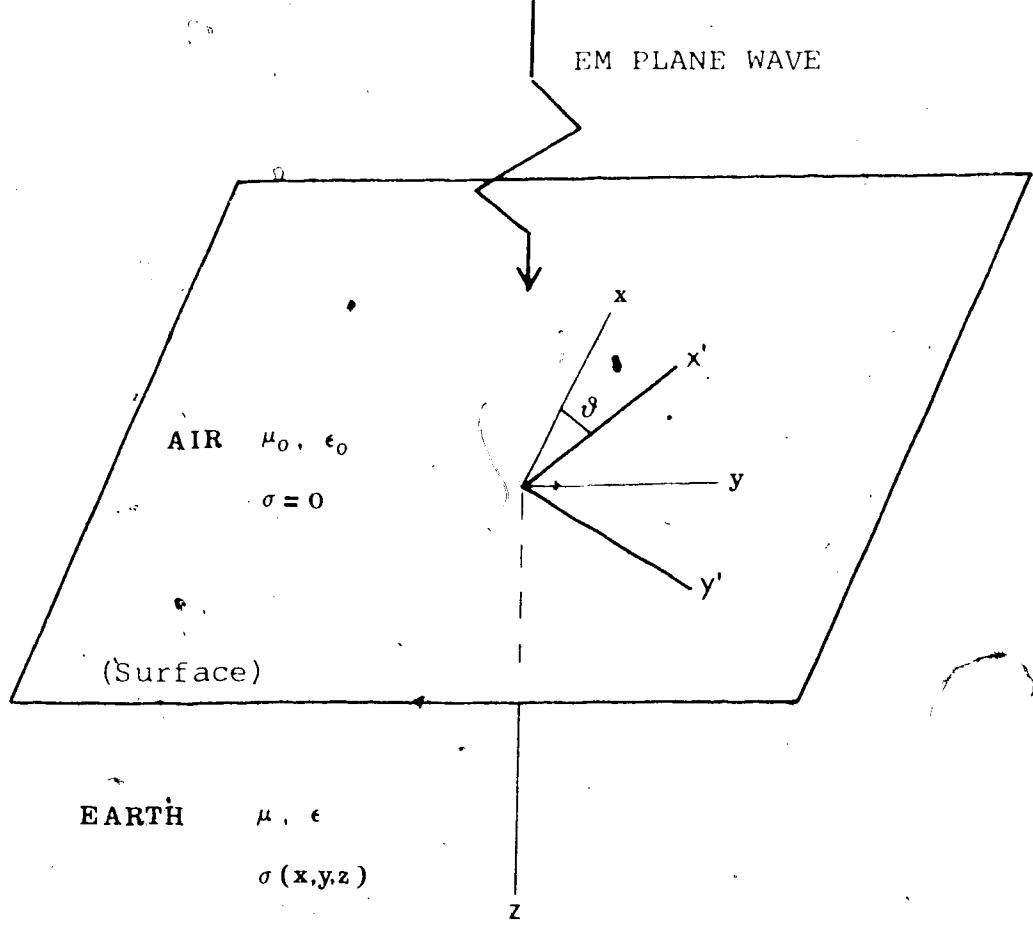


Fig. (2-1). Coordinate system and earth model.

$$\nabla \times \nabla \times \vec{A} = -j\omega\mu(\sigma + j\omega\epsilon)\vec{A} \quad (2-2)$$

where  $\vec{A}$  represents a field vector  $E$  or  $H$ . Eqn. (2-2) describes the field for all regions.

Except at discontinuities or in the source region, the free charge density is zero and hence within the medium  $\nabla \cdot E = 0$  and Eq. (2-2) becomes the Helmholtz Equation

$$\nabla^2 \vec{A} = j\omega\mu(\sigma + j\omega\epsilon)\vec{A} \quad (2-3)$$

In a homogeneous region where  $\sigma$  is constant, Eqn. (2-3) is satisfied by a plane wave solution of the form

$$\vec{A} = A_0 e^{j\vec{k} \cdot \vec{r}}$$

where

$\vec{r} = \hat{i}x + \hat{j}y + \hat{k}z$  is the position vector, and

$\vec{k} = \hat{i}k_x + \hat{j}k_y + \hat{k}k_z$  is the propagation vector, and

$$K = |\vec{k}| = \sqrt{j\omega\mu(\sigma + j\omega\epsilon)} \quad (2-4)$$

The frequency of interest in this work will range from  $10^{-3}$  — 10 Hz, and the conductivity of the earth from  $10^{-4}$  — 1 mhos. Thus  $\sigma \gg \omega\epsilon$  within the earth, and henceforth the propagation constant in the earth becomes approximately

$$K_1 = \sqrt{j\omega\mu\sigma} = (1+j)\sqrt{\omega\mu\sigma/2} \quad (2-5a)$$

In the air where  $\sigma = 0$ , the propagation constant is

$$K_0 = j\omega\sqrt{\mu\epsilon} \quad (2-5b)$$

It is obvious that  $K_0 \ll K_1$ .

### II-B. Consideration of Skin Depth Effect

When the field propagates in the earth, the real component of  $K_1$  produces an attenuation effect on the field. The attenuation factor for a given wave in a medium with conductivity  $\sigma$  is

$$\text{Re}[K_1] = \sqrt{\omega\mu\sigma/2} \quad \text{or} \quad \sqrt{\pi\mu_0/\rho T}$$

The propagation distance or penetration depth  $\delta$ , within which the strength of the field is attenuated by a factor of  $e^{-1}$ , is defined as the skin depth, where

$$\delta = 1/\text{Re}[K_1] = \sqrt{2/\omega\mu\sigma} \quad (2-6)$$

The wave length of the field in this conductive medium is related to the skin depth by

$$\lambda = 2\pi\delta$$

The field measured at the surface due to reflecting surfaces at a depth equivalent to approximately one half-wave length for example, will be attenuated by a

factor  $e^{-6.28}$ . In addition the secondary fields are reduced by the finite reflective coefficient of the anomaly and possibly out of phase relations.

The velocity of phase propagation of the wave within a medium of conductivity  $\sigma$ , is  $f\lambda$ . Thus for a conductivity  $\sigma = 10^{-1}$  mhos and  $f = 10^{-1}$  Hz, the velocity is

$$v = \sqrt{\frac{4\pi f}{\mu_0 \sigma}} \approx 3 \times 10^3 \text{ m/sec}.$$

As a consequence a plane wave incident at an oblique angle will propagate perpendicular to the surface as though it were at normal incidence.

### III-C. Plane Wave Incident on a Uniform Earth

In the case of a uniform earth where  $\sigma$  is constant, the E and H-fields are related in Eqn. (2-1a)

$$\nabla \times \vec{E} = -j\omega\mu\vec{H}.$$

Considering the E-field polarized in the x-direction and the H-field in the y-direction, the above relation can be rewritten as

$$j\omega\mu H_y = -\frac{\partial E_x}{\partial z} = K_1 E_x$$

and

$$E_x = \sqrt{j\omega\mu\rho} H_y \quad (2-7)$$

where  $\rho = 1/\sigma$  is the resistivity of the medium.

The impedance  $Z$  is defined

$$Z = \frac{E_x}{H_y} = \sqrt{j\omega\mu\rho} = \sqrt{\omega\mu\rho} e^{j\pi/4} \quad (2-8)$$

Thus the phase angle between the  $E$  and  $H$ -fields for a uniform half space is  $\phi = \pi/4$ . The resistivity in this case of the uniform earth obtained from Eqn. (2-7) or (2-8) is

$$\rho = \frac{1}{\omega\mu} |Z|^2 = \frac{1}{\omega\mu} \left| \frac{E_x}{H_y} \right|^2 \quad (2-9)$$

#### II-D. Apparent Resistivity of Layered Earth

For a uniform earth as discussed in the previous section, the true resistivity of the medium can be directly deduced from the measured fields. For a more general case, however, the true resistivity of the medium is a more complex function of the measured fields.

(1) Single layered earth: This model consists of a uniform layer of thickness  $h$  and resistivity  $\rho_1$ , overlying a half space with resistivity  $\rho_2$ , as shown in Fig. (2-2a). Choosing the source  $E$ -field polarized in the  $x$ -direction, the plane wave equation can be written (from Eqn. (2-3)) as

$$\frac{\partial^2 H_y}{\partial z^2} = K_i^2 H_y$$

A solution has the form

$$H_Y(z) = A_i e^{-K_i(z-z_{i-1})} + B_i e^{+K_i(z-z_{i-1})}$$

$$E_X(z) = K_i \rho_i [A_i e^{-K_i(z-z_{i-1})} - B_i e^{+K_i(z-z_{i-1})}]$$

in each layer. The coefficient  $A_i$  is associated with a positive going (downward) transmitted wave, and  $B_i$  with the upward reflected wave.  $z_{i-1}$  is the depth to the upper surface of the  $i$ -th layer. Thus in medium 1, where  $z_{i-1} = z_0 = 0$

$$H_Y(z) = A_1 e^{-K_1 z} + B_1 e^{+K_1 z}$$

$$E_X(z) = K_1 \rho_1 [A_1 e^{-K_1 z} - B_1 e^{+K_1 z}]$$

while in medium 2 where  $z_{i-1} = z_1 = h$  (the thickness of the first layer)

$$H_Y(z) = A_2 e^{-K_2(z-h)}$$

$$E_X(z) = K_2 \rho_2 A_2 e^{-K_2(z-h)}$$

where  $B_2 = 0$  since there is no upward reflected wave.

From the boundary condition that the tangential components of the field are equal at  $z = h$ ,

$$A_1 = \frac{1}{2}(1 + \frac{K_2 \rho_2}{K_1 \rho_1}) A_2 e^{-K_1 h}$$

$$B_1 = \frac{1}{2}(1 - \frac{K_2 \rho_2}{K_1 \rho_1}) A_2 e^{-K_1 h}$$

Thus, the surface impedance is obtained

$$\begin{aligned} Z_o &= \frac{E_x(0)}{H_y(0)} = \rho_1 K_1 \left( \frac{A_1 - B_1}{A_1 + B_1} \right) \\ &= \rho_1 K_1 \left( \frac{\sqrt{\rho_2}(1+\alpha) + \sqrt{\rho_1}(1-\alpha)}{\sqrt{\rho_2}(1-\alpha) + \sqrt{\rho_1}(1+\alpha)} \right) \end{aligned} \quad (2-10)$$

$$\text{where } \alpha = e^{-2K_1 h}$$

The apparent resistivity is defined by

$$\rho_a = \frac{1}{\omega \mu} [Z_o]^2 = \rho_1 \left( \frac{A_1 - B_1}{A_1 + B_1} \right)^2 \quad (2-11)$$

This is the apparent resistivity of a single-layer earth, which differs from the true resistivity  $\rho_1$  by a factor  $(A_1 - B_1)/(A_1 + B_1)^2$ . This factor is a function of the earth parameters, i.e., the resistivities of the two differently conductive regions and the thickness of the overlying layer.

(2) n-layered earth: A typical n-layered earth model is as shown in Fig. (2-2b). The solutions for the m-th layer have the form

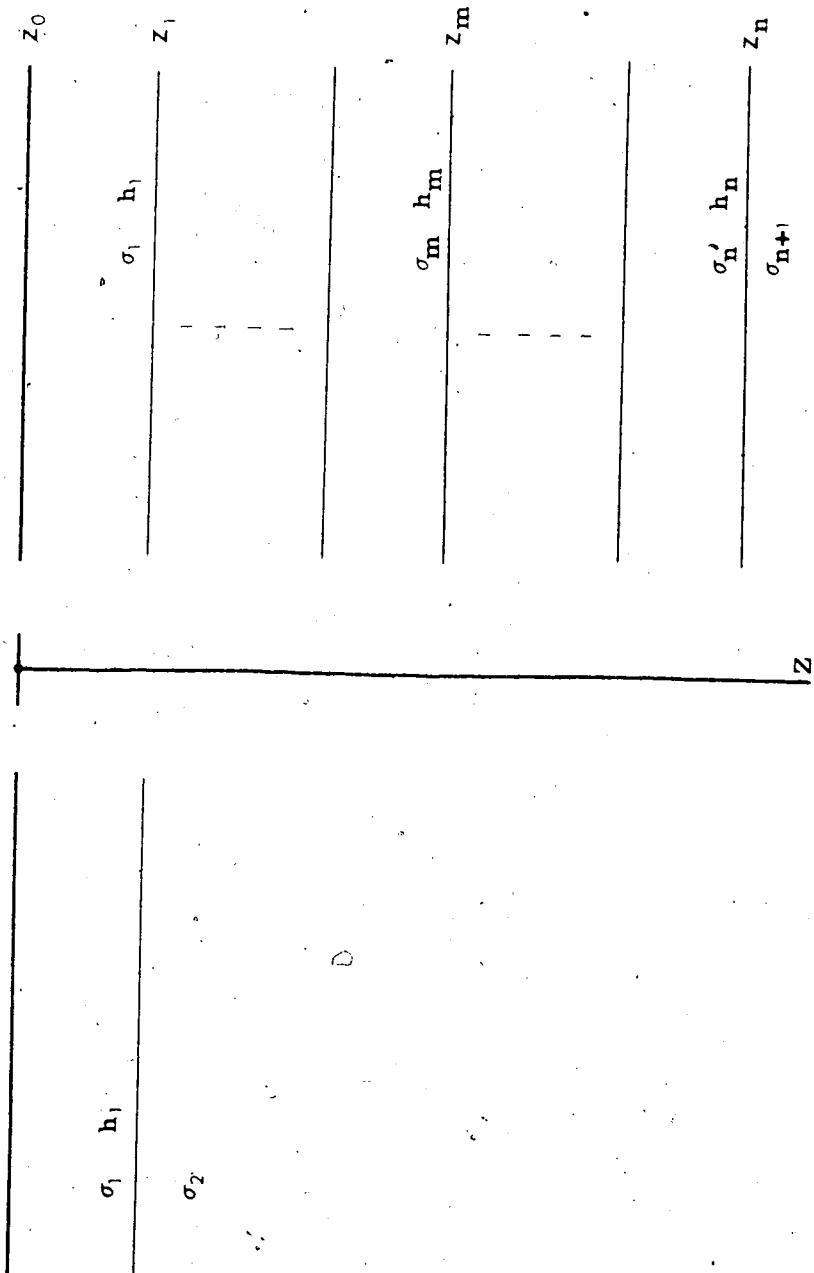


Fig. (2-2a). Single-layer model.

Fig. (2-2b).  $n$ -layer model.

$$H_{Y,m}(z) = A_m e^{-K_m(z-z_{m-1})} + B_m e^{K_m(z-z_{m-1})}$$

$$E_{x,m}(z) = K_m [A_m e^{-K_m(z-z_{m-1})} - B_m e^{K_m(z-z_{m-1})}]$$

Applying the boundary conditions at  $z = z_m$ , the transmission and reflection coefficients can be expressed by

$$A_m = \frac{1}{2} [(1+\gamma)A_{m+1} + (1-\gamma)B_{m+1}] e^{K_m h_m}$$

$$B_m = \frac{1}{2} [(1-\gamma)A_{m+1} + (1+\gamma)B_{m+1}] e^{-K_m h_m}$$

where  $\gamma = (K_{m+1}\rho_{m+1})/(K_m\rho_m)$ , and  $h_m = z_m - z_{m-1}$  is the thickness of the  $m$ -th layer. The impedance of this  $m$ -th layer is defined

$$Z_m = \frac{E_{x,m}}{H_{Y,m}} = K_m \rho_m \frac{K_m \rho_m \tanh(K_m \rho_m)}{K_m \rho_m + Z_{m+1} \tanh(K_m \rho_m)} + Z_{m+1} \quad (2-12)$$

The surface impedance is

$$Z_o = \frac{E_x(0)}{H_Y(0)} = K_1 \rho_1 \frac{A_1 - B_1}{A_1 + B_1} = K_1 \rho_1 \frac{K_1 \rho_1 \tanh(K_1 \rho_1) + Z_2}{K_1 \rho_1 + Z_2 \tanh(K_1 \rho_1)} \quad (2-13a)$$

and the apparent resistivity is

$$\rho_a = \frac{1}{\omega \mu} |Z_o|^2 = \rho_1 \left| \frac{A_1 - B_1}{A_1 + B_1} \right|^2 \quad (2-13b)$$

where  $A_1$ ,  $B_1$ , and in turn  $|A_1 - B_1/A_1 + B_1|^2$  are functions of the resistivities and thickness of all the layers.

### II-E. Tensor Impedance Analysis

In the previous section, the case of a uniform or piecewise uniform medium was discussed in which the field component variations were a function of one variable only and the resistivity was a scalar. In a more general case both the components of the fields and the resistivity are functions of 3 spatial variables.

In this case one defines an impedance tensor [Z] such that

$$\vec{[E]} = [Z] \vec{[H]} \quad (2-14)$$

where  $\vec{[E]}$  and  $\vec{[H]}$  are the electric and magnetic field vectors written as column matrices and [Z] is in general a  $3 \times 3$  impedance matrix relating each of the components of E to 3 components of H. The components of [Z] are point functions of earth parameters together with the appropriate frequency dependence. While analytic solutions are possible only in very special cases, it is of interest to develop general expressions for the electromagnetic fields in terms of the earth parameters.

Assuming the earth behaves like a linear system, the secondary field and in turn the total field is linearly related to the primary magnetic field,  $H_p(r)$ .

At the surface,  $x, y, z = 0$

$$[\vec{H}_p(0)] = (H_{px}(0), H_{py}(0), 0) \quad (2-15)$$

thus, the total H-field within the earth can then be related to the primary field at the surface

$$[\vec{H}(r)] = [\phi(r)] [\vec{H}_p(0)] \quad (2-16)$$

where  $[\phi(r)]$  is generally a  $3 \times 3$  matrix. However, since  $H_{pz}(0) = 0$ , only the  $3 \times 2$  part is present. The elements of  $[\phi(r)]$  are functions of the transmission and reflection coefficients, which are in turn, of the position  $r$  and other earth parameters.

The E-field at position  $r$  can be obtained from a generalized form of Ohms law, appropriate to an anisotropic earth, and Maxwell's equations,

$$[\vec{E}(r)] = [\rho(r)] [\vec{J}(r)] = [\rho(r)] [\nabla \times \vec{H}(r)] \quad (2-17)$$

where  $\rho(r)$  is a  $3 \times 3$  matrix, its elements representing the tensor coefficients of the resistivity of the earth; and  $J$  is the current density. At the surface, the components of  $J$  in terms of  $H_p(0)$  can be obtained by substituting  $H(r)$  from Eqn. (2-16) into  $J(r) = \nabla \times H(r)$  and evaluating at the surface

$$J_x(0) = (y\phi_{31} - z\phi_{21}) H_{px}(0) + (y\phi_{32} - z\phi_{22}) H_{py}(0) \quad (2-18a)$$

$$J_y(0) = (z\phi_{11} - x\phi_{31}) H_{px}(0) + (z\phi_{12} - x\phi_{32}) H_{py}(0) \quad (2-18b)$$

$$J_z(0) = (x\phi_{21} - y\phi_{11})H_{px}(0) + (x\phi_{22} - y\phi_{12})H_{py}(0) \quad (2-18c)$$

$$\text{where } q_{ij} = \left(\frac{\partial}{\partial q} \phi_{ij}(r)\right)_{r=0}.$$

Since no conduction current flows in the air region, the continuity condition at the air-earth interface requires that in the earth at or near the surface,  $J_z$  vanishes. Thus the E-field at the surface can be expressed in terms of  $H_p(0)$  by

$$\begin{aligned} E_x(0) &= \rho_{11} J_x(0) + \rho_{12} J_y(0) \\ &= \{\rho_{11} (y\phi_{31} - z\phi_{21}) + \rho_{12} (z\phi_{11} - x\phi_{31})\} H_{px}(0) \\ &\quad + \{\rho_{11} (y\phi_{32} - z\phi_{22}) + \rho_{12} (z\phi_{12} - x\phi_{32})\} H_{py}(0) \quad (2-19a) \end{aligned}$$

$$\begin{aligned} E_y(0) &= \rho_{21} J_x(0) + \rho_{22} J_y(0) \\ &= [\rho_{21} (y\phi_{31} - z\phi_{21}) + \rho_{22} (z\phi_{11} - x\phi_{31})] H_{px}(0) \\ &\quad + [\rho_{21} (y\phi_{32} - z\phi_{22}) + \rho_{22} (z\phi_{12} - x\phi_{32})] H_{py}(0) \quad (2-19b) \end{aligned}$$

$$\begin{aligned} E_z(0) &= \rho_{31} J_x(0) + \rho_{32} J_y(0) \\ &= [\rho_{31} (y\phi_{31} - z\phi_{21}) + \rho_{32} (z\phi_{11} - x\phi_{31})] H_{px}(0) \\ &\quad + [\rho_{31} (y\phi_{32} - z\phi_{22}) + \rho_{32} (z\phi_{12} - x\phi_{32})] H_{py}(0). \quad (2-19c) \end{aligned}$$

Eqns. (2-19) define the primary impedance relating the primary H-field to the E-field at the surface, by

$$[E(0)] = [Z_p] [H_p(0)] \quad (2-20)$$

where  $[Z_p]$  is a  $2 \times 3$  matrix as indicated in Eqns. (2-19).

The surface impedance, which defines the relationship between the total H-field and the E-field at the surface, can be determined as follows. From Eqn. (2-16), the inverse matrix of  $[\phi(0)]$  can be generated and evaluated at the surface,

$$[H_p(0)] = [\phi(0)]^{-1} [H(0)] \quad (2-21)$$

where  $[\phi(0)]^{-1}$  is defined by  $[\phi(0)]^{-1} [\phi(0)] = [I]$ .

Substituting Eqn. (2-21) to (2-20), the total H-field and the E-field at the surface are related by

$$[E(0)] = [Z_p] [\phi(0)]^{-1} [H(0)] \quad (2-22)$$

Thus, the surface impedance function for the three dimensional earth is

$$[Z] = [Z_p] [\phi(0)]^{-1} \quad (2-23)$$

where  $[Z]$  is a  $3 \times 3$  matrix. An explicit form of E-H fields relation can then be expressed in terms of the elements of  $[Z]$ .

$$E_x = Z_{11}H_x + Z_{12}H_y + Z_{13}H_z \quad (2-24a)$$

$$E_y = Z_{21}H_x + Z_{22}H_y + Z_{23}H_z \quad (2-24b)$$

$$E_z = Z_{31}H_x + Z_{32}H_y + Z_{33}H_z \quad (2-24c)$$

### 11-F. Examples of the Impedance Functions

(1) Uniform Earth:  $\phi(x, y, z) = \sigma_0 = 1/\mu_0$ .

In a uniform half-space the total H-field is the transmitted primary H-field. If the surface value of the field is denoted by

$$\vec{H}_p(0) = (H_{x0}; H_{y0}, 0)$$

then the total field at  $z > 0$  is

$$\vec{H}(z) = \vec{H}_p(z) = (H_{x0}e^{-Kz}, H_{y0}e^{-Kz}, 0)$$

for which,

$$[\phi(r)] = \begin{pmatrix} e^{-Kz} & 0 \\ 0 & e^{-Kz} \\ 0 & 0 \end{pmatrix} \quad (2-25)$$

where  $K = \sqrt{j\omega\mu\sigma_0}$ .

Thus, at the surface, the inverse of  $[\phi(0)]$  is

$$[\phi(0)]^{-1} = \begin{pmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix}$$

The resistivity of a uniform earth can be denoted by

$$[\rho] = \begin{pmatrix} \rho & 0 & 0 \\ 0 & \rho & 0 \\ 0 & 0 & \rho \end{pmatrix}$$

and so that

$$[z_p] = \begin{pmatrix} 0 & \rho K \\ -\rho K & 0 \\ 0 & 0 \end{pmatrix}$$

Therefore, the surface impedance tensor is,

$$[z] = [z_p] [\phi(0)^{-1}] = \begin{pmatrix} 0 & \rho K & 0 \\ -\rho K & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix} \quad (2-26)$$

Eqn. (2-26) indicates that

$$z_{12} = -z_{21} = \sqrt{j\omega\mu\rho} \quad (2-27a)$$

$$z_{ij} = 0 \quad i, j \neq 1, 2 \quad (2-27b)$$

i = j

The result given in Eqn. (2-27) is consistent with that given in Eqn. (2-8), for a uniform earth.

## (2) Layered earth:

Referring the model to (II-D) and letting the primary H-field at the surface be denoted by

$$\mathbf{H}_P(0) = (A_x, A_y, 0)$$

the total H-field below the surface can then be written as

$$H_x(r) = A_x e^{-K_1 z} + B_x e^{K_1 z}$$

$$H_y(r) = A_y e^{-K_1 z} + B_y e^{K_1 z}$$

Comparing the above expressions with Eqn. (2-16), it is clearly seen that  $[\phi(r)]$  has the form

$$[\phi(r)] = \begin{pmatrix} e^{-K_1 z} + \frac{B_x}{A_x} e^{K_1 z} & 0 & 0 \\ 0 & e^{-K_1 z} + \frac{B_y}{A_y} e^{K_1 z} & 0 \\ 0 & 0 & 0 \end{pmatrix} \quad (2-28)$$

where  $K_1 = \sqrt{j\omega\mu/\rho_1}$  and  $\rho_1$  is the resistivity of the first layer, and  $B_x, B_y$  represent the effective reflection coefficients from all discontinuities. Since the parameters for each layer are constant within the layer

$$\frac{B_x}{A_x} = \frac{B_y}{A_y} = \frac{B}{A}$$

By following the procedure in the last paragraph, one can obtain

$$[\phi(0)]^{-1} = \begin{pmatrix} \frac{1}{1 + \frac{B}{A}} & 0 & 0 \\ 0 & \frac{1}{1 + \frac{B}{A}} & 0 \\ 0 & 0 & \frac{1}{1 + \frac{B}{A}} \end{pmatrix}$$

$$[Z_p] = \begin{pmatrix} 0 & \rho_1 K_1 (1 - \frac{B}{A}) \\ -\rho_1 K_1 (1 - \frac{B}{A}) & 0 \\ 0 & 0 \end{pmatrix}$$

and the surface impedance tensor

$$[Z] = [Z_p] [\phi(0)]^{-1} = \begin{pmatrix} 0 & \rho_1 K_1 (\frac{A - B}{A + B}) & 0 \\ -\rho_1 K_1 (\frac{A - B}{A + B}) & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix}$$

(2-29)

which is consistent with the result given in Eqn. (2-13a) or (2-10).

$$Z_{12} = -Z_{21} = \rho_1 K_1 \frac{A - B}{A + B} \quad (2-30)$$

### (3) Vertical fault - two dimensional model:

A vertical fault is shown schematically in Fig.

(2-3), the x-axis is along the interface of the fault, or the strike direction.  $\rho_i$  ( $i = 1, 2$ ) is the resistivity

of the region i.

Let the primary H-field at the surface be

$$\vec{H}_p(0) = (A_x, A_y, 0)$$

The total H-field can then be written

$$H_x(r) = A_x(e^{-K_i z} + R_{xx}(y, z))$$

$$H_y(r) = A_y(e^{-K_i z} + R_{yy}(y, z))$$

$$H_z(r) = A_y R_{zy}(y, z)$$

where R's represent the normalized reflection coefficients and depend upon y and z in this case. The associated functions are

$$[\psi(r)] = \begin{pmatrix} e^{-K_i z} + R_{xx} & 0 \\ 0 & e^{-K_i z} + R_{yy} \\ 0 & R_{zy} \end{pmatrix} \quad (2-31)$$

$$[\phi(0)]^{-1} = \begin{pmatrix} \frac{1}{1 + R_{xx}} & 0 & 0 \\ 0 & \frac{1}{1 + R_{yy}} & 0 \end{pmatrix}_{z=0}$$

$$[Z_p] = \begin{pmatrix} 0 & \rho_i(K_i - z R_{yy} + y R_{zy}) \\ -\rho_i(K_i - z R_{xx}) & 0 \\ 0 & 0 \end{pmatrix}$$

where  $\frac{\partial}{\partial q} [R_{uv}(r)]_{z=0}$

and the surface impedance tensor

$$[Z] = \begin{pmatrix} 0 & \frac{\rho_i (K_i - z R_{yy} + y R_{zy})}{1 + R_{yy}(0)} & 0 \\ -\frac{\rho_i (K_i - z R_{xx})}{1 + R_{xx}(0)} & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix} \quad (2-32)$$

$$z_{12} = \frac{\rho_i (K_i - z R_{yy} + y R_{zy})}{1 + R_{yy}(0)} \quad (2-33a)$$

$$z_{21} = -\frac{\rho_i (K_i - z R_{xx})}{1 + R_{xx}(0)} \quad (2-33b)$$

$$z_{ij} = 0 \quad i, j \neq 1, 2 \quad (2-33c)$$

$$i = j$$

The result  $z_{12} \neq z_{21}$  shows the anisotropic behavior of the impedance tensor elements in the two dimensional case.

#### II-G. Reduction of a Three-Dimensional Problem to a Formal Two-Dimensional Formulation

The following analysis only applies for the case where the principle directions of resistivity are mutually orthogonal and thus a cartesian coordinate

system is applicable. The 9 impedance coefficients shown in Eqn. (2-24), are reduced to 4, generally non-independent, coefficients. Amongst all the components of the fields, only  $H_{px}$  and  $H_{py}$  are independent, while the others,  $H_x$ ,  $H_y$ ,  $H_z$ ,  $E_x$ ,  $E_y$  are linearly related to the two independent components. The mathematical proof can be carried out by rewriting Eqn. (2-16)

$$H_x = \phi_{11} H_{px} + \phi_{12} H_{py} \quad (2-34a)$$

$$H_y = \phi_{21} H_{px} + \phi_{22} H_{py} \quad (2-34b)$$

$$H_z = \phi_{31} H_{px} + \phi_{32} H_{py} \quad (2-34c)$$

It is obvious that  $H_x$ ,  $H_y$ , and  $H_z$  are not independent; and each one of the three components are dependent on the other two. To put  $H_z$  in terms of  $H_x$  and  $H_y$  for instance,  $H_{px}$  and  $H_{py}$  can first be obtained in terms of  $H_x$  and  $H_y$  from Eqn. (2-34a,b), and then substituted in the Eqn. (2-34c)

$$H_z = \alpha_x H_x + \alpha_y H_y \quad (2-35)$$

where

$$\alpha_x = \frac{\phi_{31}\phi_{22} - \phi_{32}\phi_{21}}{\phi_{11}\phi_{22} - \phi_{12}\phi_{21}}$$

$$\alpha_y = \frac{\psi_{32}\phi_{11} - \psi_{31}\phi_{12}}{\phi_{11}\phi_{22} - \phi_{12}\phi_{21}}$$

An alternative proof can be achieved by considering a model in which the earth resistivity  $\rho(r)$  approaches some constant value in the region near the surface. At the surface,  $J_z(0) = 0$ , and thus  $E_z(0) = 0$ . Substituting  $E_z(0) = 0$  into Eqn. (2-24c), an equivalent relationship to Eqn. (2-35) is obtained,

$$H_z(0) = \alpha_x H_x(0) + \alpha_y H_y(0) . \quad (2-36)$$

Substituting  $H_z$  from Eqn. (2-35) or (2-36) into (2-24),

$$E_x = z_{11} H_x + z_{12} H_y$$

$$E_y = z_{21} H_x + z_{22} H_y$$

where

$$z_{11} = z_{11} + z_{31} \alpha_x$$

$$z_{12} = z_{12} + z_{32} \alpha_y$$

$$z_{21} = z_{21} + z_{31} \alpha_x$$

$$z_{22} = z_{22} + z_{32} \alpha_y$$

For convenience, the unprimed  $z_{ij}$  with  $i,j = x,y$  will be used in the following text for surface impedance for two-dimensional problem, unless otherwise indicated.

$$E_x = Z_{xx} H_x + Z_{xy} H_y \quad (2-37a)$$

$$E_y = Z_{yx} H_x + Z_{yy} H_y \quad (2-37b)$$

The vertical component of magnetic field  $H_z$  is a secondary field, and is related to the total H-field components by Eqn. (2-35) or (2-36). At the surface

$$H_z(0) = \alpha_x H_x(0) + \alpha_y H_y(0)$$

The relation between  $H_z$  and the E-field at the surface can be written

$$H_z(0) = Y_x E_x(0) + Y_y E_y(0) \quad (2-38)$$

which defines the surface admittances  $[Y_i]$ . Eqns. (2-35), (2-37), and (2-38) give the following relations

$$\alpha_x = Y_x Z_{xx} + Y_y Z_{yx} \quad (2-39a)$$

$$\alpha_y = Y_x Z_{xy} + Y_y Z_{yy} \quad (2-39b)$$

As a function of frequency, the impedance and admittance are related to the field components at a given point at the surface for any polarization of the source and for any arbitrary earth conductivity distribution, and hence they contain diagnostic information for geological interpretation.

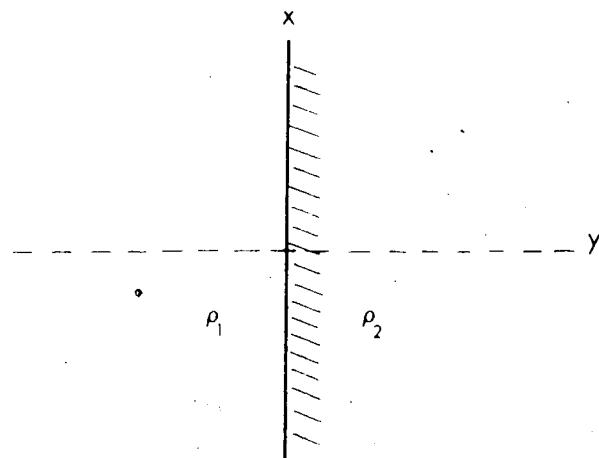


Fig. (2-3). Vertical fault.

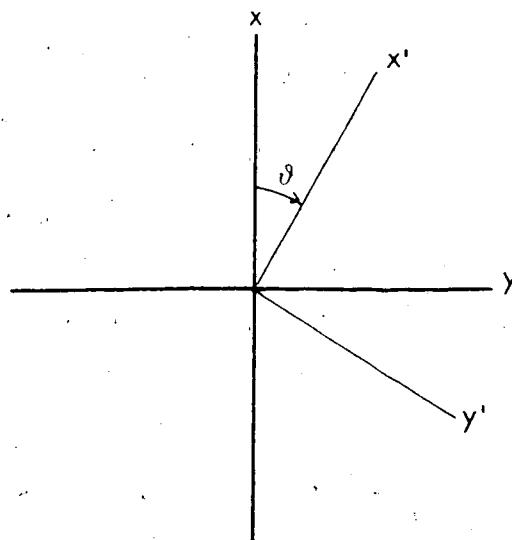


Fig. (2-4). Rotated coordinates.

### III-B. Orientation Behavior

Surface impedance and admittance elements  $Z_{ij}$ ,  $Y_i$ , and  $\mu_i$  which are functions of the earth parameters, contain the information for determination of geological substructure including their directional behavior.

Taking the 2-dimensional problem as an example; as shown in Eqn. (2-33), if the axes of the coordinate system coincide with the principal directions, i.e. the strike direction and the perpendicular,  $Z_{xx}$  and  $Z_{yy}$  vanish. Physically, in either of the principal directions, a component of the E-field is correlated only with its orthogonal component of the H-field; thus the induction parameters  $Z_{xx}$  and  $Z_{yy}$  are zero.

However, if the axes of the coordinates lie in any other directions, where the effective resistivity is no longer isotropic, each component of the E-field is induced by a combination of all components of the H-field, and the magnitude of each  $Z_{ij}$  is a measure of the induction effect between the corresponding E and H components.

In order to examine the directional behavior of  $Z_{ij}$ , one can rotate the fields E and H from an arbitrary direction x-y, i.e. the measuring direction, to a new direction x'-y' as shown in Fig. (2-4), where

$$E'_x = \cos\theta E_x + \sin\theta E_y$$

$$E'_y = -\sin\theta E_x + \cos\theta E_y$$

This defines the rotation matrix which is orthogonal

$$[R] = \begin{pmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{pmatrix} \quad (2-40)$$

Orthogonality of the matrix enables one to write

$$[E'] = [R] [E]$$

$$[H'] = [R] [H]$$

$$[E'] = [Z'] [H']$$

and the rotated impedance matrix is obtained

$$[Z'] = [R] [Z] [R]^{-1} \quad (2-41)$$

Thus, the rotated surface impedance elements  $Z'_{ij}$  are related to those in the measuring direction

$$Z'_{xx} = \frac{1}{2}[(Z_{xx} + Z_{yy}) + (Z_{xx} - Z_{yy})\cos 2\theta + (Z_{xy} + Z_{yx})\sin 2\theta] \quad (2-42a)$$

$$Z'_{yy} = \frac{1}{2}[(Z_{xx} + Z_{yy}) - (Z_{xx} - Z_{yy})\cos 2\theta - (Z_{xy} + Z_{yx})\sin 2\theta] \quad (2-42b)$$

$$Z'_{xy} = \frac{1}{2}[(Z_{xy} - Z_{yx}) + (Z_{xy} + Z_{yx})\cos 2\theta - (Z_{xx} + Z_{yy})\sin 2\theta] \quad (2-42c)$$

$$Z'_{yx} = \frac{1}{2}[-(Z_{xy} - Z_{yx}) + (Z_{xy} + Z_{yx})\cos 2\theta - (Z_{xx} + Z_{yy})\sin 2\theta] \quad (2-42d)$$

where  $z'_{ij}(\theta)$  are functions of the rotation angle  $\theta$ ,  
and also have the following relations

$$z'_{yy}(\theta) = z'_{xx}(\theta+90^\circ) \quad (2-43)$$

$$z'_{yx}(\theta) = -z'_{xy}(\theta+90^\circ) \quad (2-44)$$

$$z'_{xy} - z'_{yx} = z_{xy} - z_{yx} = 2z_1 \quad (2-45)$$

$$z'_{xx} + z'_{yy} = z_{xx} + z_{yy} = 2z_2 \quad (2-46)$$

where  $z_1$  and  $z_2$  are new constants. As shown by Sims (1969) the loci of  $z'_{ij}(\theta)$  in the complex plane are in general elliptical as shown in Fig. (2-5). This graphical representation is very helpful in visualizing the source behavior as a function of rotation and in the explanation of dimensionality.

For convenience all ellipses of  $z'_{ij}$  in Fig. (2-5) are shown to the same scale and correspond to the same angle  $\theta_0$  between the measuring axis and the major axis of the resistivity structure.

$$\text{major axis} = z'_{xy}(\theta_0) + z'_{yx}(\theta_0)$$

$$\text{minor axis} = z'_{xx}(\theta_0) + z'_{yy}(\theta_0)$$

The angle of the major axis can be found by maximizing either

$$Z'_{xy}(\psi) - Z'_{11} \text{ or } Z'_{xy}(\psi) + Z'_{yx}(\psi) \\ \frac{d}{d\psi} \left| Z'_{xy}(\psi) + Z'_{yx}(\psi) \right| = 0 \quad (2-47a)$$

The 4 possible solutions of Eqn. (2-47a), corresponding to the major and minor axis intercepts, are

$$\phi = \frac{1}{4} \tan^{-1} \frac{2 \operatorname{Re}[(Z_{xx} - Z_{yy})^* (Z_{xy} + Z_{yx})]}{|Z_{xx} - Z_{yy}|^2 - |Z_{xy} + Z_{yx}|^2} \quad (2-47b)$$

This result is ambiguous since either of the principle directions of resistivity is found. However, this ambiguity can be removed by apparent resistivity information.

The rotated admittance elements can also be obtained

$$Y'_x(\psi) = Y_x \cos \psi + Y_y \sin \psi \quad (2-48a)$$

$$Y'_y(\psi) = -Y'_x \cos \psi + Y'_y \sin \psi \quad (2-48b)$$

and are related to

$$Y'_x(\psi) = Y'_y(\psi + 90^\circ)$$

The locus of  $Y'_x(\psi)$  is also an ellipse as shown in Fig. (2-5). The ellipse is centered on the origin and oriented at the angle  $\theta_{ze}$ . The rotation period for this locus is  $360^\circ$  in  $\psi$ , instead of  $180^\circ$  as for  $Z'_{ij}$ .

The major and minor axes are

$$\text{major axis} = 2Y'_x(\theta_{ze}) \quad (2-49a)$$

$$\text{minor axis} = 2Y'_y(\theta_{ze} + 90^\circ) \quad (2-49b)$$

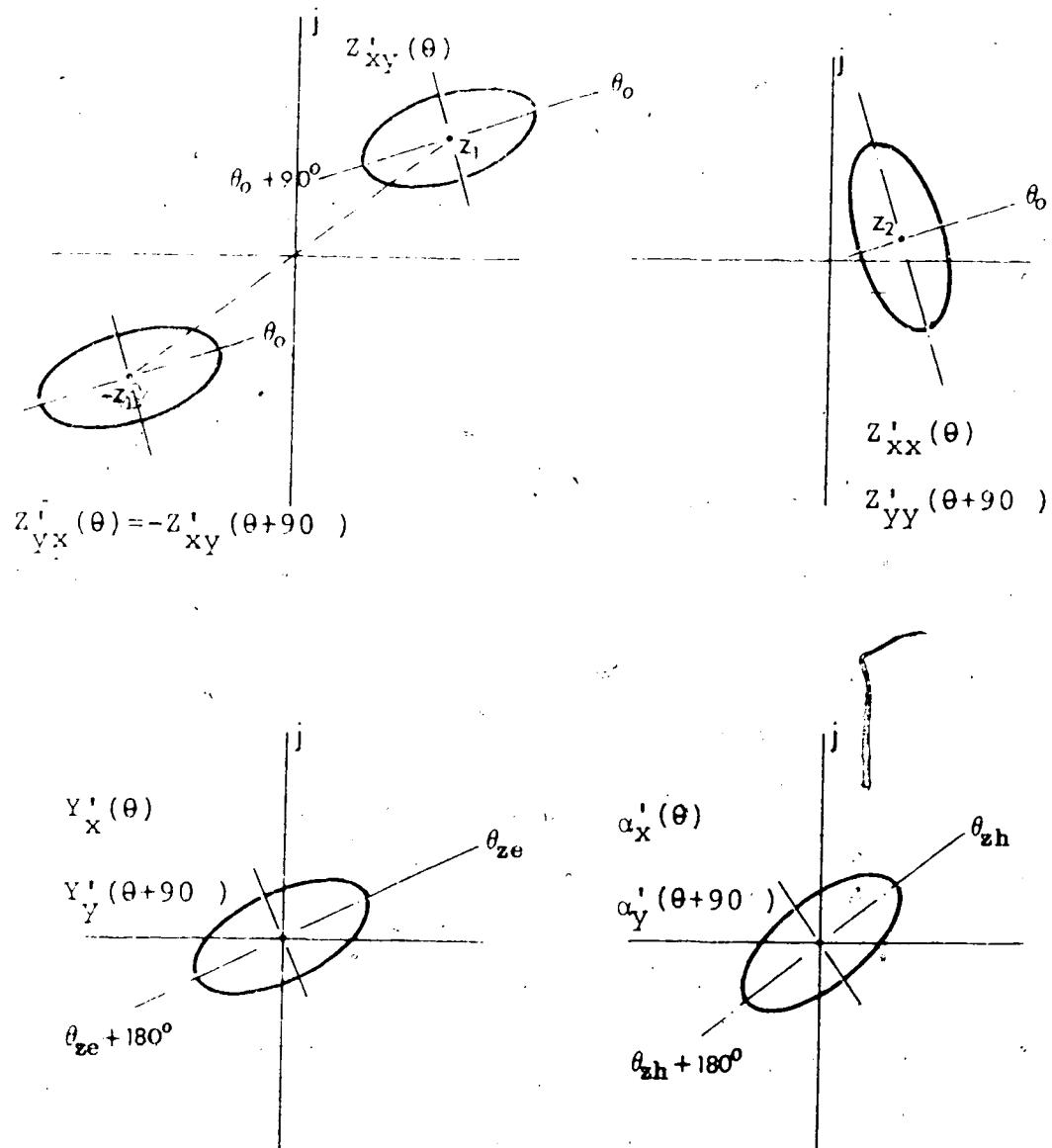


Fig. (2-5).  $Z_{ij}^j(\theta)$ ,  $Y_i^j(\theta)$  and  $\alpha_i^j(\theta)$  loci in complex plane. ( $\theta=0$  corresponding to the measuring axis not shown).

Using the procedure as used in dealing with  $z'_{ij}$ ,  
the angle  $\theta_{ze}$  is found

$$\theta_{ze} = \frac{1}{2} \tan^{-1} \frac{2 \operatorname{Re}(Y_x^* Y_y)}{|Y_x|^2 - |Y_y|^2} \quad (2-50)$$

The rotated  $\alpha'_i$  behaves in a similar fashion to  
the  $y'_i$ , except that  $\alpha'_x \rightarrow y'_y$  and  $\alpha'_y \rightarrow y'_x$ .

### III-I. Special Cases and the Indications of Dimensionality

#### (1) One-dimensional case:

When the earth conductivity is uniform or is a function of depth only, the surface impedance is invariant with rotation angle  $\theta$ . Eqns. (2-27), (2-30), and (2-42) give these results

$$z'_{xx} = z'_{yy} = 0 \quad (2-51a)$$

$$z'_{xy} = -z'_{yx} = z_1 \quad (2-51b)$$

where  $z_1$  is a complex scalar for all  $\theta$ .

The rotation loci for  $z'_{ij}$ , as shown in Fig. (2-6), reduce to points in the complex plane centered at  $\pm z_1$  for  $z'_{xy}$  and  $z'_{yx}$ , and at the origin for  $z'_{xx}$  and  $z'_{yy}$ .

In the one-dimensional case, no vertical magnetic field exists ( $H_z = 0$ ), so that  $y'_x = y'_y = 0$ . The rotation loci for  $y'_i$  and  $\alpha'_i$  are also point ellipses

centered on the origin as shown in Fig. (2-6).

(2) Two-dimensional case:

As illustrated in Eqns. (2-33) for a two-dimensional model,  $z'_{xx}$  and  $z'_{yy}$  vanish along the principal axes, i.e.  $\theta = \theta_0, \theta_0 + 90^\circ$ . Eqn. (2-46) also shows that the equation,  $z'_{xx} + z'_{yy} = 2z_2$ , is invariant with rotation angle  $\theta$ , so that, for a two-dimensional model  $z'_{xx} + z'_{yy} = 2z_2 = 0$ , and thus,

$$z'_{xx} = -z'_{yy} \quad \text{for all } \theta \quad (2-52a)$$

$$z'_{xx} = z'_{yy} = 0 \quad \text{for } \theta = \theta_0, \theta_0 + 90^\circ \quad (2-52b)$$

The loci for  $z'_{xx}$  and  $z'_{yy}$  are straight line ellipses centered on the origin. The loci for  $z'_{xy}$  and  $z'_{yx}$  are also straight line ellipses centered at  $\pm z_1$ , as shown in Fig. (2-7).

A vertical magnetic field  $H_z$  exists when the primary magnetic field has a component perpendicular to the strike. Elements of  $Y'_i$  and  $\alpha'_i$  repeat for each  $180^\circ$  increment in  $\theta$ . Suppose  $y'$ -axis is perpendicular to the strike (referring to Figs. (2-3) and (2-4)),  $Y'_i$  and  $\alpha'_i$  are

$$Y'_x \approx \alpha'_y = 0 \quad \theta = (\theta_{ze} + 90^\circ \pm 180^\circ) \quad (2-53a)$$

$$Y'_y = \alpha'_x = 0 \quad \theta = (\theta_{ze} \pm 180^\circ) \quad (2-53b)$$

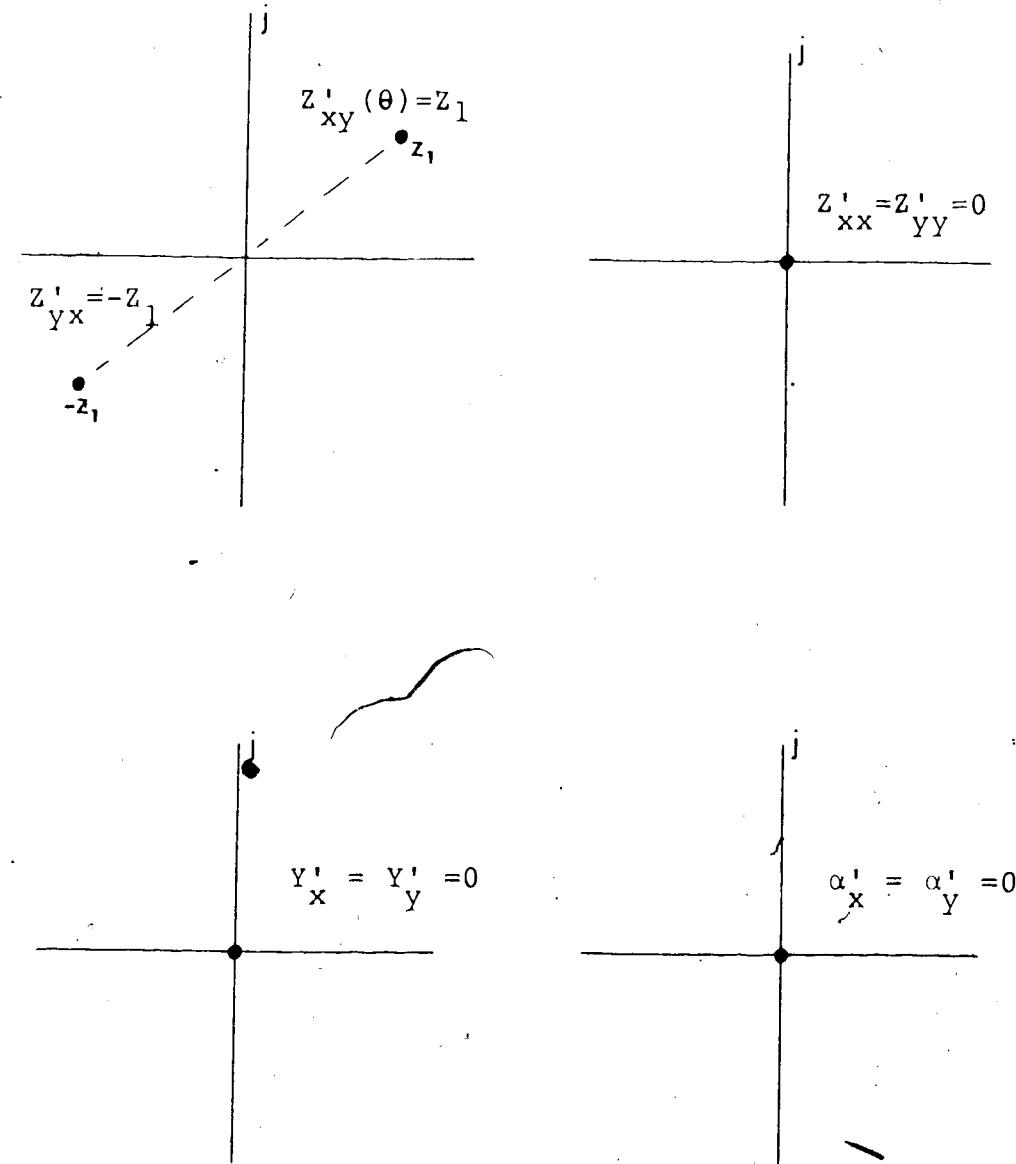


Fig. (2-6).  $Z_{ij}'(\theta)$ ,  $Y_i'(\theta)$  and  $\alpha_i'(\theta)$  loci for one-dimensionality.

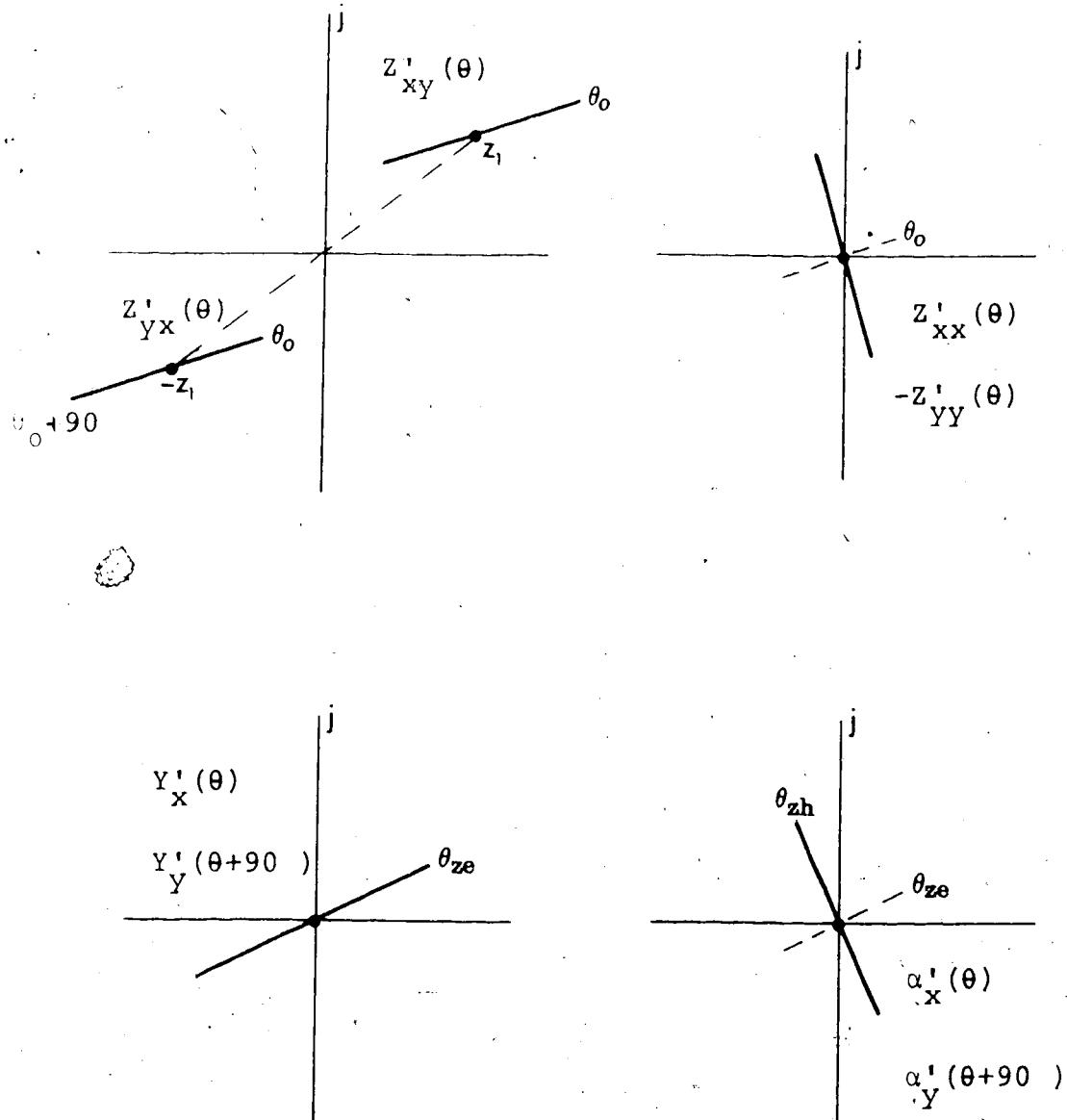


Fig. (2-7).  $z_{ij}^j(\theta)$ ,  $y_{ij}^j(\theta)$  and  $\alpha_{ij}^j(\theta)$  loci for  
two-dimensionality.

$$\theta_{ze} = \theta_{zh} \pm 90^\circ . \quad (2-53c)$$

The loci of  $y'_i(\theta)$  and  $\alpha'_i(\theta)$  are straight line ellipses centered on the origin as shown in Fig. (2-7).

The direction corresponding to the angle  $\theta_o$  determined by  $z'_{ij}$  can be one of the two principle directions. The angle  $\theta_{ze}$ , however, determined by  $y'_i$  (or  $\theta_{zh}$  by  $\alpha'_i$ ) corresponds to the direction parallel (or perpendicular) to the strike for a fault.

### (3) Three-dimensional case:

Arbitrary conductivity structures could produce the elements of impedance and admittance,  $z_{ij}$  and  $y_i$ , having arbitrary values, and the rotation loci may in general be ellipses as given in Fig. (2-5). However, if there exists a plane of symmetry through the measuring point, the behavior of  $z_{ij}$  and  $y_i$  with respect to the rotation angle  $\theta$  appears as those of a two-dimensional model. These special cases of three-dimensionality can be classified into the equivalent two-dimensional problems; some examples are shown in Fig. (2-8). The simplification in these cases results from the loss of information caused by making measurements at a point of symmetry. In fact the generalized two-dimensional impedance  $z_{ij}$  and admittance  $y_i$  can lead to unique interpretations only for two or one-dimensional models. However the more general result

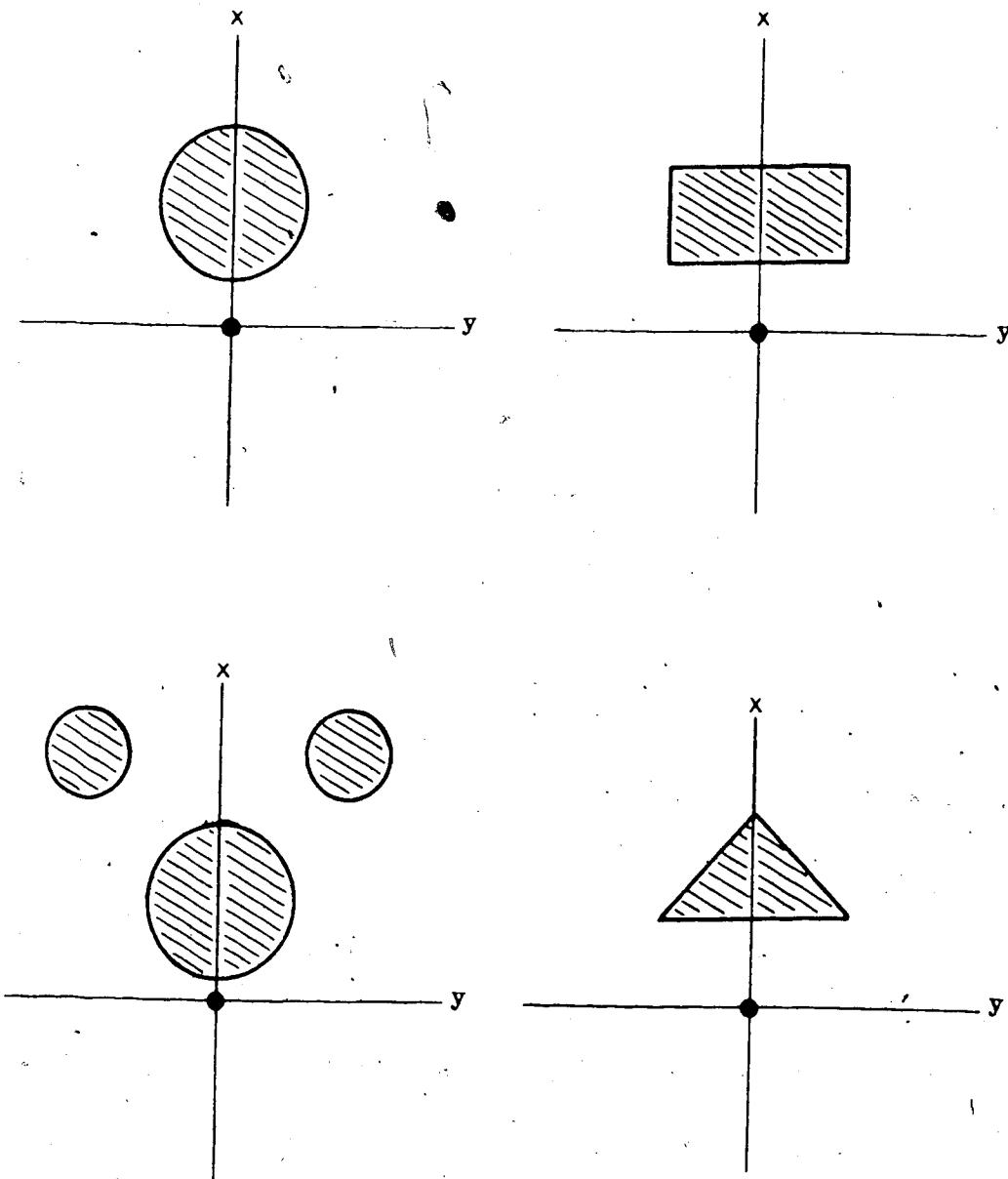


Fig. (2-8). Three-dimensional models with a plane of symmetry through the measuring point •.

can give information on the dimensionality of the structure in terms of the dimensionality indicators

$$\bar{\alpha}(\text{skew}) = \left| \begin{array}{c} z'_{xx} + z'_{yy} \\ z'_{xy} - z'_{yx} \end{array} \right| = \left| \begin{array}{c} z'_{xx} + z'_{yy} \\ z'_{xy} - z'_{yx} \end{array} \right| = \left| \begin{array}{c} z_2 \\ z_1 \end{array} \right| \quad (\text{Swift 1967})$$

(2-54a)

$$\beta_o = \frac{\left| \begin{array}{c} z'_{xx} - z'_{yy} \\ z'_{xy} + z'_{yx} \end{array} \right|}{\left| \begin{array}{c} z'_{xy} + z'_{yx} \end{array} \right|} \theta_o \quad (\text{Word et al. 1970})$$

(2-54b)

$$\beta_{oo} = \frac{\left| \begin{array}{c} z'_{xx} - z'_{yy} \\ z'_{xy} - z'_{yx} \end{array} \right|}{\left| \begin{array}{c} z'_{xy} - z'_{yx} \end{array} \right|} \theta_o \quad (\text{Word et al. 1970})$$

(2-54c)

and

$$\beta_{ze} = \frac{\left| \begin{array}{c} y' \\ y'_x \end{array} \right|}{\left| \begin{array}{c} y'_x \end{array} \right|} \theta_{ze} \quad \beta_{zh} = \frac{\left| \begin{array}{c} \alpha'_x \\ \alpha'_y \end{array} \right|}{\left| \begin{array}{c} \alpha'_y \end{array} \right|} \theta_{zh} \quad (\text{Word et al. 1970})$$

(2-54d)

Skew  $\bar{\alpha}$  is invariant with rotation angle  $\theta$ , and equal to the ratio between the two displacements from the origins to the centers of the corresponding ellipses as shown in Fig. (2-5). It is obvious that the condition  $\bar{\alpha} = 0$  can indicate either one or two-dimensionality. Word et al. (1970) propose that  $\bar{\alpha} = 0$  is a necessary but not a sufficient condition for two-dimensionality. The sufficient condition must include the condition  $\beta_o = 0$ .  $\beta_o$  is the ratio of minor to major axes of the  $z'_{ij}$  loci as shown in Fig. (2-5). For two-dimensional cases or equivalent, the rotation

loci are line ellipses, so that  $\beta_o$  is zero. However, for the one-dimensional cases,  $\beta_o = (0/0)$  becomes an undetermined quantity.

For one or two-dimensional cases,  $\beta_{oo} = 0$ .  $\beta_{oo}$  has the advantage as an indicator by avoiding  $\beta_o = 0/0$  for one-dimensional cases, in addition, it also has potential for judging the degree of three-dimensionality.

$\beta_{ze} = \beta_{zh} = 0$  for two-dimensionality, but is not necessarily non-zero for three-dimensionality.

The behavior of the dimensional indicators are summarized as following:

One dimensionality     $\bar{\alpha} = 0$

$$\beta_o = 0/0$$

$$\beta_{oo} = 0$$

$$\beta_{ze} = \beta_{zh} = 0/0$$

Two dimensionality     $\bar{\alpha} = 0$

and special               $\beta_o = 0$

case of symmetry        $\beta_{oo} = 0$

$$\beta_{ze} = \beta_{zh} = 0$$

Three dimensionality     $\bar{\alpha} \neq 0$

$$\beta_o \neq 0$$

$$\beta_{oo} \neq 0$$

$$\beta_{ze} = \beta_{zh} \stackrel{(may)}{=} 0$$

While the use of the dimensionality indicators provides some information regarding the nature of the structure, quantitative interpretation is restricted to one and two-dimensional models. In practice it is frequently possible to interpret the more usual three-dimensional structure in terms of these two and one-dimensional models to some reasonable degree of approximation. In this case, it may be desirable to define the following parameters:

$$\beta_1 = \frac{\rho'_{xx} + \rho'_{yy}}{\rho'_{xy} + \rho'_{yx}} \quad (2-55a)$$

$$\beta_2 = \frac{\rho'_M - \rho'_m}{\rho'_M} \quad (\text{anisotropy factor}) \quad (2-55b)$$

where  $\rho'_{ij}$  are the apparent resistivities corresponding to  $z'_{ij}(\theta_0)$ , and  $\rho'_M$  and  $\rho'_m$  correspond to the maximum and minimum values between  $\rho'_{xy}$  and  $\rho'_{yx}$ .

3-dimensionality  $\beta_1 \neq 0$ , measuring the degree of 3-dimensionality.

$$\beta_2 = 0 - 1$$

2-dimensionality  $\beta_1 = 0$  ( $\rho'_{xx} = \rho'_{yy} = 0$ )

$$\beta_2 \neq 0 \quad (\rho'_{xy} \neq \rho'_{yx})$$

1-dimensionality  $\beta_1 = 0$

$$\beta_2 = 0 \quad (\rho'_{xy} = \rho'_{yx})$$

## CHAPTER III

### INSTRUMENTATION AND DATA ANALYSIS

#### III-A. Instrumentation

The instruments used for data gathering have been described by Allsopp et al. (1974), and are shown in Fig. (3-1). It is designed to measure three components of the magnetic field ( $H_x, H_y, H_z$ ) and two components of the electric field ( $E_x, E_y$ ) of the micropulsation signals in the frequency range from 0.0001 to 10 Hz. A short description of this system will follow.

##### (1) Magnetic system:

A time varying magnetic field with amplitude of the order of  $10^{-2}$  to  $10^2$  gammas is detected with a specially designed induction type sensor (Burke and Rankin 1975) and produces  $\mu$ V level output in the frequency range of interest. Preamplification stage consists of a RATEK RA-11 parametric amplifier and two FET input operational amplifiers, with a total gain of 85 dB. The output is fed to a post amplifier with switchable voltage gains at -20, -10, 0, 10, 20 dB, and then through a Butterworth low pass filter to eliminate the aliasing effect. The overall amplitude and phase responses are shown in Fig. (3-2).

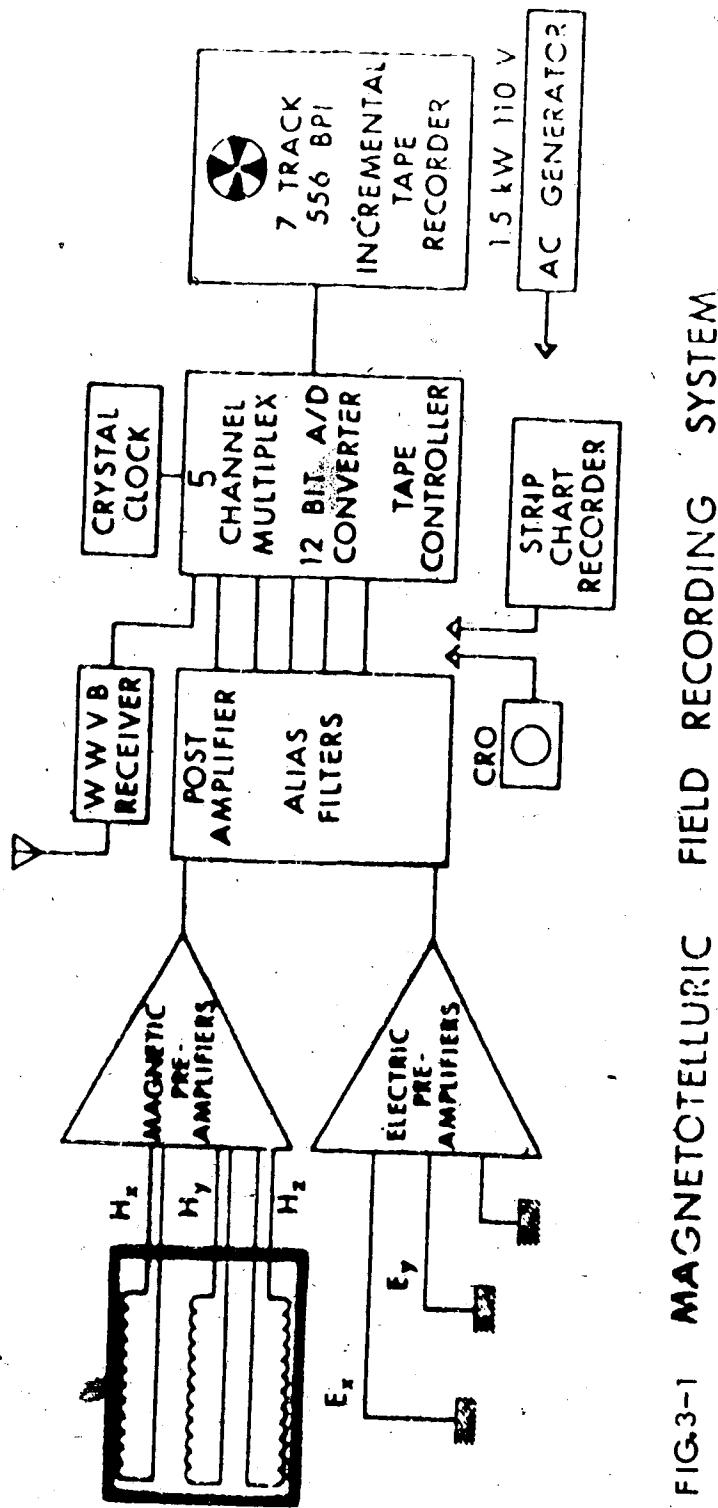


FIG. 3-1 MAGNETOTELLURIC FIELD RECORDING SYSTEM

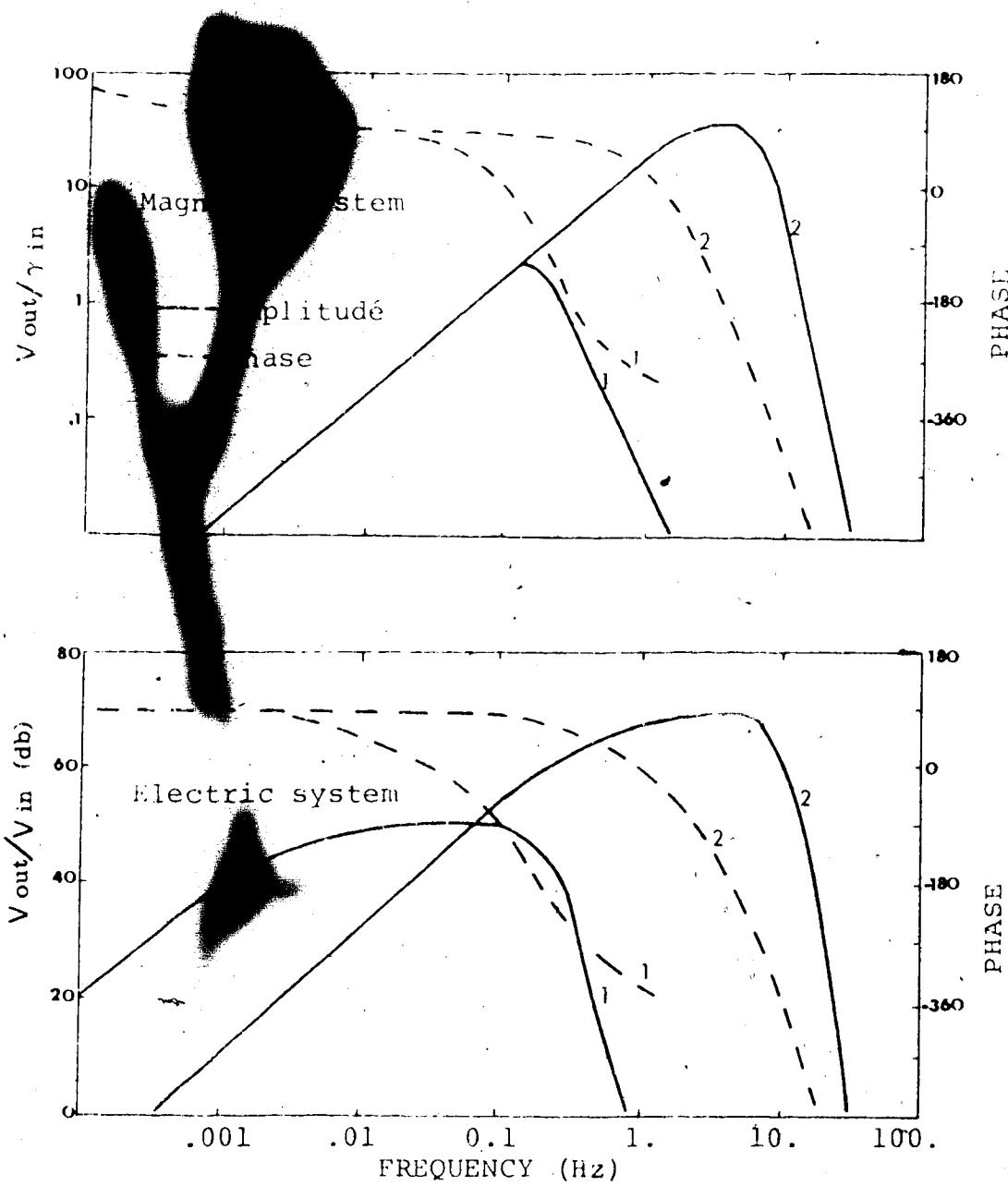


Fig.(3-2). Amplitude and phase responses of the magnetic and electric systems. (1) for low mode and (2) for high mode of recording.

(2) Electric system:

The electric field variations are detected as a voltage difference between copper electrodes in the ground. A chopper stabilized operational amplifier is used for preamplification. The post amplifier and aliasing filter are identical to that of the magnetic system as described above. The overall amplitude and phase responses are shown in Fig. (3-2).

(3) Digital recording:

The recording system consists of the multiplexer, A-D converter, crystal clock and a seven track tape recorder, with a dynamic range of  $\pm 10$  V. The sampling rates were chosen to be 1.25 and 40 samples/second/channel in low and high mode respectively. A 2700-feet tape with 556 bpi can last up to 20 days or 16 hours depending on the mode of operation.

### III-B. Data Analysis

The first step in the interpretation of the data is to obtain power spectral estimates of the measured quantities,  $H_x, H_y, H_z, E_x, E_y$ .

(1) Power spectrum computation:

Let  $f(t)$  represents the time sequence of a recorded signal. The procedure is as follows.

a) Truncation

Let  $N$  represent the number of samples per data set;  $T_o = N\Delta t$  is the sample time length (sec) per data set, where  $\Delta t$  is the sampling interval (sec).

A data set with length of  $N$  samples can be obtained by truncating the possible infinite time sequence  $f(t)$

$$f_o(t) = f(t) \cdot W_o(t) \quad (3-1)$$

where

$$W_o(t) = \begin{cases} 1, & (0 \leq t \leq T_o) \\ 0, & (t = \text{otherwise}) \end{cases}$$

The effect of this convolution is removed in the frequency domain.

b) Removing mean and trend

Since only the AC component of the signal is of interest, it is customary to eliminate the DC component at this stage by removing the mean. Removing the trend is done to remove the effect of contributions at periods longer than the fundamental and is accomplished by removing the slope corresponding to the difference between the average of the first and last third points.

c) Tapering

A large discontinuity may occur between the value of the function at the beginning and the end of the

data set; this discontinuity produces undesirable high frequency harmonics in the power spectrum. It is customary to taper the data set making the first and the last points approach the mean value which is now zero; and it is done by using a cosine bell window with one-tenth of the samples tapered at each end of the data set.

d) Fourier transform

The Fast Fourier Transform method is used to transform the time sequence  $f_o(t)$ , from the time domain into the frequency domain.

$$F_o(\omega_n) = \frac{2}{T_o} \int_0^{T_o} f_o(t) e^{-j\omega_n t} dt = F(\omega_n) * W_o(\omega_n) \quad (3-2)$$

where  $F(\omega)$  and  $W_o(\omega)$  are the Fourier transfer function of  $f(t)$  and  $W_o(t)$  respectively, and  $\omega = 2\pi f$ . From N data points,  $N/2$  complex coefficients  $F_o(\omega_n)$  are obtained with

$$\omega_n = n\omega_1, \text{ where } \omega_1 = 2\pi/T_o, \text{ and } n = 1, 2, \dots, N/2.$$

The corresponding frequency range is

$$(f_1 = 1/T_o) \leq f_n \leq (f_N = 1/2\Delta t).$$

The highest frequency  $f_N$  is called the Nyquist or folding frequency. Components of higher frequencies than  $f_N$  are analog filtered out of the recorded signals, to avoid aliasing effects.

e) Hamming window

The Fourier coefficients  $F_O(\omega)$  obtained from a truncated data set are different from the actual ones  $F(\omega)$  due to the truncating effect. The correction can be done by multiplication with  $W_O(\omega)$  in the frequency domain. However, the Hamming or a similar window can be used as an excellent approximation for this correction. The Fourier transfer of the Hamming window has the form

$$W_h(\omega_n) = 0.54 W_O(\omega_n) + 0.23 [W_O(\omega_{n-1}) + W_O(\omega_{n+1})], \quad (3-3)$$

Applying this window by convolution in the frequency domain, the Fourier coefficients of the data signal are obtained as

$$F(\omega_n) = 0.54 F_O(\omega_n) + 0.23 [F_O(\omega_{n-1}) + F_O(\omega_{n+1})]. \quad (3-4)$$

f) System response correction

The overall system response must be removed to obtain the true Fourier coefficients of the E and H-field quantities.

g) Power spectrum computation

Auto and cross-power spectra are computed as

$$P_{IJ}(\omega_n) = F_I^*(\omega_n) \cdot F_J(\omega_n) \quad (3-5)$$

where \* indicates the complex conjugate and I,J

(=1,2,3,4,5) are integer indexes referring to the data channels. The number of samples per data set used is usually large (say, 4096), so that there are a large number ( $N/2$ ) of harmonics in the power spectrum. At this point, smoothing is carried out to obtain average power spectral estimates. This reduces the computing time in later stages of the analysis, and eliminates noise effect (see Chapter IV). Ten frequencies are obtained in each frequency decade; the ratio of center frequencies of successive bands is

$$\log \frac{f_{k+1}}{f_k} = \frac{1}{10}$$

Thus, the bandwidth averaging power spectral estimates at a representative frequency  $\omega_k$  is

$$\langle P_{IJ}(\omega_k) \rangle = \frac{1}{M_b} \sum_{\text{(band)}} P_{IJ}(\omega_n) \quad (3-6)$$

where  $M_b$  is the number of harmonics involved in the bandwidth averaging.

If more than one data set is used, then the overall averaging spectral estimate is computed as

$$\langle P_{IJ}(\omega_k) \rangle = \frac{1}{M_s} \sum_{m=1}^{M_s} P_{IJ}^m(\omega_k) \quad (3-7)$$

where  $m = 1, 2, \dots, M_s$  represents a given data set and  $M_s$  is the number of data sets contributing to this average.

(2) Polarization:

The polarization characteristics of the MT-field have been studied, using power spectral estimates, by Fowler et al. (1967), Paulson (1968), and Rankin and Kurtz (1970). The three important polarization parameters are: Degree of Polarization, Polarization Angle, and Ellipticity.

a) Degree of Polarization (DP)

For quasi-monochromatic EM-signals, the power spectral estimates may consist of polarized and unpolarized portions, and can be represented by a matrix of the following form

$$\begin{pmatrix} \langle P_{xx} \rangle & \langle P_{xy} \rangle \\ \langle P_{yx} \rangle & \langle P_{yy} \rangle \end{pmatrix} = \begin{pmatrix} A & B \\ B^* & C \end{pmatrix} + \begin{pmatrix} D & 0 \\ 0 & D \end{pmatrix} \quad (3-8)$$

where x and y refer to the components of either E or H. The first matrix on the right hand side corresponds to the completely polarized contribution, i.e. that part with unit coherency ( $AC = B^*B$ ); while the second matrix corresponds to the incoherent contribution. The degree of polarization is defined as the ratio of the polarized intensity to the total intensity of the signals.

$$DP = \frac{A + C}{\langle P_{xx} \rangle + \langle P_{yy} \rangle} = \left( 1 - \frac{4|J|}{(\langle P_{xx} \rangle + \langle P_{yy} \rangle)^2} \right)^{\frac{1}{2}} \quad (3-9)$$

where  $J = \langle P_{xx} \rangle \langle P_{yy} \rangle - \langle P_{xy} \rangle \langle P_{yx} \rangle$ , is the determinant of the generalized power matrix. For a completely polarized signal, the coherency between x and y components is unity so that  $J = 0$  and  $DP = 1$ . For a completely unpolarized signal, where the coherency is zero,  $DP = 0$ .

b) Polarization Angle ( $\theta_p$ )

The polarization angle  $\theta_p$  is defined as the major axis azimuth angle of the polarization ellipse of the polarized field

$$\tan 2\theta_p = \frac{2 \operatorname{Re} \langle P_{xy} \rangle}{\langle P_{xx} \rangle - \langle P_{yy} \rangle} \quad (3-10)$$

c) Ellipticity ( $\epsilon$ )

Ellipticity is defined as the ratio of minor to major axis of the polarization ellipse

$$\epsilon = \tan \left( \frac{1}{2} \sin^{-1} \frac{2 \operatorname{Im} \langle P_{xy} \rangle}{[(\langle P_{xx} \rangle - \langle P_{yy} \rangle)^2 + 4 \langle P_{xy} \rangle \langle P_{yx} \rangle]^{\frac{1}{2}}} \right) \quad (3-11)$$

$\epsilon$  values ranges from 0 for a linearly polarized field, to 1 for a circularly polarized field. A positive or negative value of  $\epsilon$  represents the sense of polarization of righthand or lefthand as measured when looking into the propagating wave.

(3) Impedance computation:

In order to determine the surface impedance matrix elements  $z_{ij}$ , using Eqns. (2-37), two independent data determinations required. However, determination can be done by using power spectral estimates in the frequency domain. Consider the frequency domain expression defined the surface impedance tensor, as

$$E_x = Z_{xx} H_x + Z_{xy} H_y \quad (3-12)$$

[ (2-37) ]

$$E_y = Z_{yx} H_x + Z_{yy} H_y$$

Let each of Eqns. (3-12) be multiplied in turn by the complex conjugate of each of the 5 field quantities ( $H_x, H_y, H_z, E_x, E_y$ ). If the source field is randomly or partly randomly polarized with respect to the frequency, there is some degree of statistical independence between the cross-powers of the various field components.

Consequently, the power spectral estimates  $\langle P_{IJ}(\omega) \rangle$  as determined in Eqn. (3-6) will produce 5 linearly independent equations for each of Eqns. (3-12) of the form

$$1. \quad \langle E_x^* E_i \rangle = z_{ix} \langle E_x^* H_x \rangle + z_{iy} \langle E_x^* H_y \rangle$$

$$2. \quad \langle E_y^* E_i \rangle = z_{ix} \langle E_y^* H_x \rangle + z_{iy} \langle E_y^* H_y \rangle$$

$$3. \quad \langle H_x^* E_i \rangle = z_{ix} \langle H_x^* H_x \rangle + z_{iy} \langle H_x^* H_y \rangle$$

$$4. \quad \langle H_y^* E_i \rangle = Z_{ix} \langle H_y^* H_x \rangle + Z_{iy} \langle H_y^* H_y \rangle$$

$$5. \quad \langle H_z^* E_i \rangle = Z_{ix} \langle H_z^* H_x \rangle + Z_{iy} \langle H_z^* H_y \rangle \quad (3-13)$$

where  $i = x, y$ .

Using Equations 3 and 4 of (3-13) is equivalent to the least square error fitting method (Sims et al. 1971).

Eqns. (3-13) provide 10 pairs of simultaneous equations for solution of each impedance element. However, the pairs (1,4), (2,3), and those involving (5) are not suitable since under some circumstances the solutions for impedance functions become indeterminate. There are 4 remaining pairs (1,2), (1,3), (2,4), (3,4) which are well behaved as long as the source field is at least partially unpolarized. Thus, up to 4 independently computed values of each of the impedance elements  $Z_{ij}$  can be obtained at each frequency  $\omega_k$ . It is useful to average

$$Z_{ij}(\omega_k) = \frac{1}{L} \sum_{\ell=1}^L z_{ij}^\ell(\omega_k) \quad (3-14)$$

where  $\ell = 1, 2, 3, 4$  represents a given solution and L is the number of acceptable solutions among the possible 4.

In order for a solution to exist by the method described, it is necessary that the smoothing be carried out i.e. at least over 2 frequency estimates, or to use 2 independent data sets. On the other hand,

in order to obtain satisfactory value of  $Z_{ij}^{\circ}$ , the bandwidth used for smoothing must be limited to a range for which  $Z_{ij}$  does not vary significantly with respect to frequency.

The admittance elements defined by

$$H_z = Y_x E_x + Y_y E_y \quad (3-15)$$

[ (2-38) ]

and the coefficients  $\alpha_x$  and  $\alpha_y$  in the relation

$$H_z = \alpha_x E_x + \alpha_y E_y \quad (3-16)$$

[ (2-35) ]

are computed in a similar manner and are subject to the same conditions as the impedance elements  $Z_{ij}$  just discussed, and the result is

$$Y_i = \frac{1}{L} \sum_{\ell=1}^L Y_i^{\ell}(\omega_k) \quad (3-17)$$

$$\alpha_i = \frac{1}{L} \sum_{\ell=1}^L \alpha_i^{\ell}(\omega_k) \quad (3-18)$$

The apparent resistivity for an impedance element  $Z_{ij}$  is computed as

$$\rho_{ij}(\omega_k) = \frac{0.2}{f_k} |Z_{ij}(\omega_k)|^2 \quad (\text{in ohm-meter}) \quad (3-19)$$

where the unit of  $Z_{ij}$  is (millivolt/kilometer)/gamma.

(4) Criteria of data reliability:

Experimental data is inevitably recorded in the presence of noise of one kind or another, and the establishment of reliable criteria for judging the quality of a computed result is important. For the case at hand, noise is defined as any portion of the measured EH-field quantities not conforming to the impedance and admittance relationships defined in Eqns. (3-12), (3-15), and (3-16). Both phasor coherency and predicted coherency, have been proposed and used as a measure of data quality.

a) Phasor Coherency (CP) (Word et al. 1970)

When noise is present in the measured EH-fields, the independently computed solutions of  $z_{ij}^l$ ,  $y_i^l$ , and  $\alpha_i^l$  are not completely consistent. The degree of scatter between the independent solutions can be used as an effective measure of the reliability of the results.

The phasor coherency, representing a measure of the degree of data quality, is defined as

$$CP = 1 - \frac{D}{|R|}$$

$$\text{where } R = \frac{1}{L} \sum_{\ell=1}^L R_\ell \quad \text{and} \quad D = \frac{1}{L} \sum_{\ell=1}^L |R_\ell - R|$$

and  $R_\ell$  represents the  $\ell$ -th solution for  $z_{ij}^l$ ,  $y_i^l$ , or  $\alpha_i^l$ , and  $L$  the number of acceptable solutions. The phasor

coherency has the range of

$$-\infty \leq CP \leq 1$$

b) Predicted Coherency (PCH) (Swift 1967)

The predicted coherency is defined as the coherency between the measured field component and the corresponding component predicted from the average value of  $z_{ij}^l$ ,  $y_i^l$ , or  $\alpha_i^l$ . The PCH for the impedance is given as

$$PCH(E^*E^p_x) = \frac{|z_{xx}^{*}e_x^*h_x^* + z_{xy}^{*}e_x^*h_y^*|}{\{<P_{Ex}>[|z_{xx}|^2<P_{Hx}> + |z_{xy}|^2<P_{Hy}> + 2\operatorname{Re}(z_{xx}^*z_{xy}^{*}e_x^*h_y^*)]\}^{\frac{1}{2}}$$

$$PCH(E^*E^p_y) = \frac{|z_{yx}^{*}e_y^*h_x^* + z_{yy}^{*}e_y^*h_y^*|}{\{<P_{Ey}>[|z_{yx}|^2<P_{Hx}> + |z_{yy}|^2<P_{Hy}> + 2\operatorname{Re}(z_{yx}^*z_{yy}^{*}e_y^*h_y^*)]\}^{\frac{1}{2}}$$

and for the admittance

$$PCH(H^*H^p_z) = \frac{|y_x^{*}h_z^*e_x^* + y_y^{*}h_z^*e_y^*|}{\{<P_{Hz}>[|y_x|^2<P_{Ex}> + |y_y|^2<P_{Ey}> + 2\operatorname{Re}(y_x^*y_y^{*}e_x^*e_y^*)]\}^{\frac{1}{2}}$$

where the superscript p represents the predicted quantity. The PCH has the range of

$$0 \leq PCH \leq 1$$

The phasor and predicted coherency represent a measure of the degree to which all influences in the measured EH-fields are related by the same tensor.  $CP = 1$  and  $PCH = 1$  if all solutions of a given  $R^{\ell}$  coincides in both magnitude and phase. But the converse is not necessary, since noise can exist and be coherent between channels. As part of the process of analysis, the ordinary coherencies ( $E_{x,y}^{H_x}$ ) and ( $E_{y,x}^{H_x}$ ) are computed for the raw data. If these ordinary coherencies are low the data is rejected from the subsequent processing stream.

The problem of noise contamination will be discussed in the following chapter.

## CHAPTER IV

### NOISE DISCUSSION

In the statistical sense, the primary magnetic signals appearing in the magnetotelluric method can be considered as noise. However in arbitrary measuring directions the orthogonal components of the magnetic fields, consisting of primary plus secondary contribution, show a considerable degree of coherence. The components of the electric field at the surface of the earth are substantially secondary and are thus highly coherent with both the orthogonal magnetic component and between themselves. In the subsequent discussion those components of the fields which are coherent in the sense just described will be considered to be signal whereas all other contributions, whether coherent or not, will be considered as noise.

As an example of coherent noise, the 60 Hz power line contribution is frequently present and strenuous efforts must be made to remove it. Other examples are power line spikes and steps, which can affect all channels of information. This noise is usually eliminated by discarding sections of record during the pre-editing processes.

Random noise cannot be eliminated by any process except in a statistical sense, but since it is incoherent

its effect can be taken into account and noisy data can be rejected according to a predetermined criterion for coherency.

#### IV-A. General Discussion

##### (1) Covariance and Coherence

The covariance functions for time sequences can be defined by

$$a_I(\tau) = \frac{1}{N} \sum_{t=1}^{N-\tau} X_I(t) X_I(t+\tau) \quad \text{auto-covariance} \quad (4-1a)$$

$$c_{IJ}(\tau) = \frac{1}{N} \sum_{t=1}^{N-\tau} X_I(t) X_J(t+\tau) \quad \text{cross-covariance} \quad (4-1b)$$

where  $X_I$  and  $X_J$  are time sequences with zero mean values, and  $\tau = 0, 1, 2, \dots$  represents the time lag.

For signals which are quasi-periodic and coherent,  $a(\tau)$  and  $c(\tau)$  are also quasi-periodic functions of  $\tau$ . For random noise which is non-coherent,  $a(\tau)$  and  $c(\tau)$  are also random, and when  $N$  is large  $a(\tau, \tau \neq 0)$  and  $c(\tau)$  becomes very small and can be considered negligible when compared with  $a(0)$ , where  $a(0)$  is the zero lag auto-covariance which is also the energy density of the time sequence.

$$a(0) = \frac{1}{N} \sum_{t=1}^N |X(t)|^2 = \sigma^2 \quad (4-2)$$

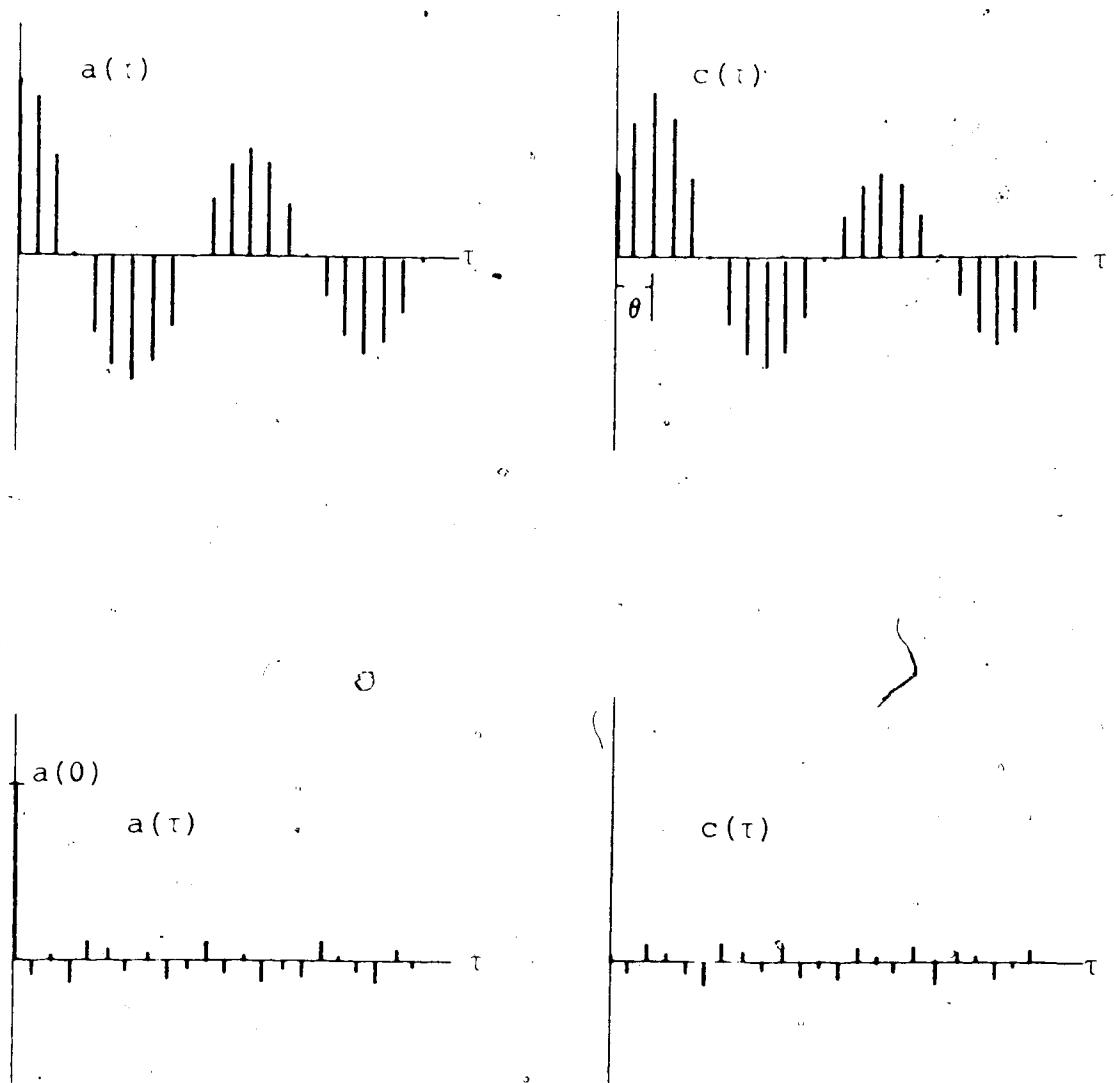


Fig. (4-1). Relative amplitudes of auto-covariance  $a(\tau)$  and cross-covariance  $c(\tau)$  for two coherent sinusoidal sequences with a phase  $\theta$  (upper), and for two random noise sequences (lower).

This also turns out to be the variance estimate of  $X(t)$  which, if  $X(t)$  is stationary, approaches a constant value when  $N$  is large.

Fig. (4-1) shows the relative amplitudes of  $a(\tau)$  and  $c(\tau)$  for two coherent sinusoidal sequences with a phase difference  $\theta$ , and for two random noise sequences.

Now consider time sequences which consist of the sum of signal  $s(t)$  and random noise  $n(t)$

$$x_I(t) = s_I(t) + n_I(t) \quad (4-3a)$$

$$x_J(t) = s_J(t) + n_J(t) \quad (4-3b)$$

The auto- and cross-covariance functions are

$$\begin{aligned} a_I(\tau) &= \frac{1}{N} \sum_{t=1}^{N-\tau} x_I(t) x_I(t+\tau) \\ &= \frac{1}{N} \left[ \sum_{t=1}^{N-\tau} s_I(t) s_I(t+\tau) + \sum_{t=1}^{N-\tau} n_I(t) n_I(t+\tau) \right. \\ &\quad \left. + \sum_{t=1}^{N-\tau} s_I(t) n_I(t+\tau) + \sum_{t=1}^{N-\tau} n_I(t) s_I(t+\tau) \right] \quad (4-4a) \end{aligned}$$

$$\begin{aligned} c_{IJ}(\tau) &= \frac{1}{N} \sum_{t=1}^{N-\tau} x_I(t) x_J(t+\tau) \\ &= \frac{1}{N} \left[ \sum_{t=1}^{N-\tau} s_I(t) s_J(t+\tau) + \sum_{t=1}^{N-\tau} n_I(t) n_J(t+\tau) \right. \\ &\quad \left. + \sum_{t=1}^{N-\tau} s_I(t) n_J(t+\tau) + \sum_{t=1}^{N-\tau} n_I(t) s_J(t+\tau) \right] \quad (4-4b) \end{aligned}$$

The first term in both  $a(t)$  and  $c(t)$  is signal and the second term is noise; the last two terms are the noise-signal interference terms and pertain to the noise effect. Except for the second term of the zero lag auto-covariance  $a_n(0)$ , which is also the noise energy density, each of the remaining terms containing  $n$  is a sum of random functions as indicated in Eqns. (4-4), and thus decreases as  $N$  increases. For sufficiently long time sequences, it will be shown later that terms in Eqns. (4-4) involving random functions can be neglected.

## (2) Power Spectrum of Signal and Noise

### a) Power spectrum of random noise

For a random time sequence  $n(t)$ , its complex Fourier coefficients are defined by

$$n(\omega_n) = \frac{2}{T} \int_0^T n(t) e^{-j\omega_n t} dt \quad (4-5a)$$

which for digitized data becomes

$$n(\omega_n) = \frac{2}{N} \sum_{t=1}^N n(t) e^{-j2\pi nt/N} \quad (4-5b)$$

The corresponding power spectrum is defined by

$$P_n(\omega_n) = n^*(\omega_n) n(\omega_n) \quad (4-6)$$

$n(\omega)$  and  $P_n(\omega)$  are also random in both amplitude and

phase. For digitized data one obtains for the average noise power

$$\begin{aligned}\bar{P}_n - \bar{\varepsilon} &= \frac{2}{N} \sum_{n=1}^{N/2} P_n(\omega_n) \\ &= \frac{2}{N} \cdot \frac{2}{N} \sum_{t=1}^N |\eta(t)|^2 = \frac{4}{N} \sigma_N^2\end{aligned}\quad (4-7)$$

where

$$\sigma_N^2 = \frac{1}{N} \sum_{t=1}^N |\eta(t)|^2 (=a(0))$$

is the variance estimate which approaches a constant value as  $N$  becomes sufficiently large. Henceforth we will consider discretely digitized data of finite sample length  $N$ .

As mentioned earlier the distribution of the amplitudes of a random noise power spectrum is also random. According to an experimental study of a typically random process described in section A-4 of this chapter, the maximum value in the spectrum

$P_n(\omega_n) \sim 4\bar{\varepsilon}$ . It can be considered that the power spectrum  $P_n(\omega_n)$  can be represented by the average value  $\bar{\varepsilon}$  weighted by a factor  $K_n^2$  so that

$$P_n(\omega_n) = K_n^2 \bar{\varepsilon} = K_n^2 \frac{4}{N} \sigma_N^2 \quad (4-8)$$

The  $K_n^2$  form a set of random positive numbers with

$$\frac{2}{N} \sum_{n=1}^{N/2} K_n^2 = 1$$

Now let the weighting factor for the maximum value of  $P_{\eta}(\omega)$  be represented by  $K_M^2$ . The magnitude of  $K_M^2$  is found to be limited; for instance,  $K_M^2 \sim 4$  for the case given above. Also, as mentioned earlier,  $\sigma^2$  has a constant value, and thus it is clear that for a stationary random process the amplitude of the power spectrum as given in Eqn. (4-8) is inversely proportional to  $N$ , the number of sample points.

b) Power spectrum of signal

For a monofrequency sinusoidal signal  $S(t) = s_0 \sin(\omega_0 t)$ , the Fourier coefficient is

$$S(\omega_0) = \frac{2}{T} \int_0^T s_0 \sin(\omega_0 t) e^{-j\omega_0 t} dt \quad (4-9)$$

and the magnitude of the power spectrum at this frequency is

$$P_S(\omega_0) = S^*(\omega_0) S(\omega_0) = s_0^2 \quad (4-10)$$

which is independent of the number of sample points  $N$ .

The same condition holds for each harmonic of a more general time sequence.

It is seen from Eqn. (4-8) and Eqn. (4-10) that an obvious distinction exists between signal and random noise. For a signal the amplitude of the spectrum is independent of  $N$ , while for random noise, it is inversely proportional to  $N$ . This fact can also be

understood by considering that for a continuously periodic signal the power is distributed to only those frequencies that are present in the signal, while for random noise, the power is distributed over all  $N/2$  harmonics with equal probability.

c) Spectral densities with mixed signal and noise

For a mixed signal and noise time sequence as given in Eqn. (4-3)

$$x(t) = s(t) + n(t) \quad (4-11a)$$

the Fourier coefficients are

$$X(\omega_n) = S(\omega_n) + N(\omega_n) \quad (4-11b)$$

where

$$S(\omega_n) = \frac{2}{T} \int_0^T s(t) e^{-j\omega_n t} dt$$

and

$$N(\omega_n) = \frac{2}{T} \int_0^T n(t) e^{-j\omega_n t} dt$$

The auto-power spectrum is

$$P(\omega_n) = X^*(\omega_n) X(\omega_n) = P_s(\omega_n) + P_n(\omega_n) \quad (4-12)$$

where the signal is  $P_s(\omega_n) = S^*(\omega_n) S(\omega_n)$ , and the noise contribution is  $P_n(\omega_n) = N^*(\omega_n) N(\omega_n) + S^*(\omega_n) N(\omega_n) + N^*(\omega_n) S(\omega_n)$ . Using the result deduced in the last section,

$$P_S(\omega_n) = s_o^2 \quad (4-13a)$$

$$P_\eta(\omega_n) = \frac{1}{N} 4\sigma^2 K_n^2 + \frac{1}{\sqrt{N}} 4s_o K_n \cos(\gamma_n) \quad (4-13b)$$

where  $\gamma_n$  is the phase angle of  $s^*(\omega_n)\eta(\omega_n)$  and is random, ranging from 0 to  $2\pi$ .

The noise-signal ratio of auto-power  $(\eta/S)_a$  for a given harmonic is defined as

$$(\eta/S)_a = \frac{P_\eta(\omega_n)}{P_S(\omega_n)} = \frac{1}{N} \frac{4\sigma^2}{s_o^2} K_n^2 + \frac{1}{\sqrt{N}} \frac{4\sigma}{s_o} K_n \cos(\gamma_n) \quad (4-14)$$

and since  $K_n \leq K_M$  and  $\cos(\gamma_n)$  can be either positive or negative

$$\left( \frac{4\sigma^2}{Ns_o^2} K_M^2 - \frac{4\sigma}{\sqrt{N} s_o} K_M \right) \leq (\eta/S)_a \leq \left( \frac{4\sigma^2}{Ns_o^2} K_M^2 + \frac{4\sigma}{\sqrt{N} s_o} K_M \right) \quad (4-15)$$

The general cross-power spectrum between  $X_I(t)$  and  $X_J(t)$  for a given frequency  $\omega$  is

$$P_{IJ}(\omega) = X_I^*(\omega)X_J(\omega) = P_{SIJ}(\omega) + P_{\eta IJ}(\omega) \quad (4-16a)$$

where the signal is

$$P_{SIJ}(\omega) = S_I^*(\omega)S_J(\omega) \quad (4-16b)$$

and the noise contribution is

$$P_{\eta IJ}(\omega) = \eta_I^*(\omega)\eta_J(\omega) + S_I^*(\omega)\eta_J(\omega) + \eta_I^*(\omega)S_J(\omega) \quad (4-16c)$$

which can be written as

$$P_{SIJ}(\omega) = s_o^2 e^{j\theta_s} s$$

$$P_{nIJ}(\omega) = \frac{1}{N} 4\sigma^2 K_{nI} K_{nJ} e^{j\theta_n} + \frac{2s_o\sigma}{\sqrt{N}} (K_{nI} e^{j\theta_I} + K_{nJ} e^{j\theta_J})$$

where  $\theta$ 's are the phase angles for the corresponding terms in Eqns. (4-16b,c).

The noise-signal ratio of cross-power for a given harmonic is defined as

$$(n/S)_c = \left| \frac{P_{nIJ}(\omega)}{P_{SIJ}(\omega)} \right| = \left| \frac{\frac{1}{N} 4\sigma^2 K_{nI} K_{nJ} e^{j\theta_n} + \frac{1}{\sqrt{N}} 2s_o (K_{nI} e^{j\theta_I} + K_{nJ} e^{j\theta_J})}{s_o^2 e^{j\theta_s} s} \right| \quad (4-17a)$$

$$\leq \left( \frac{4\sigma^2}{Ns_o^2} K_M^2 + \frac{4\sigma}{\sqrt{N} s_o} K_M \right) \quad (4-17b)$$

since  $K_{nI}^2, K_{nJ}^2 \leq K_M^2$ .

It is seen from Eqns. (4-14) and (4-17) that the noise-signal ratio in the power spectrum depends upon  $N$ , the number of sample points used for analysis. In some cases, the power estimate resulting from an averaging is desired; these will be discussed in the next section.

### (3) Power Estimate and Averaging Effect

In some cases such as MT prospecting, the power spectrum which is used for further analysis consists of smoothed estimates, averaged over a frequency band (Mb), or averaged at a given frequency over a number of data sets (Ms), or both. The resulting auto-power is estimated by

$$\langle P \rangle = \langle P_S \rangle + \langle P_\eta \rangle \quad (4-18a)$$

From equations (4-13)

$$\begin{aligned} \langle P_S \rangle &= \frac{1}{Ms} \sum_{k=1}^{Ms} \frac{1}{Mb} \sum_{n=1}^{Mb} P_{Sk}(\omega_n) \\ &= \frac{1}{M} \sum_{n=1}^M P_S(\omega_n) = \overline{s^2} \end{aligned} \quad (4-18b)$$

where  $P_{Sk}(\omega)$  is the signal power density of the k-th set,  $M = Ms \cdot Mb$ , and  $\sum_{n=1}^M$  stands for the double summation.

$$\begin{aligned} \langle P_\eta \rangle &= \frac{1}{M} \sum_{n=1}^M P_\eta(\omega_n) \\ &= \frac{4c^2}{N} \frac{1}{M} \sum_{n=1}^M K_n^2 + \frac{4c\bar{s}}{\sqrt{N}} \frac{1}{M} \sum_{n=1}^M (K_n \cos(\gamma_n)) \end{aligned} \quad (4-18c)$$

$\frac{1}{M} \sum_{n=1}^M K_n^2 \rightarrow 1$  as previously shown. Define  $\sum_{n=1}^M (K_n \cos(\gamma_n)) = \langle K_n \rangle$ , Eqn. (4-18c) becomes

$$\langle P \rangle = \frac{4c^2}{N} + \frac{4c\bar{s}}{\sqrt{N}} \frac{\langle K_n \rangle}{M} \quad (4-19)$$

The noise-signal ratio of the auto-power estimate is estimated from Eqn. (4-18b) and (4-19),

$$\langle N/S \rangle_a = \frac{\langle P_n \rangle}{\langle P_S \rangle} = \frac{4\sigma^2}{N s^2} + \frac{4\sigma}{\sqrt{N s M}} \langle K_n \rangle \quad (4-20a)$$

or, by using the reasonable estimate of  $\langle K_n \rangle$

$$\left( \frac{4\sigma^2}{N s^2} + \frac{4\sigma K_M}{\sqrt{N s M}} \right) \lesssim \langle N/S \rangle_a \lesssim \left( \frac{4\sigma^2}{N s^2} + \frac{4\sigma K_M}{\sqrt{N s M}} \right) \quad (4-20b)$$

The averaged cross-power estimate is

$$\langle P_{IJ} \rangle = \langle P_{SIJ} \rangle + \langle P_{nIJ} \rangle \quad (4-21a)$$

where

$$\langle P_{SIJ} \rangle = \frac{1}{M} \sum_{n=1}^M P_{SIJ}(\omega_n) = \frac{1}{M} \sum_{n=1}^M (s_n^2 e^{j\theta_{nn}}) \quad (4-21b)$$

$$\begin{aligned} \langle P_{nIJ} \rangle &= \frac{4\sigma^2}{N} \frac{1}{M} \sum_{n=1}^M (K_{nI} K_{nJ} e^{j\theta_{nn}}) \\ &+ \frac{2\sigma s}{\sqrt{N}} \frac{1}{M} \left( \sum_{n=1}^M K_{nI} e^{j\theta_{In}} + \sum_{n=1}^M K_{nJ} e^{j\theta_{Jn}} \right) \end{aligned} \quad (4-21c)$$

Since the signals are coherent, and assuming that  $e^{j\theta_{nn}}$  are in phase within a given band, Eqn. (4-21b) can be rewritten as

$$\langle P_{SIJ} \rangle = s^2 e^{j\theta_S} \quad (4-22a)$$

However, the phases of noise terms are random, so that

Eqn. (4-21c) can be written

$$\langle P_{nIJ} \rangle = \frac{4\sigma^2}{N} \frac{\langle K_n \rangle^2 e^{j\langle \theta_n \rangle}}{M} + \frac{4\sigma s}{\sqrt{N}} \frac{\langle K_n \rangle e^{j\langle \theta_n \rangle}}{M} \quad (4-22b)$$

where  $\langle \theta_n \rangle$ 's represent the average of the random phases, ranging from 0 to  $2\pi$ . From Eqns. (4-22a) and (4-22b), the noise-signal ratio of the cross-power is

$$\langle \eta/S \rangle_c = \left| \frac{\langle P_{nIJ} \rangle}{\langle P_{SIJ} \rangle} \right| = \left| \frac{\frac{4\sigma^2}{N} \frac{\langle K_n \rangle^2 e^{j\langle \theta_n \rangle}}{M}}{\frac{4\sigma^2}{N s^2} \frac{\langle K_M \rangle}{M}} + \frac{4\sigma}{\sqrt{N s}} \frac{\langle K_n \rangle e^{j\langle \theta_n \rangle}}{M} \right| \quad (4-23a)$$

$$\approx \left( \frac{4\sigma^2 K_M^2}{N s^2 M} + \frac{4\sigma K_M}{\sqrt{N s} M} \right) \quad (4-23b)$$

Eqns (4-20b) and (4-23b) represent the limits of the noise effect in the auto- and cross-power estimates. However, both terms of  $\langle P_{nIJ} \rangle$  as given in Eqn. (4-22b) and the second term of  $\langle P_n \rangle$  as given in Eqn. (4-19) are average values of random functions, and thus become negligible for sufficiently large  $M$ . In other words, the noise contribution can be largely removed from the cross-power estimate and partially removed from the auto-power estimate. In addition, there are generally some mechanisms which can further diminish the noise contribution from auto-power estimates through the use of a good cross-power estimate (White 1973).

#### (4) Experiment

For the purpose of investigating the noise effect quantitatively, three noise-signal mixed sequences were generated.

$$X_I(t) = S_I(t) + N_I(t), \quad I = 1, 2, 3. \quad (4-24)$$

where

$$S_I(t) = \sum_{n=1}^N s_n \sin(\omega_n t + \phi_n), \quad \text{with } s_n^2 = 10.$$

The three independent noise components  $N_I$  were taken from the random noise sequence provided by the Computing Services Library of the University of Alberta. The noise sequences are normalized to  $s_I^2 = 5$ , and  $K_M^2 = 4$  for these sequences. The formulas used for computing the noise-signal ratio are:

For individual harmonics

$$(\cdot/S)_I = \frac{P_{XI}(\omega) - P_{SI}(\omega)}{P_{SI}(\omega)} \quad (\text{auto-power}) \quad (4-25a)$$

$$(\cdot/S)_{IJ} = \left| \frac{P_{XIJ}(\omega) - P_{SIJ}(\omega)}{P_{SIJ}(\omega)} \right| \quad (\text{cross-power}) \quad (4-25b)$$

For power estimates

$$\langle \cdot/S \rangle_I = \frac{\langle P_{XI} \rangle - \langle P_{SI} \rangle}{\langle P_{SI} \rangle} \quad (\text{auto-power}) \quad (4-26a)$$

$$\langle \cdot/S \rangle_{IJ} = \left| \frac{\langle P_{XIJ} \rangle - \langle P_{SIJ} \rangle}{\langle P_{SIJ} \rangle} \right| \quad (\text{cross-power}) \quad (4-26b)$$

where

$$P_{XI}(\omega) = X_I^*(\omega)X_I(\omega)$$

$$P_{SI}(\omega) = S_I^*(\omega)S_I(\omega)$$

$$P_{XIJ}(\omega) = X_I^*(\omega)X_J(\omega)$$

$$P_{SIJ}(\omega) = S_I^*(\omega)S_J(\omega)$$

$$P_{XI}^{(1)} = \frac{1}{M} \sum P_{XI}(\omega)$$

$$P_{SI}^{(1)} = \frac{1}{M} \sum P_{SI}(\omega)$$

$$P_{XIJ}^{(1)} = \frac{1}{M} \sum P_{XIJ}(\omega)$$

$$P_{SIJ}^{(1)} = \frac{1}{M} \sum P_{SIJ}(\omega)$$

The computed results are given in Table (4-1) with various values of M and N. In the Table, the first three columns are the noise-signal ratio of the auto-power for the three channels defined in Eqs. (4-25b), and (4-26b). Data in the line with each M-number are the theoretical values defined in Eqs. (4-15), (4-17b), (4-20b), and (4-23b) with the same values of M, N and  $\cos(\theta_n) = .1$ .

#### IV-B. Investigation of Noise Effect on Impedance Elements

##### 1. General Solution

The surface impedance elements  $Z_{ij}$  are computed from the 4 pairs [(1,2), (1,3), (2,4), (3,4)] of equations (III-B-3)

$$\text{eqn. } L_x^* E_1 = Z_{ix} L_x^* H_x + Z_{iy} L_y^* H_y$$

## 22315 4-7. NOISE AND SIGNAL DATA

A(TE)-P(TE) 10

A(TE)-P(TE) 10

V<sub>124</sub>

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{14} \\ \text{V}_{12} \\ \text{V}_{15} \\ \text{V}_{19} \\ \text{V}_{15} \\ \text{V}_{27} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.14 & -.14\Delta & -.14 \\ -.12 & -.11 & -.12 \\ -.15 & -.13 & -.15 \\ -.19 & -.19 & -.19 \\ -.15 & -.15 & -.15 \\ -.27 & -.14 & -.14 \nabla \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{12} \\ \text{V}_{14} \\ \text{V}_{15} \\ \text{V}_{19} \\ \text{V}_{27} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.17 & -.17\Delta & -.17 \\ -.17 & -.14 & -.14 \\ -.18 & -.13 & -.13 \\ -.17 & -.17 & -.17 \\ -.17 & -.16 & -.16 \\ -.22 & -.14 & -.14 \nabla \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11} \\ \text{V}_{27} \\ \text{V}_{11}\Delta \end{array} \quad [(-.12) - (.12)]$$

$\begin{array}{ccc} -.01 & -.02\Delta & -.02 \\ -.07 & -.04 & -.04 \nabla \\ -.11\Delta & -.02 & -.02 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11} \\ \text{V}_{27} \\ \text{V}_{11}\Delta \end{array} \quad [(-.12) - (.12)]$$

$\begin{array}{ccc} -.08 & -.08 & -.08 \\ -.03 & -.03 & -.03 \\ -.09 & -.09\Delta & -.09 \\ -.09 & -.09 & -.09 \end{array}$

V<sub>124</sub>

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11}\Delta \\ \text{V}_{30} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{34}\nabla \end{array} \quad [(-.1) - (.1)\nabla]$$

$\begin{array}{ccc} -.21\Delta & -.03 & -.07 \\ -.30 & -.17 & -.14 \\ -.17 & -.03 & -.02 \\ -.17 & -.06\Delta & -.06 \\ -.34\nabla & -.06 & -.04 \\ -.17 & -.07 & -.07 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11}\Delta \\ \text{V}_{30} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{34}\nabla \end{array} \quad [(-.1) - (.1)\nabla]$$

$\begin{array}{ccc} -.11\Delta & -.17 & -.17 \\ -.11 & -.18 & -.18 \\ -.12 & -.03 & -.17 \\ -.12 & -.12 & -.12 \\ -.11 & -.21 & -.21 \\ -.11 & -.13 & -.13 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{22} \\ \text{V}_{12} \\ \text{V}_{12}\Delta \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.22 & -.26 & -.26 \\ -.12 & -.09\Delta & -.07 \\ -.12 & -.12 & -.12 \\ -.12\Delta & -.21 & -.21 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{22} \\ \text{V}_{12} \\ \text{V}_{12}\Delta \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.23 & -.24 & -.24 \\ -.15 & -.15\Delta & -.15 \\ -.15 & -.15 & -.15 \\ -.12 & -.21 & -.21 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{11} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.17 & -.12 & -.12 \\ -.17 & -.22\Delta & -.14 \\ -.17 & -.17 & -.21 \\ -.15 & -.06 & -.02 \\ -.07 & -.12 & -.04 \\ -.11 & -.11 & -.04 \nabla \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{17} \\ \text{V}_{11} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.17 & -.15 & -.15 \\ -.17 & -.25\Delta & -.17 \\ -.17 & -.16\Delta & -.17 \\ -.15 & -.06 & -.06 \\ -.07 & -.13 & -.06 \\ -.11 & -.26 & -.06 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11}\Delta \\ \text{V}_{11}\Delta \\ \text{V}_{11} \\ \text{V}_{11} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.11\Delta & -.04 & -.07 \\ -.11\Delta & -.05\Delta & -.05 \\ -.11 & -.01 & -.02 \\ -.11 & -.01 & -.02 \\ -.11 & -.02 & -.02 \end{array}$

$$\begin{array}{c} \text{V}_1 \\ \text{V}_{11}\Delta \\ \text{V}_{11}\Delta \\ \text{V}_{11} \\ \text{V}_{11} \end{array} \quad [(-.1) - (.1)]$$

$\begin{array}{ccc} -.11\Delta & -.04 & -.07 \\ -.11\Delta & -.05\Delta & -.05 \\ -.11 & -.01 & -.02 \\ -.11 & -.01 & -.02 \\ -.11 & -.02 & -.02 \end{array}$

$$(2) \quad E_y^* E_i = Z_{ix}^* E_y^* H_x + Z_{iy}^* E_y^* H_y \quad (4-30)$$

$$(3) \quad E_x^* E_i = Z_{ix}^* H_x^* H_x + Z_{iy}^* H_x^* H_y$$

$$(4) \quad E_y^* E_i = Z_{ix}^* H_y^* H_x + Z_{iy}^* H_y^* H_y \quad (i=x, y)$$

and 4 solutions for each element are obtained. These are represented by

$$Z_{ij}^l = \frac{N_{ij}^l}{D_{ij}^l} \quad (4-31)$$

where  $l (=1, 2, 3, 4)$  represents a given pair, and  $N_{ij}^l$  and  $D_{ij}^l$  are functions of the power estimates. The generalized power estimates are considered to contain the signal and the noise components, thus D and N can be represented symbolically by

$$N_{ij}^l = N_{ijo}^l + \Delta N_{ij}^l, \quad D_{ij}^l = D_{ijo}^l + \Delta D_{ij}^l$$

where  $N_{ijo}^l$  and  $D_{ijo}^l$  are the signal, and  $\Delta N_{ij}^l$  and  $\Delta D_{ij}^l$  are noise components.

When the noise-signal ratio is small, Eqn. (4-31) at first order becomes

$$\begin{aligned} Z_{ij}^l &= Z_{ijo}^l + \Delta Z_{ij}^l \\ \frac{N_{ij}^l}{D_{ij}^l} &= \frac{N_{ijo}^l + \Delta N_{ij}^l}{D_{ijo}^l + \Delta D_{ij}^l} = \frac{N_{ijo}^l}{D_{ijo}^l} + \frac{\Delta N_{ij}^l}{\Delta D_{ij}^l} \end{aligned} \quad (4-32)$$

$Z_{ij}^0 = N_{ijo}^\ell / D_{ijo}^\ell$  is the major component of the data and thus is a good approximation from which to deduce uniquely the resistivity structure of the earth.

$\Delta Z_{ij} = \frac{\Delta N_{ij}^\ell}{D_{ijo}^\ell} - \frac{N_{ijo}^\ell}{D_{ijo}^\ell} \frac{\Delta D_{ij}^\ell}{D_{ijo}^\ell}$  is due to the noise and varies

with different  $\ell$  value, since noise is random. For the case of  $\ell = (3, 4)$  and  $i = x$ ,  $N_{ij}$  and  $D_{ij}$  are:

$$D = \langle P_{Hx} \rangle \langle P_{Hy} \rangle - \langle H_x^* H_y \rangle \langle H_y^* H_x \rangle$$

$$\Delta D = \langle P_{Hx} \rangle \langle P_{Hy} \rangle_n + \langle P_{Hy} \rangle \langle P_{Hx} \rangle_n - \langle H_x^* H_y \rangle \langle H_y^* H_x \rangle_n - \langle H_y^* H_x \rangle \langle H_x^* H_y \rangle_n$$

$$N_{xx} = \langle P_{Hy} \rangle \langle H_x^* E_x \rangle - \langle H_y^* E_x \rangle \langle H_x^* H_y \rangle$$

$$\Delta N_{xx} = \langle P_{Hy} \rangle \langle H_x^* E_x \rangle_n + \langle H_x^* E_x \rangle \langle P_{Hy} \rangle_n - \langle H_y^* E_x \rangle \langle H_x^* H_y \rangle_n - \langle H_x^* H_y \rangle \langle H_y^* E_x \rangle_n$$

$$N_{xy} = \langle P_{Hx} \rangle \langle H_y^* E_x \rangle - \langle H_x^* E_x \rangle \langle H_y^* H_x \rangle$$

$$\Delta N_{xy} = \langle P_{Hx} \rangle \langle H_y^* E_x \rangle_n + \langle H_y^* E_x \rangle \langle P_{Hx} \rangle_n - \langle H_x^* E_x \rangle \langle H_y^* H_x \rangle_n - \langle H_y^* H_x \rangle \langle H_x^* E_x \rangle_n$$

After a tedious calculation, the first order approximation of the four  $\Delta Z_{ij}^\ell$  are written in the following forms, where the superscript denotes the pairs of equations used from (4-30).

$$\begin{aligned}
 \Delta Z_{xy}^{(1,2)} D_o^{(1,2)} &= \langle E_y^* H_x \rangle_o \langle P_{Ex} \rangle_n - \langle E_x^* H_x \rangle_o \langle E_y^* E_x \rangle_n \\
 &\quad - Z_{xy}^o [\langle E_y^* H_x \rangle_o \langle E_x^* H_y \rangle_n - \langle E_x^* H_x \rangle_o \langle E_y^* H_y \rangle_n] \\
 &\quad - Z_{xx}^o [\langle E_x^* H_x \rangle_o \langle E_y^* H_x \rangle_n - \langle E_y^* H_x \rangle_o \langle E_x^* H_x \rangle_n]
 \end{aligned}
 \tag{4-33a}$$

$$\begin{aligned}
 \Delta Z_{xy}^{(1,3)} D_o^{(1,3)} &= Z_{xx}^o \langle E_x^* H_x \rangle_o \langle P_{Hx} \rangle_n + \langle P_{Hx} \rangle_o \langle P_{Ex} \rangle_n \\
 &\quad - Z_{xy}^o [\langle P_{Hx} \rangle_o \langle E_x^* H_y \rangle_n - \langle E_x^* H_x \rangle_o \langle H_x^* H_y \rangle_n] \\
 &\quad - Z_{xx}^o \langle P_{Hx} \rangle_o \langle E_x^* H_x \rangle_n - \langle E_x^* H_x \rangle_o \langle E_x^* H_x \rangle_n
 \end{aligned}
 \tag{4-33b}$$

$$\begin{aligned}
 \Delta Z_{xy}^{(2,4)} D_o^{(2,4)} &= -Z_{xy}^o \langle E_y^* H_x \rangle_o \langle P_{Hy} \rangle_n + Z_{xy}^o \langle H_y^* H_x \rangle_o \langle E_y^* H_y \rangle_n \\
 &\quad + Z_{xx}^o [\langle H_y^* H_x \rangle_o \langle E_y^* H_x \rangle_n - \langle E_y^* H_x \rangle_o \langle H_y^* H_x \rangle_n] \\
 &\quad + \langle E_y^* H_x \rangle_o \langle H_y^* E_x \rangle_n - \langle H_y^* H_x \rangle_o \langle E_y^* E_x \rangle_n
 \end{aligned}
 \tag{4-33c}$$

$$\begin{aligned}
 \Delta Z_{xy}^{(3,4)} D_o^{(3,4)} &= -Z_{xy}^o \langle P_{Hx} \rangle_o \langle P_{Hy} \rangle_n + Z_{xx}^o \langle H_y^* H_x \rangle_o \langle P_{Hx} \rangle_n \\
 &\quad + Z_{xy}^o \langle H_y^* H_x \rangle_o \langle H_x^* H_y \rangle_n - Z_{xx}^o \langle P_{Hx} \rangle_o \langle H_y^* H_x \rangle_n \\
 &\quad + \langle P_{Hx} \rangle_o \langle H_y^* E_x \rangle_n - \langle H_y^* H_x \rangle_o \langle H_x^* E_x \rangle_n
 \end{aligned}
 \tag{4-33d}$$

$$\begin{aligned}
 \Delta Z_{\mathbf{xx}}^{(1,2)} D_{\mathbf{o}}^{(1,2)} = & - \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle P_{\mathbf{Ex}} \rangle_n + \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{y}}^* E_{\mathbf{x}} \rangle_n \\
 & - Z_{\mathbf{xy}}^o [ \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{y}}^* H_{\mathbf{x}} \rangle_n - \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{x}}^* H_{\mathbf{x}} \rangle_n ] \\
 & + Z_{\mathbf{yy}}^o [ \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_n - \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_n ]
 \end{aligned} \tag{4-33e}$$

$$\begin{aligned}
 Z_{\mathbf{xx}}^{(1,3)} D_{\mathbf{o}}^{(1,3)} = & - Z_{\mathbf{xx}}^o \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle P_{\mathbf{Hx}} \rangle_n - \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle P_{\mathbf{Ex}} \rangle_n \\
 & - Z_{\mathbf{xy}}^o [ \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_n - \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_n ] \\
 & + Z_{\mathbf{xx}}^o [ \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle E_{\mathbf{x}}^* H_{\mathbf{x}} \rangle_n + \langle E_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{x}}^* E_{\mathbf{x}} \rangle_n ]
 \end{aligned} \tag{4-33f}$$

$$\begin{aligned}
 Z_{\mathbf{xx}}^{(2,4)} D_{\mathbf{o}}^{(2,4)} = & Z_{\mathbf{xy}}^o \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle P_{\mathbf{Hy}} \rangle_n - Z_{\mathbf{xy}}^o \langle P_{\mathbf{Hy}} \rangle_o \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_n \\
 & - Z_{\mathbf{xx}}^o [ \langle P_{\mathbf{Hy}} \rangle_o \langle E_{\mathbf{y}}^* H_{\mathbf{x}} \rangle_n - \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{x}}^* H_{\mathbf{x}} \rangle_n ] \\
 & + \langle P_{\mathbf{Hy}} \rangle_o \langle E_{\mathbf{y}}^* E_{\mathbf{x}} \rangle_n - \langle E_{\mathbf{y}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{y}}^* E_{\mathbf{x}} \rangle_n
 \end{aligned} \tag{4-33g}$$

$$\begin{aligned}
 Z_{\mathbf{xx}}^{(3,4)} D_{\mathbf{o}}^{(3,4)} = & - Z_{\mathbf{xx}}^o \langle P_{\mathbf{Hy}} \rangle_o \langle P_{\mathbf{Hx}} \rangle_n + Z_{\mathbf{xy}}^o \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle P_{\mathbf{Hy}} \rangle_n \\
 & + Z_{\mathbf{xx}}^o \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{y}}^* H_{\mathbf{x}} \rangle_n - Z_{\mathbf{xy}}^o \langle P_{\mathbf{Hy}} \rangle_o \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_n \\
 & + \langle P_{\mathbf{Hy}} \rangle_o \langle H_{\mathbf{x}}^* E_{\mathbf{x}} \rangle_n - \langle H_{\mathbf{x}}^* H_{\mathbf{y}} \rangle_o \langle H_{\mathbf{y}}^* E_{\mathbf{x}} \rangle_n
 \end{aligned} \tag{4-33h}$$

where

$$D_o^{(2,4)} = \langle P_{Hy} \rangle_o \langle E_y^* H_x \rangle_o - \langle E_y^* H_y \rangle_o \langle H_y^* H_x \rangle_o$$

$$D_o^{(3,4)} = \langle P_{Hx} \rangle_o \langle P_{Hy} \rangle_o - \langle H_y^* H_x \rangle_o \langle H_x^* H_y \rangle_o$$

$$D_o^{(1,2)} = \langle E_y^* H_x \rangle_o \langle E_x^* H_y \rangle_o - \langle E_x^* H_x \rangle_o \langle E_y^* H_y \rangle_o$$

$$D_o^{(1,3)} = \langle P_{Hx} \rangle_o \langle E_x^* H_y \rangle_o - \langle E_x^* H_x \rangle_o \langle H_x^* H_y \rangle_o$$

(2) Impedance Solutions for Noise-free Cross-power Estimates

Supposing that noise exists in the auto-power estimate only, and not in the cross-power estimate as is often assumed, then all terms in Eqns. (4-33) having a noise cross-power contribution will vanish. With this assumption, and some rearranging to make all the D's real and positive Eqns. (4-33) can be rewritten as:

$$\Delta Z_{xy}^{(1,2)} = z_{xy}^o \frac{\langle P_{Ex} \rangle_n}{D \cdot 121} + z_{xx}^o \frac{\langle H_y^* H_x \rangle_o \langle P_{Ex} \rangle_n}{D \cdot 122} \quad (4-34a)$$

$$\Delta Z_{xy}^{(1,3)} = z_{xy}^o \frac{\langle P_{Ex} \rangle_n}{D \cdot 131} + z_{xx}^o \left( \frac{\langle H_y^* H_x \rangle_o \langle P_{Ex} \rangle_n}{D \cdot 132} + \frac{\langle H_y^* H_x \rangle_o \langle P_{Hx} \rangle_n}{D \cdot 133} \right) \quad (4-34b)$$

$$\Delta Z_{xy}^{(2,4)} = -z_{xy}^o \frac{\langle P_{Hy} \rangle_n}{D \cdot 241} \quad (4-34c)$$

$$\Delta Z_{xy}^{(3,4)} = -z_{xy}^o \frac{\langle P_{Hy} \rangle_n}{D \cdot 341} + z_{xx}^o \frac{\langle H_y^* H_x \rangle_o \langle P_{Hx} \rangle_n}{D \cdot 342} \quad (4-34d)$$

$$\Delta Z_{xx}^{(1,2)} = - z_{xx}^o \frac{\langle P_{Ex} \rangle}{D \cdot 123} + z_{xy}^o \frac{\langle H_x^* H_y \rangle_o \langle P_{Ex} \rangle}{D \cdot 124} \quad (4-34e)$$

$$\Delta Z_{xx}^{(1,3)} = - z_{xx}^o \frac{\langle P_{Ex} \rangle}{D \cdot 134} + \frac{\langle P_{Hx} \rangle}{D \cdot 135} - z_{xy}^o \frac{\langle H_x^* H_y \rangle_o \langle P_{Ex} \rangle}{D \cdot 136} \quad (4-34f)$$

$$\Delta Z_{xx}^{(2,4)} = + z_{xy}^o \frac{\langle H_x^* H_y \rangle_o \langle P_{Hy} \rangle}{D \cdot 242} \quad (4-34g)$$

$$\Delta Z_{xx}^{(3,4)} = - z_{xx}^o \frac{\langle P_{Hx} \rangle}{D \cdot 343} + z_{xy}^o \frac{\langle H_x^* H_y \rangle_o \langle P_{Hy} \rangle}{D \cdot 342} \quad (4-34h)$$

where

$$D \cdot 122 = | \langle E_x^* H_y \rangle |^2 - \langle E_x^* H_x \rangle \langle E_y^* H_y \rangle \langle H_x^* E_x \rangle / \langle E_y^* H_x \rangle$$

$$D \cdot 121 = D \cdot 122 / \langle P_{Hy} \rangle$$

$$D \cdot 124 = \langle E_y^* H_x \rangle \langle E_x^* H_y \rangle \langle H_x^* E_x \rangle / \langle E_y^* H_y \rangle - | \langle E_x^* H_x \rangle |^2$$

$$D \cdot 123 = D \cdot 124 / \langle P_{Hx} \rangle$$

$$D \cdot 132 = | \langle E_x^* H_y \rangle |^2 - \langle E_x^* H_x \rangle \langle H_x^* H_y \rangle \langle H_y^* E_x \rangle / \langle P_{Hx} \rangle$$

$$D \cdot 131 = D \cdot 132 / \langle P_{Hy} \rangle$$

$$D \cdot 133 = \langle P_{Hx} \rangle \langle E_x^* H_y \rangle \langle H_y^* H_x \rangle / \langle E_x^* H_x \rangle - | \langle E_x^* H_y \rangle |^2$$

$$D \cdot 135 = \langle P_{Hx} \rangle - \langle E_x^* H_x \rangle \langle H_x^* H_y \rangle / \langle E_x^* H_y \rangle$$

$$D \cdot 136 = \langle P_{Hx} \rangle \langle E_x^* H_y \rangle \langle H_x^* E_x \rangle / \langle H_x^* H_y \rangle - | \langle E_x^* H_x \rangle |^2$$

$$D \cdot 134 = D \cdot 136 / \langle P_{Hx} \rangle$$

$$D \cdot 241 = \langle P_{Hy} \rangle - \langle E_{Y Y}^* H_x^* \rangle \langle H_y^* H_x \rangle / \langle E_{Y Y}^* H_x^* \rangle$$

$$D \cdot 242 = \langle P_{Hy} \rangle \langle E_{Y X}^* H_x^* \rangle \langle H_x^* H_y \rangle / \langle E_{Y Y}^* H_x^* \rangle = | \langle H_x^* H_y \rangle |^2$$

$$D \cdot 342 = \langle P_{Hx} \rangle \langle P_{Hy} \rangle - \langle H_x^* H_y \rangle \langle H_y^* H_x \rangle$$

$$D \cdot 341 = D \cdot 342 / \langle P_{Hx} \rangle$$

$$D \cdot 343 = D \cdot 342 / P_{Hy}$$

If  $M$  is large, the noise component of the auto-power estimate as presented in Eqns. (4-34) is biased positive as can be seen from Eqn. (4-19)

$$\langle P_n \rangle = \frac{4\sigma^2}{N} \pm \frac{4\sigma s \langle K_n \rangle}{\sqrt{N} M} \quad (4-19)$$

The second term on the right hand side can be positive or negative whereas the first term is always positive.

An average taken from solutions of Eqns. (3-20) or (4-34)

$$\langle z_{ij} \rangle = z_{ij}^o + \frac{1}{L} \sum_{l=1}^L (\Delta z_{ij}^l) = z_{ij}^o + \langle \Delta z_{ij}^l \rangle \quad (4-35)$$

provides a better estimate.

As just described, the noise contribution to the auto-power estimate causes noise components  $\Delta z_{ij}^l$  in the impedance elements, and the average  $\langle \Delta z_{ij}^l \rangle$  can become more accurate than the original  $z_{ij}^l$ . Consequently,

this improved  $\langle z_{ij} \rangle$  is consistent with the existence of auto-power estimates which contain less noise than the original ones do. Thus, substituting  $\langle z_{ij} \rangle$  back to Eqn. (4-30) gives the following relations:

$$\langle P_{Ex} \rangle = R_x \quad (4-36a)$$

$$\langle P_{Ey} \rangle = R_y \quad (4-36b)$$

$$|\langle z_{xx} \rangle|^2 \langle P_{Hx} \rangle + |\langle z_{xy} \rangle|^2 \langle P_{Hy} \rangle + 2\operatorname{Re}[\langle z_{xx}^* \rangle \langle z_{xy} \rangle \langle H_x^* H_y \rangle] = R_x \quad (4-36c)$$

$$|\langle z_{yx} \rangle|^2 \langle P_{Hx} \rangle + |\langle z_{yy} \rangle|^2 \langle P_{Hy} \rangle + 2\operatorname{Re}[\langle z_{yx}^* \rangle \langle z_{yy} \rangle \langle H_x^* H_y \rangle] = R_y \quad (4-36d)$$

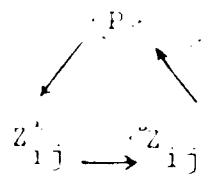
where

$$\begin{aligned} R_x &= \operatorname{Re}[\langle z_{xx} \rangle \langle E_x^* H_x \rangle + \langle z_{xy} \rangle \langle E_x^* H_y \rangle] \\ &= |\langle z_{xx} \rangle \langle E_x^* H_x \rangle + \langle z_{xy} \rangle \langle E_x^* H_y \rangle| \\ &= \frac{1}{2}\{\operatorname{Re}[\langle z_{xx} \rangle \langle E_x^* H_x \rangle + \langle z_{xy} \rangle \langle E_x^* H_y \rangle] + |\langle z_{xx} \rangle \langle E_x^* H_x \rangle + \langle z_{xy} \rangle \langle E_x^* H_y \rangle|\} \end{aligned}$$

$$\begin{aligned} R_y &= \operatorname{Re}[\langle z_{yx} \rangle \langle E_y^* H_x \rangle + \langle z_{yy} \rangle \langle E_y^* H_y \rangle] \\ &= |\langle z_{yx} \rangle \langle E_y^* H_x \rangle + \langle z_{yy} \rangle \langle E_y^* H_y \rangle| \\ &= \frac{1}{2}\{\operatorname{Re}[\langle z_{yx} \rangle \langle E_y^* H_x \rangle + \langle z_{yy} \rangle \langle E_y^* H_y \rangle] + |\langle z_{yx} \rangle \langle E_y^* H_x \rangle + \langle z_{yy} \rangle \langle E_y^* H_y \rangle|\} \end{aligned}$$

Solving Eqns. (4-36) for  $\langle P \rangle$  provides improved auto-power estimates corresponding to the more accurate  $\langle z_{ij} \rangle$ .

### A cyclic operation



(4-37)

can finally result in solution having the noise substantially removed.

$$\langle P \rangle \rightarrow \langle P \rangle_0, \quad \langle P \rangle_0 \rightarrow 0$$

$$\langle z_{ij} \rangle \rightarrow z_{ij}^0, \quad \langle \Delta z_{ij} \rangle \rightarrow 0$$

These final solutions are consistent with the conditions corresponding to a unit predicted coherency, which are

$$\begin{aligned} \langle P_{Ex} \rangle &= |\langle z_{xx} \rangle|^2 \langle P_{Hx} \rangle + |\langle z_{xy} \rangle|^2 \langle P_{Hy} \rangle \\ &\quad + 2 \operatorname{Re}[\langle z_{xx}^* \rangle \langle z_{xy} \rangle \langle H_x^* H_y \rangle] \end{aligned} \quad (4-38a)$$

$$\begin{aligned} \langle P_{Ey} \rangle &= |\langle z_{yx} \rangle|^2 \langle P_{Hx} \rangle + |\langle z_{yy} \rangle|^2 \langle P_{Hy} \rangle \\ &\quad + 2 \operatorname{Re}[\langle z_{yx}^* \rangle \langle z_{yy} \rangle \langle H_x^* H_y \rangle] \end{aligned} \quad (4-38b)$$

If the cross-power estimates are noise-free, the

following conditions can be simultaneously achieved:

a) Predicted coherency is unity.

b) 4-solutions of  $z_{ij}^x$  are identical.

As a consequence the apparent resistivity curves are smooth.

Figs. (4-2) show the result of the operation of Eqn. (4-37). Data was taken from a typical section of MT record, M, at the high frequency end ranges from 1000 to 3000. Such an averaging may remove the noise contribution from the cross-power estimates and leave noise only in the auto-power estimates. The 4-solutions of  $Z_{ij}^l$  and their average  $\langle Z_{ij} \rangle$  are plotted from left to right in each row. The first row represents the raw  $Z_{ij}^l$  and the second row represents  $Z_{ij}^l$  after the number of cyclic operations indicated on the individual plots. The scatter in  $Z_{ij}^l$  appears in the raw data, however, after a certain number of the cyclic operations, the noise in the auto-power estimates is substantially removed, and then the solutions of  $Z_{ij}^l$  become unique.

### (3) Impedance Solutions as Noise in Both Auto- and Cross-power Estimates

When noise exists in both auto- and cross-power estimates, the solutions of  $\Delta Z_{ij}^l$  are those given by Eqns. (4-33), and an average value given by

$$\langle Z_{ij} \rangle = Z_{ij}^0 + \langle \Delta Z_{ij} \rangle$$

also provides a noise-reduction-effect. Consecutive manipulations as performed by Eqns. (4-36) and (4-37) results in a set of  $\langle Z_{ij} \rangle^m$ , where  $m = 1, 2, \dots, M_c$  the number of cycles performed. None of  $\langle Z_{ij} \rangle^m$  is

expected to be noise-free, and it is even possible that the noise content is increased. Whether an average given by

$$\langle z_{ij} \rangle^m = \frac{1}{Mc} \sum_{m=1}^{Mc} z_{ij}^m \quad (4-39)$$

can provide a noise-reduction-effect depends on the form of the noise contribution. However, it is significant that even when no noise reduction is effected the results can be used to establish a confidence limit, which can be used as a criterion for final data selection.

The confidence limit is defined by

$$CL = \frac{\Delta z_{ij}^M}{\langle z_{ij} \rangle^m} \quad (4-40)$$

Where  $\Delta z_{ij}^M$  is the maximum variation among various  $z_{ij}^m$ 's.

Fig. (4-3) shows the noise disturbance behavior in the four  $z_{ij}$ 's and their average  $\langle z \rangle$  for raw data and for the indicated cyclic operations. Data was selected from sections of MT records with high noise contents. The result indicates that data between t = 10 to 100 sec are acceptable for the example shown.

Fig.(4-4) shows the result which is effected by the cyclic operation of Eqn(4-37). Data was taken from the MT record at the site-10 for both short and long period bands. Fig.(4-4.1) shows the average apparent resistivity,  $\bar{\rho}_{xy}$ , corresponding to  $\sigma_{xy}^2$  of the raw data, which have low predicted coherencies. Predicted coherency and smoothness of the curve are usually used as criteria for final data selection. If predicted coherency  $< 0.9$  is taken to be the criterion, none of the data at this site will be accepted.

To examine the effects of the cyclic processing, the low coherency of this data is ignored and a smooth curve is drawn through the data (Fig.4-4.1). Fig.(4-4.2) shows the average apparent resistivity after processing by the cyclic operation. If  $CL - \Delta\rho/\rho > 10\%$  is taken to be the criterion, some of the data is now found to be acceptable, while other parts are still of no use. Using only this acceptable data gives an entirely different picture from the curve of Fig.(4-4.1). The results shown in figure (4-4.2) are also used in figure (5-5.10).

Fig. (4-2). Curves of  $\langle z_{xy}^2 \rangle$ ,  $\langle z_{yx}^2 \rangle$  and  $\langle z_{xy}z_{yx} \rangle$

for the case where noise exists only in the auto-power.

(is deduced from  $\langle z_{ij}^2 \rangle$  in arbitrary units).

Upper sets: before cyclic operation.

Lower sets: after 10 cyclic operations.

Col. 1-4 are for the pairs appropriate to Eqn (4-30)

Col. 5 is appropriate to the average  $\langle z_{ij}^2 \rangle$ .

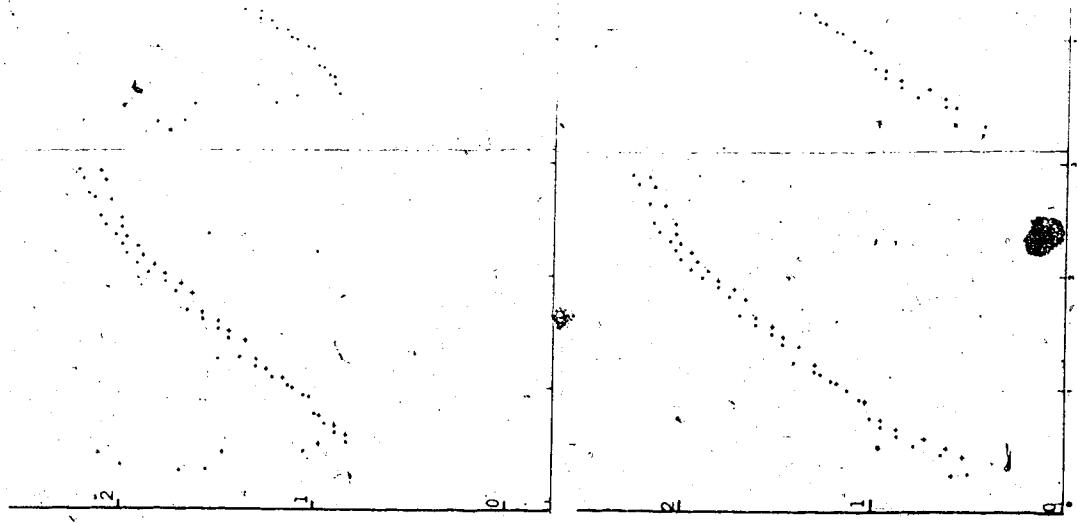
Fig. (4-3). Same as above for case when noise exists in both auto- and cross-power.

Fig. (4-4.1). Average apparent resistivity  $\rho_{xy}$  (corresponding to  $\langle z_{xy}^2 \rangle$ ) of raw data for both short (left) and long (right) period bands. The numbers are the predicted coherency values for each indicated period interval. The curves are estimated by taking all the data into account.

Fig. (4-4.2). Same data as above but after 10 cyclic operations. Acceptable data ( $CL \geq 10\%$ ) are barred. The curves are established by using this acceptable data, and are also used for interpretation as shown in

Fig. (5-5.10).

Fig. (4-2). LOG PERIOD (SEC)



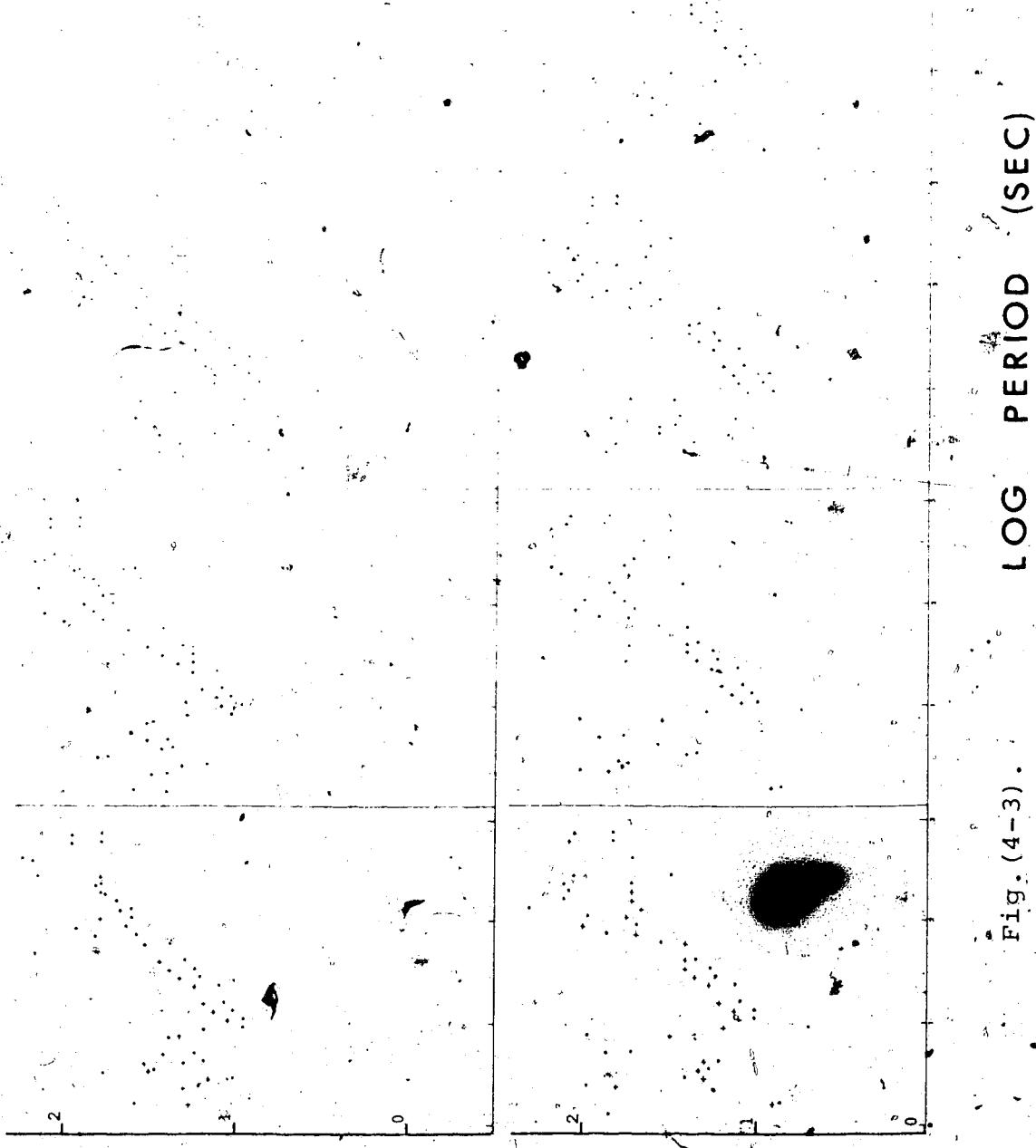


Fig. (4-3).

FIG. 4-3-1

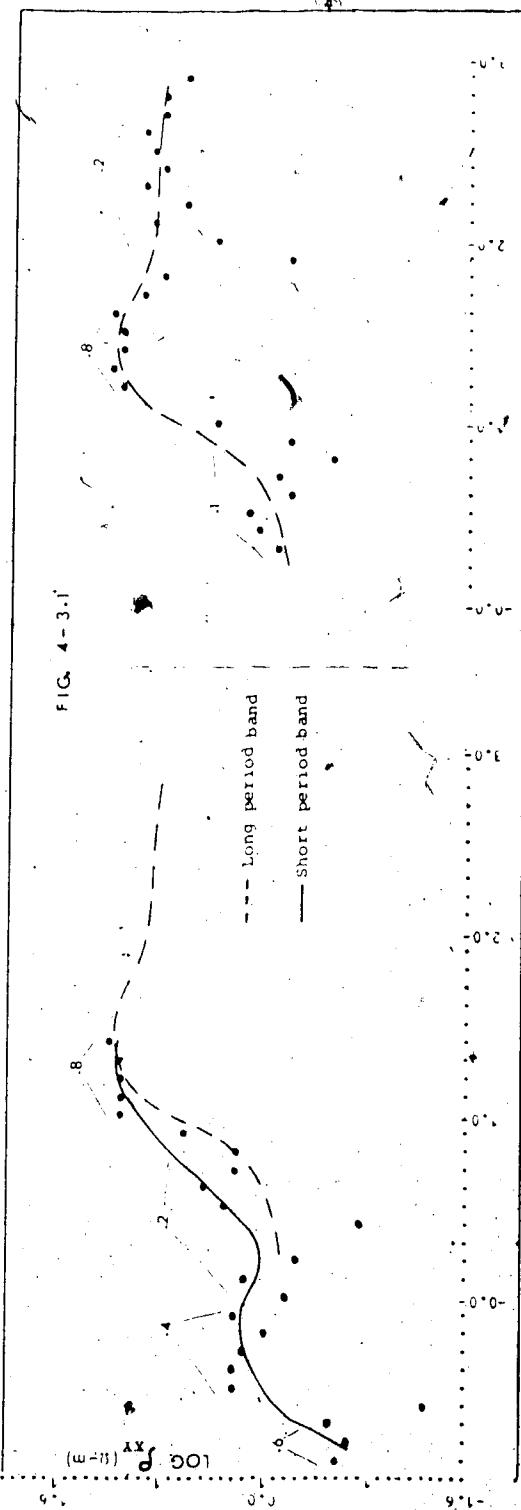
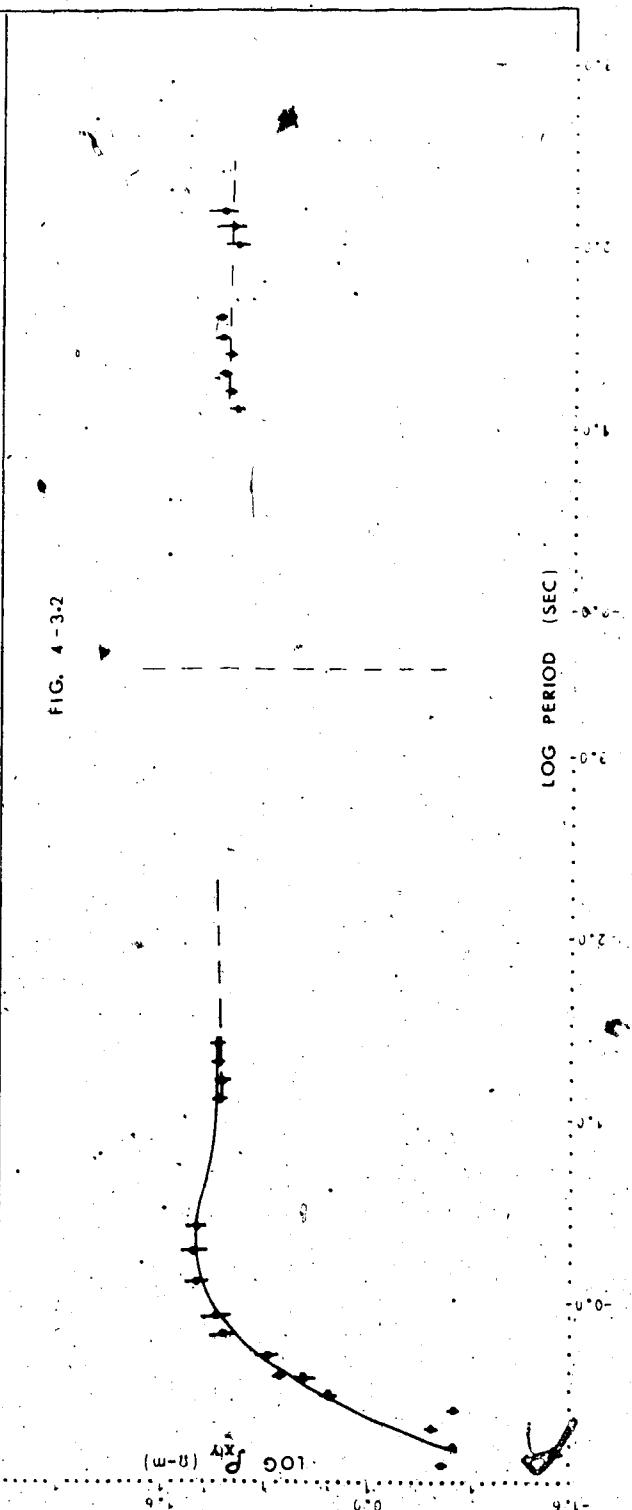


FIG. 4-3-2



## CHAPTER V

### MT SOUNDING

#### V-A. Geological Background

The Black Hills are a pronounced surface feature which rise several thousand feet above the surrounding Great Plains of North America. The location is shown in Fig.(5-1). This island like mountain structure which is approximately 200 km long and 100 km wide extends North West from the South West corner of South Dakota into the North East corner of Wyoming. A considerable number of geological and geophysical studies have been carried out in this region and it is possible to give a general geological description.

As shown in Fig.(5-1.2, 5-1.3, 5-1.4), the central Precambrian core of the Black Hills consists of a succession of highly folded schists, sedimentary in the main, but intruded by a succession of large and small granitic masses. Flanking this Algonkian mass are the upturned truncated edges of the sediments ranging in age from upper Cambrian to Tertiary. The dips of the sediments away from the centre are steeper on the East than on the West. Lidiak (1971) has mapped a metamorphic belt under the sediments and trending somewhat East of the strike of the Black Hills. This belt runs right through the central exposed metamorphic rocks of

the Black Hills, shown in Fig. (5-2).

Gough and Camfield (1971) have found from their deep sounding studies a very pronounced induction anomaly which is in striking agreement with Lidiak's metamorphic belt, as shown in Fig. (5-3). They attribute their results to a suitable graphitic schist in the basement. Mathisrud and Sumner (1967) have found highly conductive graphitic schists in the lead mine district of the Hills, but these were highly localized.

Three heat flow measurements made by Sass et al (1971), were either anomalously high or low. However these measurements were not of high quality and therefore no reliable deductions can be drawn from them. Surface manifestations of thermal activity such as hot springs and elevated ground water temperatures exist in many regions in and around the Black Hills.

The age dating results quoted by Lidiak place the Black Hills' orogeny at 1600 to 1800 m.y. ago. Various silicic intrusives associated with volcanic activity are dated at 1450 m.y. ago. It should be noted that the extension of the metamorphic belt mapped by Lidiak was deduced on the basis of geophysical measurements such as gravity and magnetic and is reasonably consistent with drill hole results and the aeromagnetic survey reported by Zietz et al (1971) and also shown in Fig. (5-2).

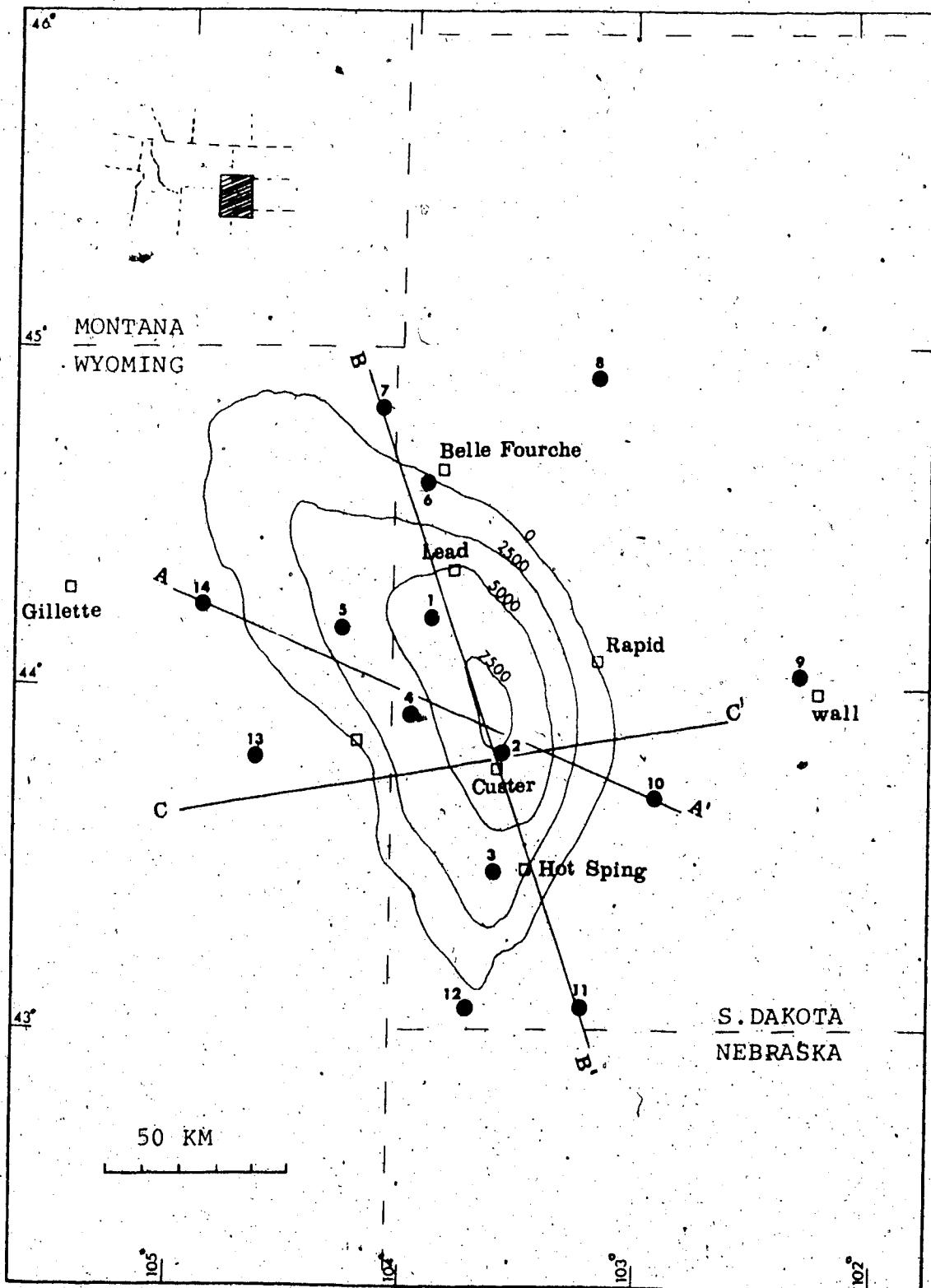
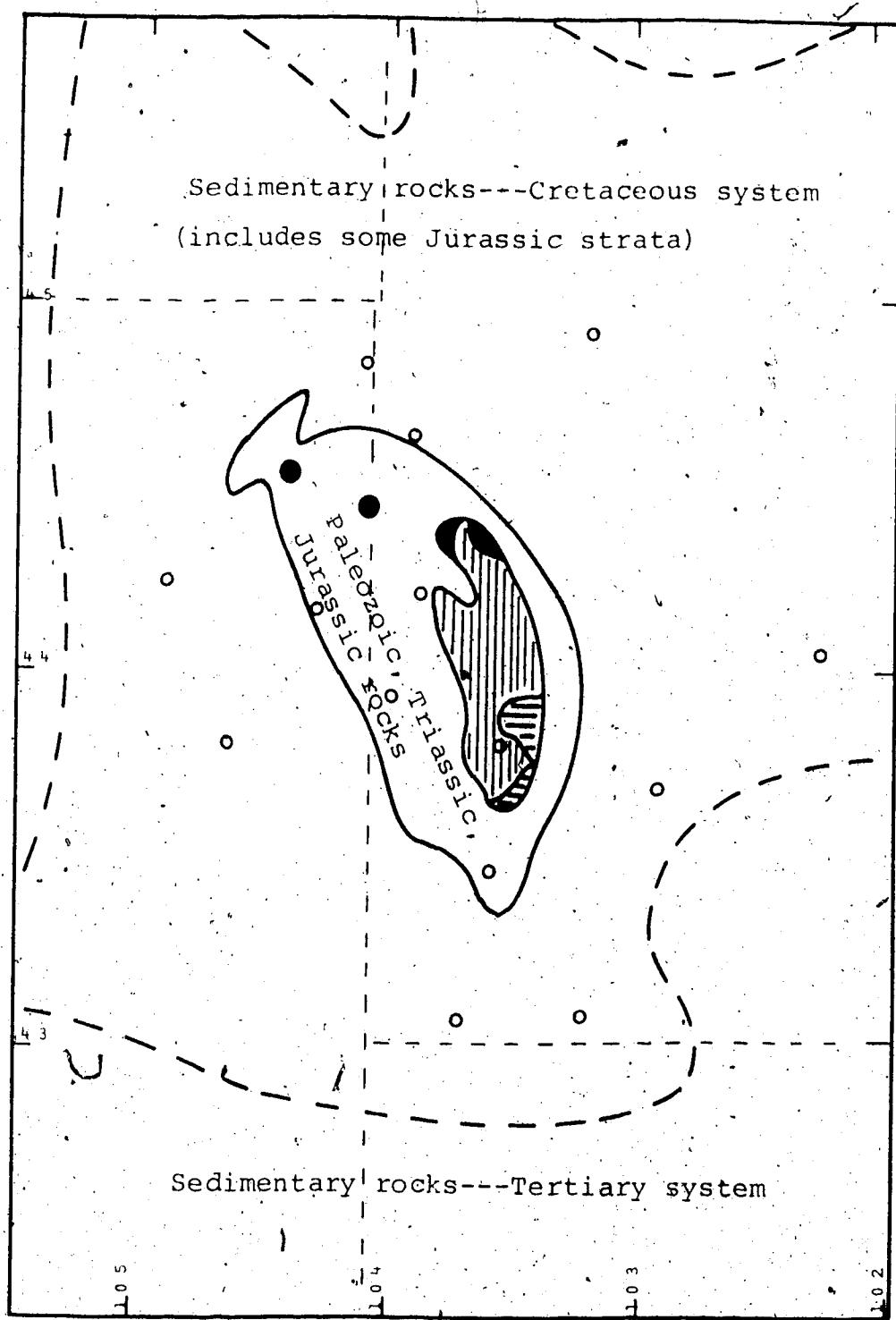


Fig. (5-1). Geological location of the Black Hills and sounding sites (●). Contour interval 2500-ft. Datum is sea level.



Fig(5-1.2) Areal geologic map of the Black Hills region.

- Precambrian rocks--- Sedimentary rocks.
- Precambrian rocks--- Igneous and metamorphic.
- Intrusive rocks --- Tertiary system.

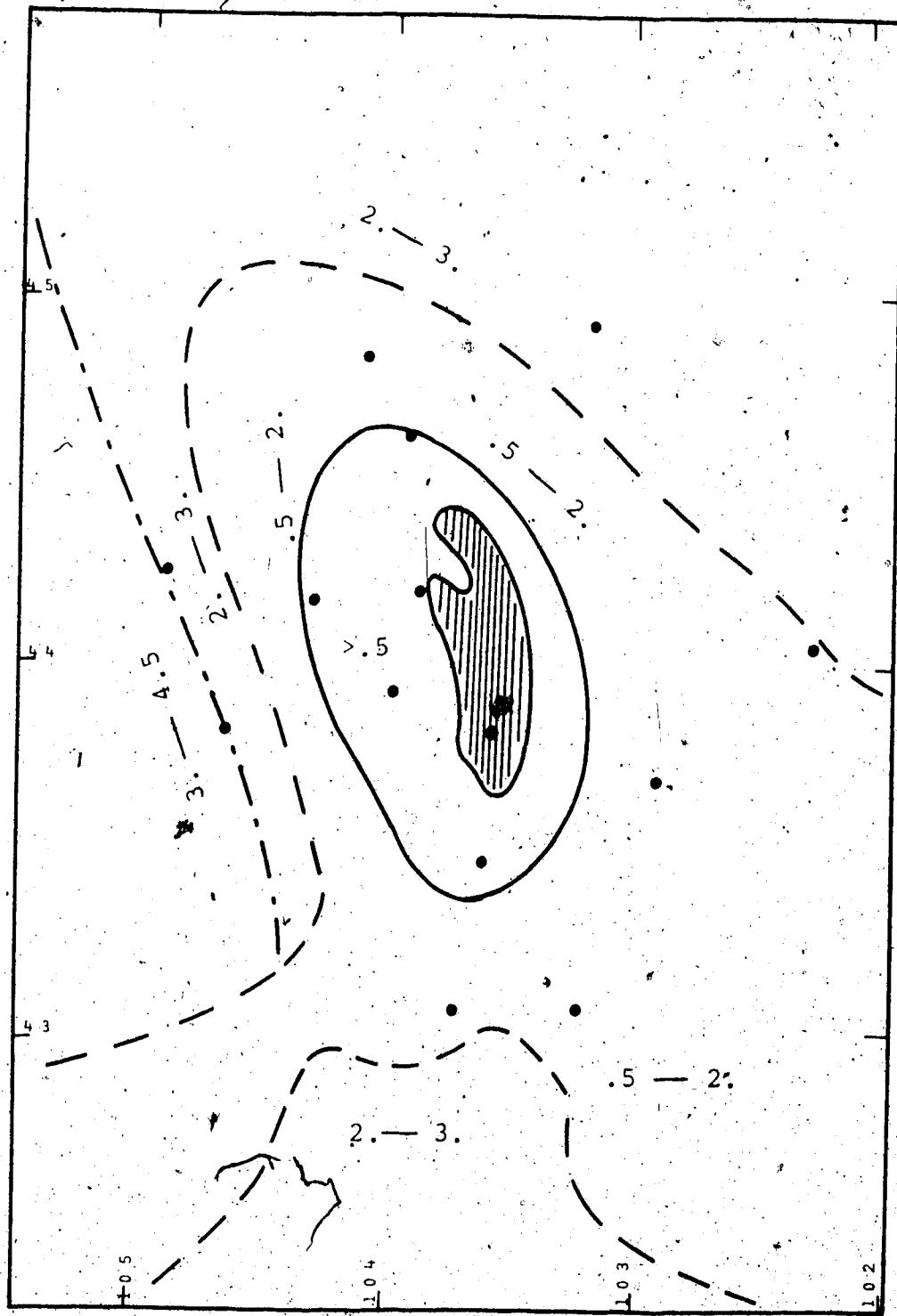


Fig.(5-1.3). Thickness of Phanerozoic rocks.

(Depth to Precambrian basement, in KM).



Black Hills uplift(Precambrian crystalline rocks exposed).

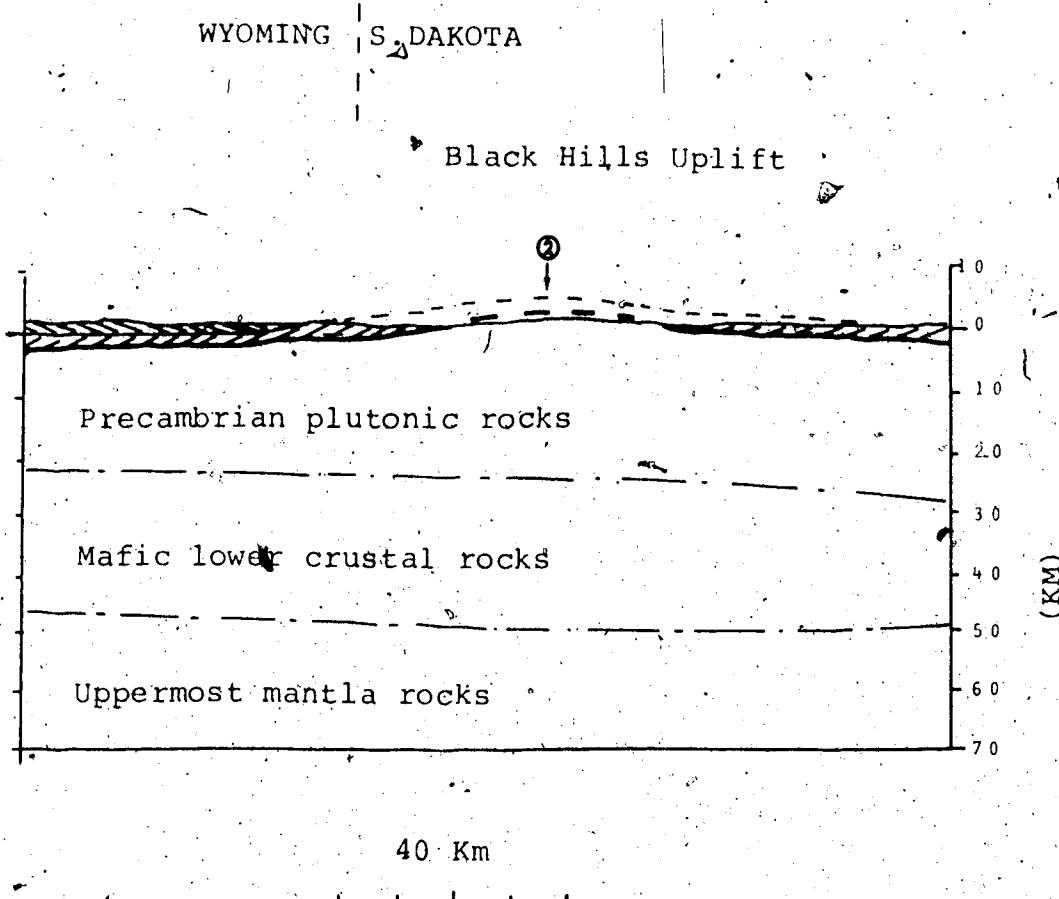
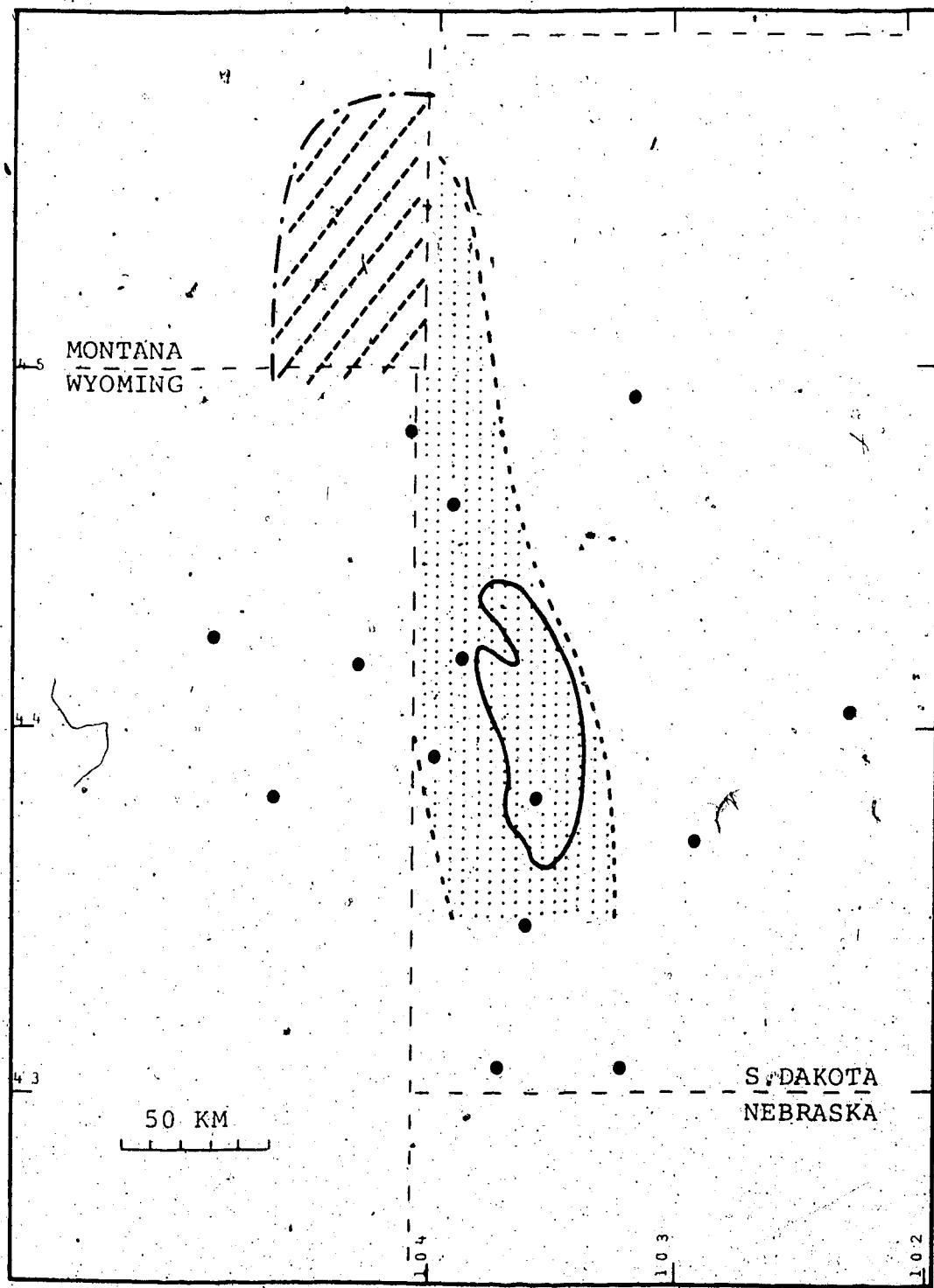


Fig.(5-1.3). Structural cross section along C-C'.

- Cenozoic sedimentary rocks, continental, detrital, with local volcanic rocks.
- Mesozoic sedimentary rocks, marine and continental, detrital.

[Fig. 5-1.2,3,4; Geologic Atlas of the Rocky Mountain region, Denver, Colorado, 1972]



Fig(5-2). The Black Hills uplift.

- Metamorphic belt (Lidiak 1971).
- Region of high magnetic intensity (Zietz et al.).

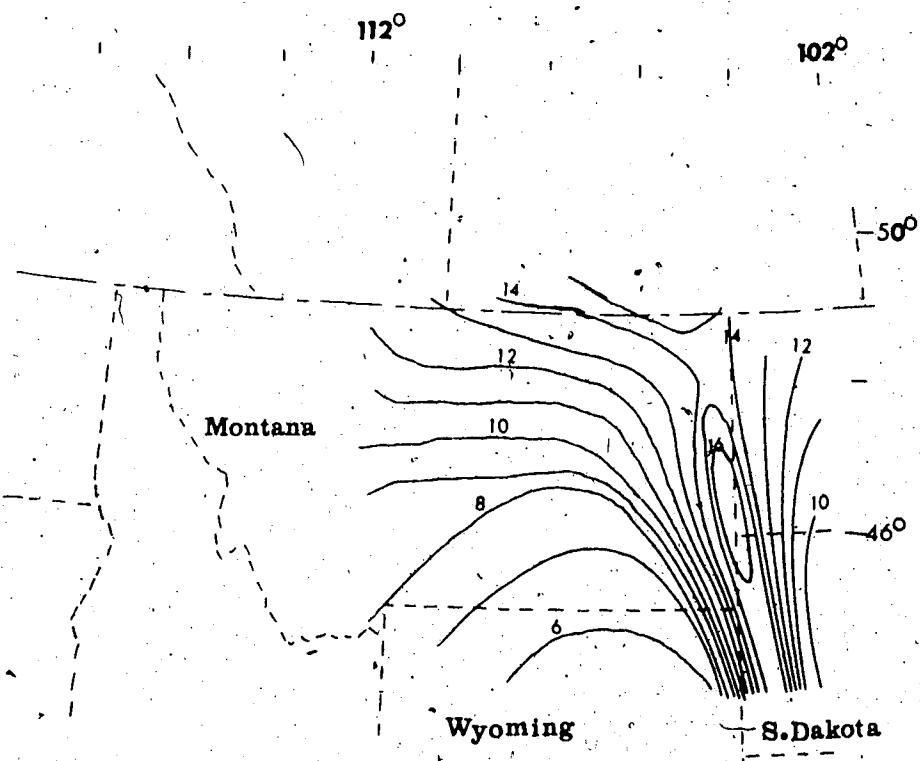


Fig. (5-3). Contour map of Fourier amplitudes of east-west component of magnetic field at period 48 min. Geomagnetic deep sounding result by Gough et al. (1971).

Since electromagnetic measurements are affected by structure lying within several skin depths an averaging effect is anticipated and thus one would expect to measure a relatively poorly conducting central region of the Black Hills characteristic of old crystalline material, surrounded by the relatively better conducting material characteristic of more recent sediments. The striking high conductivity anomaly in the central region of the Black Hills will be discussed in the following sections.

#### IV-B. Data Analysis and Computed Results

Records were selected visually to be of good amplitude and free of spikes and steps. Six data sets were usually obtained at each station in both high and low mode. Details of this are shown in Table (5-1). In all cases a data set consists of 4096 digitized values with the high mode digitized at 40/sec and the low at 1.25/sec. Following the techniques described in Chapter III, the Fourier coefficients were computed using a constant Q filter with  $M_b = .6$  at the low frequency end and  $M_b \approx 300$  at the high frequency end of the spectrum. Fig. (5-4) shows a pair of representative smoothed power density spectra for the E and H fields. The computed numerical results appear in the tables in Appendix B.

While it has been customary in the past to use only the criterion of high predicted coherency between the conjugate pairs of  $E$  and  $H$  as a basis for acceptance or rejection of data, the cyclic operation presented in Chapter IV permits the inclusion of a considerable amount of data with lower predicted coherencies in the initial stages of computation. The actual criterion adopted in the final stage of computation was that  $\rho_a$  was derived from the average values of  $z_{ij}^k$ , where the individual  $z_{ij}^k$  lie within  $2\frac{1}{2}\%$  of the average of the 4 possible values. The results of this process can be seen in Figs. (4-2) and (4-3). By this procedure 'good' data was obtained over the complete range of recorded frequencies, except for stations 3 and 9 where only a portion of the spectral range was acceptable.

The apparent resistivities are shown in the 'principal' directions in Figs. (5-5.1 to 5-5.14) for each station. Real principal directions actually exist only for genuinely two-dimensional structures. These directions are given for each station as  $\theta_o$  in Figs. (5-6.1) to (5-6.2).

In addition to the apparent resistivity, and its principal directions, the directional parameters  $\theta_{ze}$  and  $\theta_{zh}$  are also used. These are the directions which produce a maximum coherency between  $H_z$  and a component

Table 5-1 (L = long period band, S = short period band)

	Site	Period Band	No. of Data Set Average	Sample Length
1	Nahant	L	6	5.6 h 10. m
		S	6	
2	Custer	L	10	9.2 h 10. m
		S	6	
3	Kahta	L	7	6.5 h 7. m
		S	4	
4	Moon	L	10	9.2 h 10. m
		S	6	
5	Smith	L	6	5.6 h 10. m
		S	6	
6	Belle Fourche	L	10	9.2 h 10. m
		S	6	
7	Colony	L	6	5.6 h 10. m
		S	6	
8	Newell	L	6	5.6 h 10. m
		S	6	
9	Wall	L	5	4.6 h 10. m
		S	6	
10	Cheyenne River	L	6	5.6 h 10. m
		S	6	
11	Walker	L	10	9.2 h 10. m
		S	6	
12	Ardmore	L	6	5.6 h 7. m
		S	4	
13	Clareton	L	10	9.2 h 7. m
		S	6	
14	Morecraft	L	10	9.2 h 10. m
		S	6	

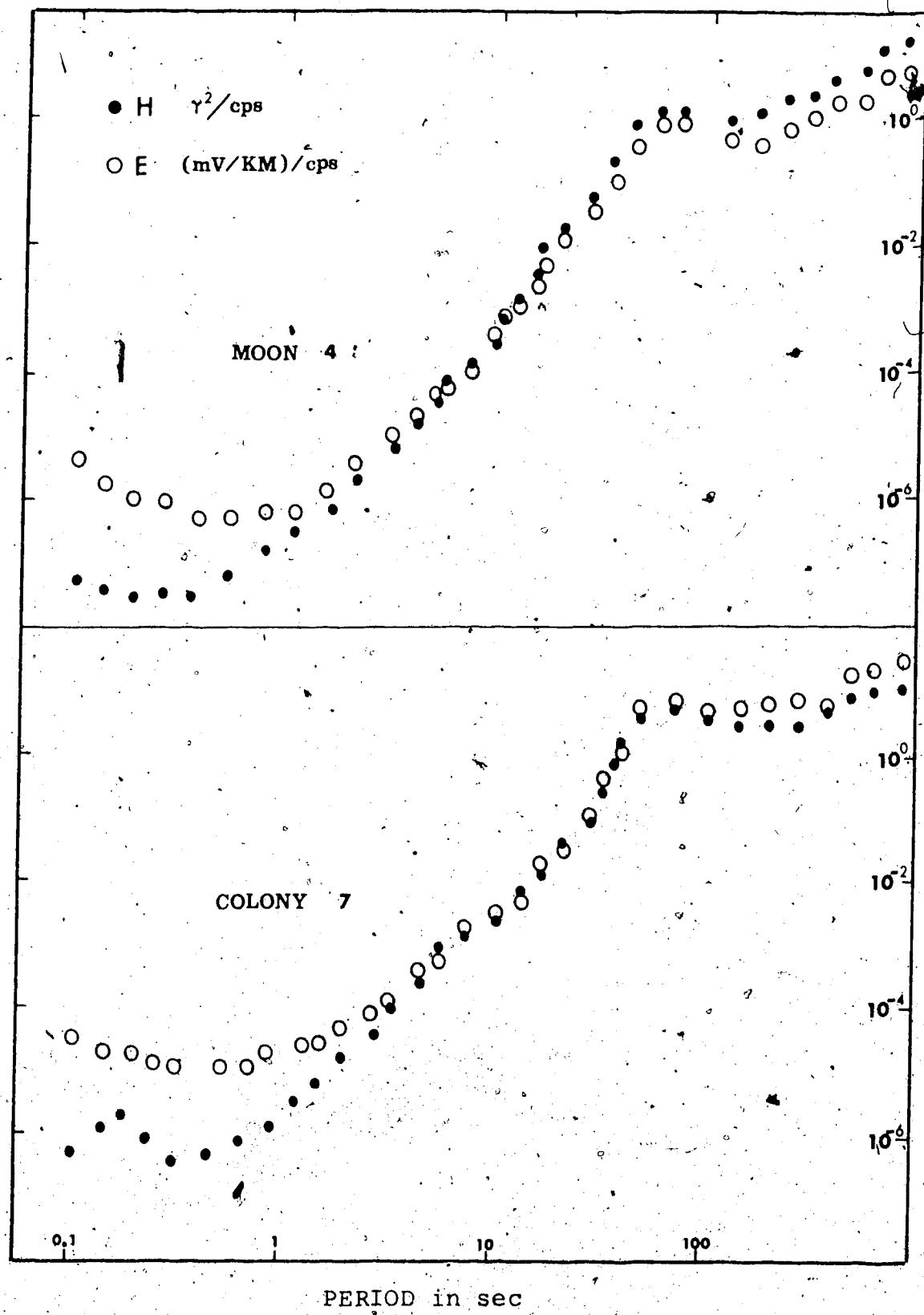
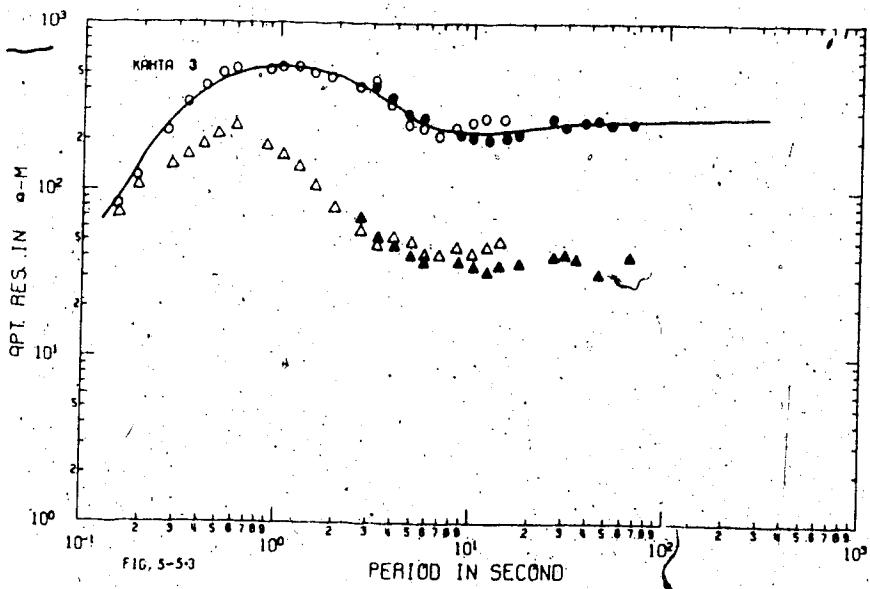
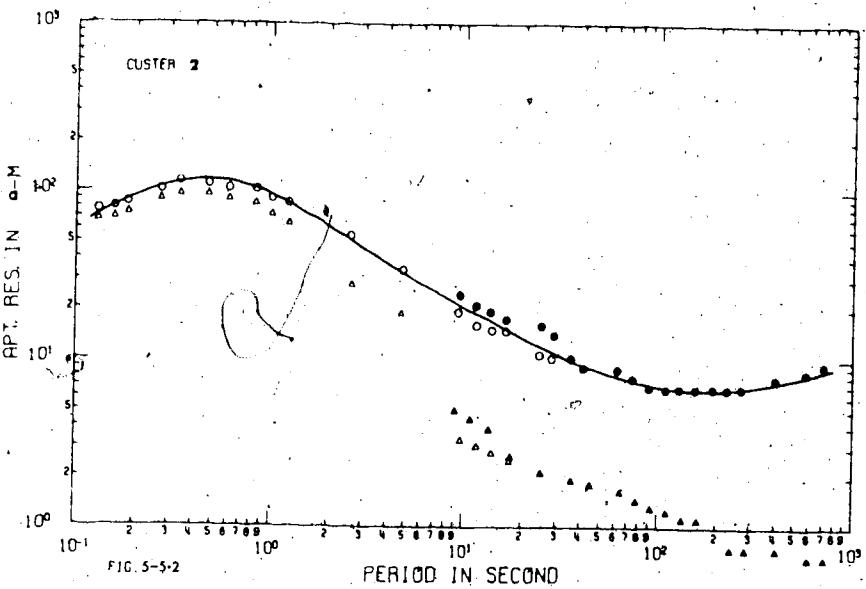
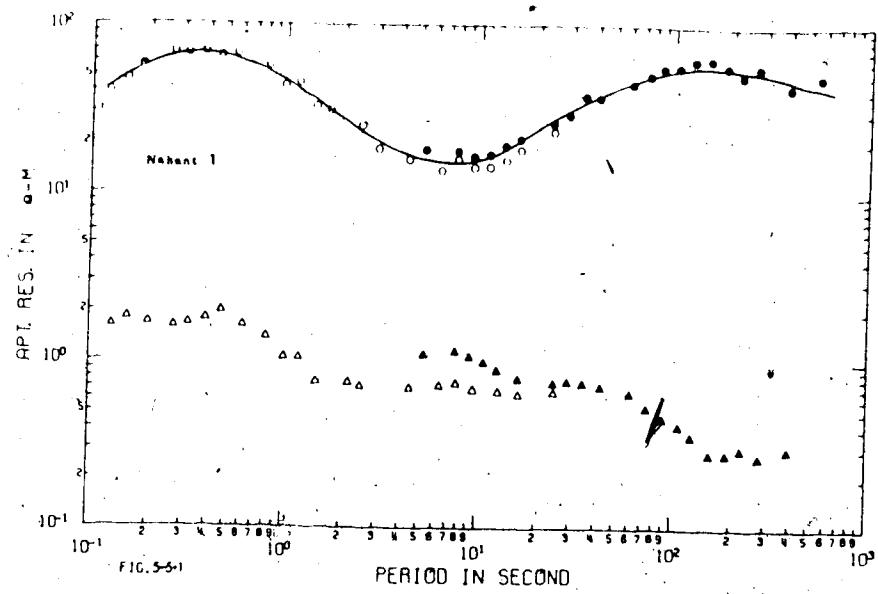


Fig.(5-4). Smoothed power spectral density.

Figs. (5-5). Apparent Resistivities in the Principal  
Directions vs Period.

- O --- ρ maximum of short period band
- Δ --- ρ minimum of short period band
- --- ρ' maximum of long period band
- ▲ --- ρ minimum of long period band
- --- Result of layered model fitting



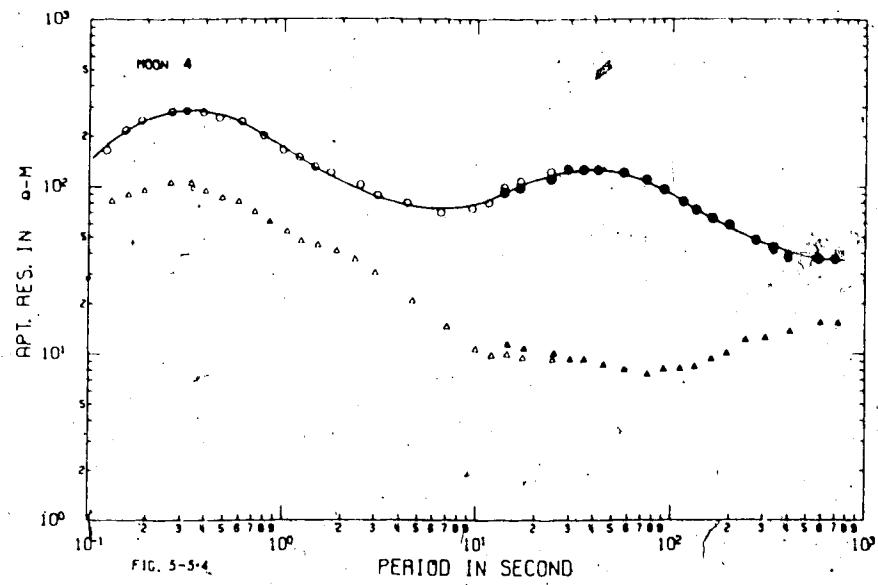


FIG. 5-5-4

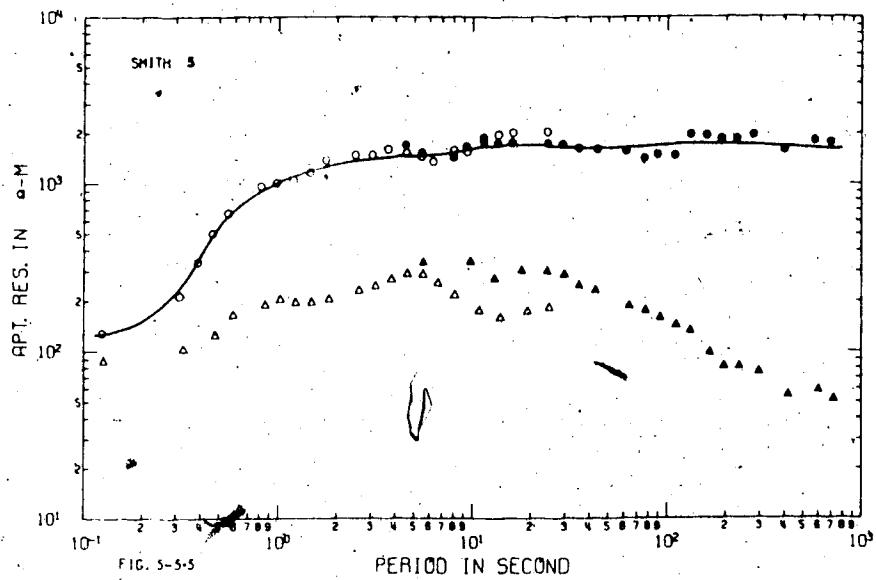


FIG. 5-5-5

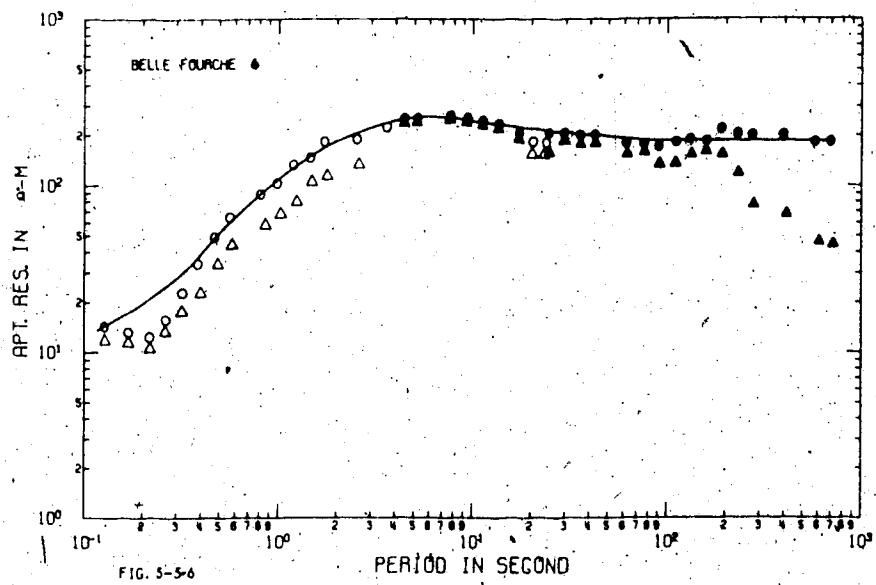
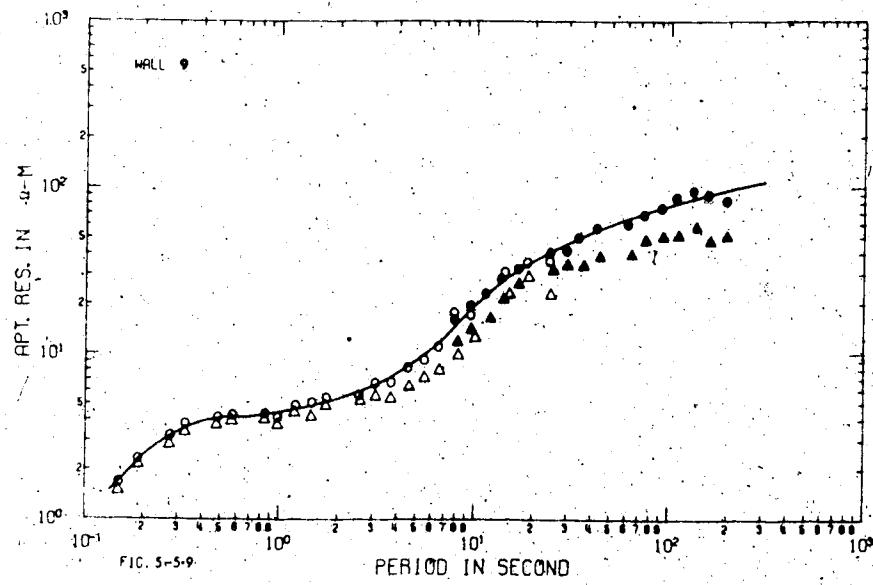
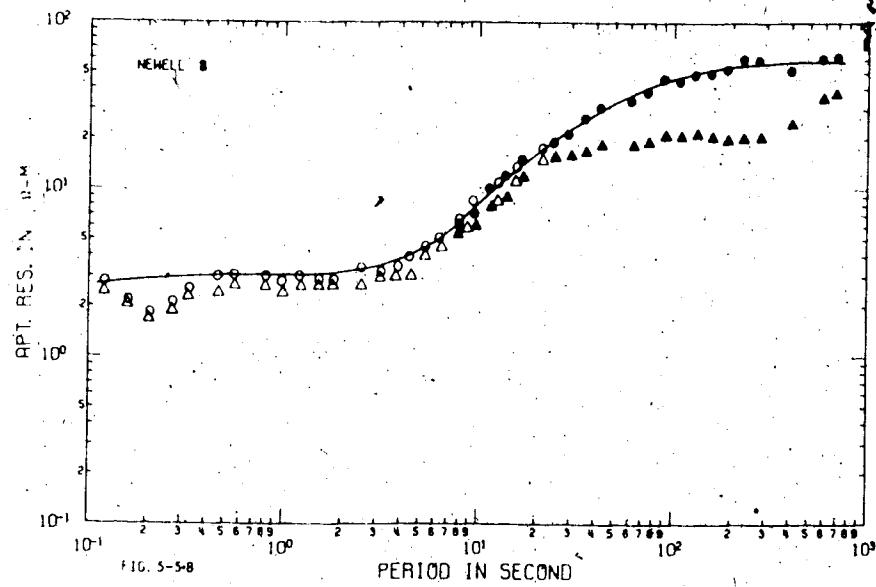
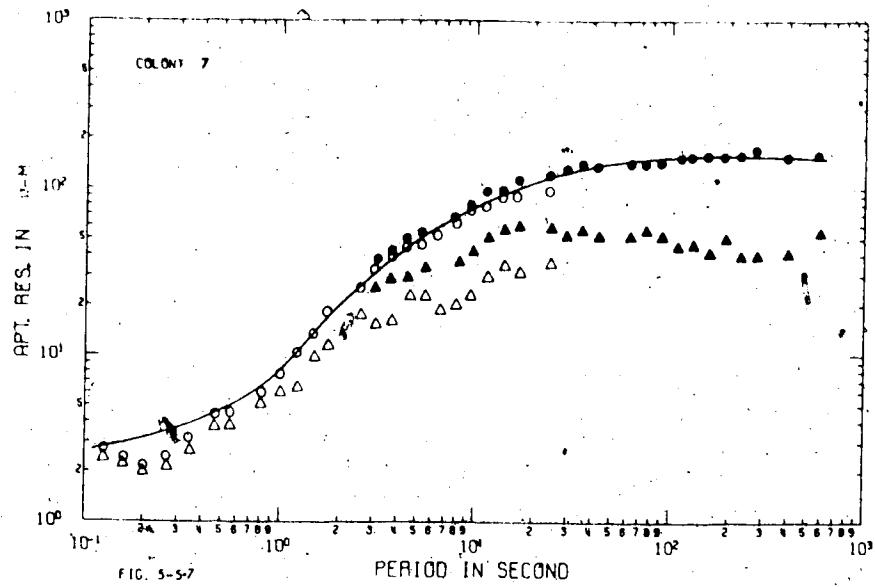
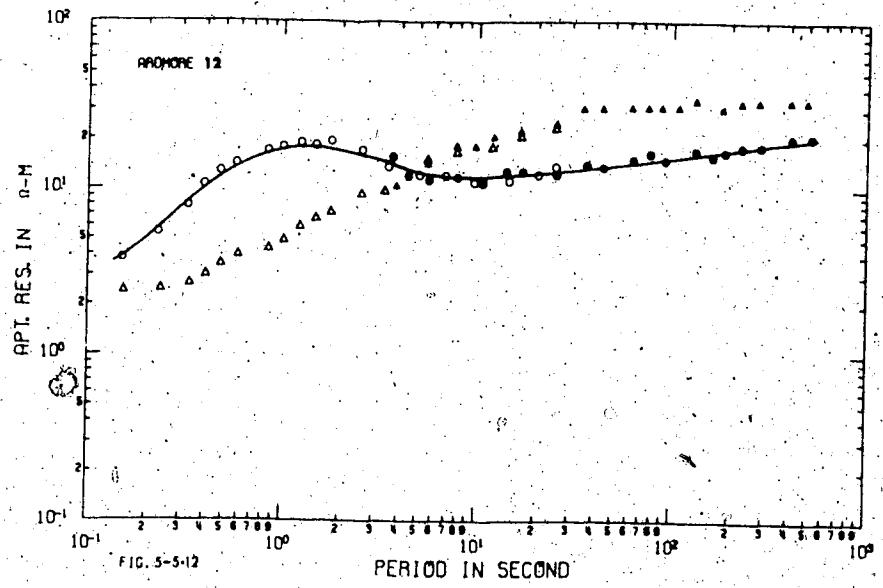
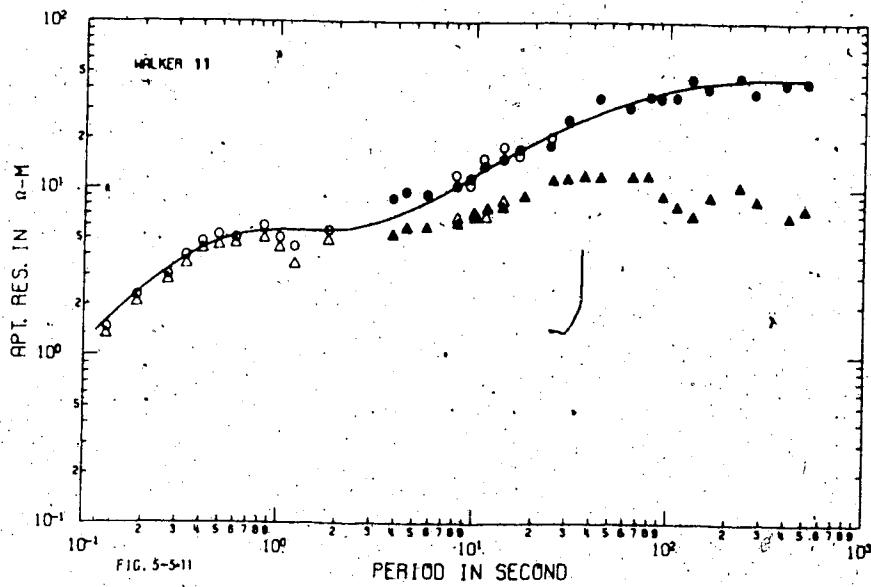
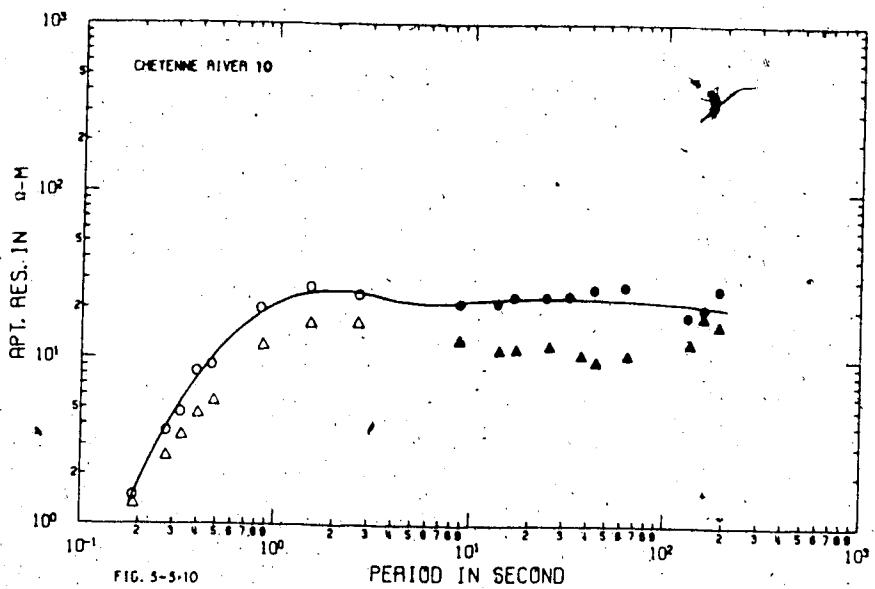
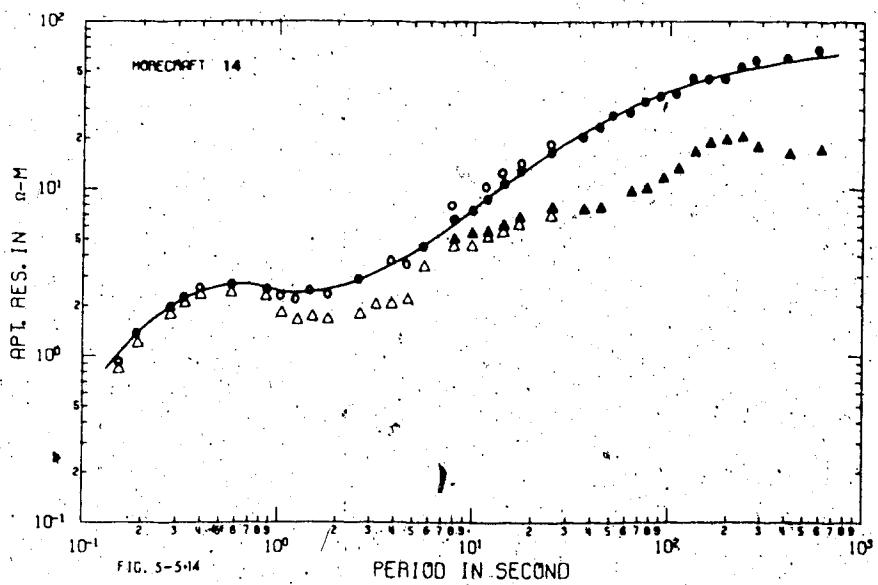
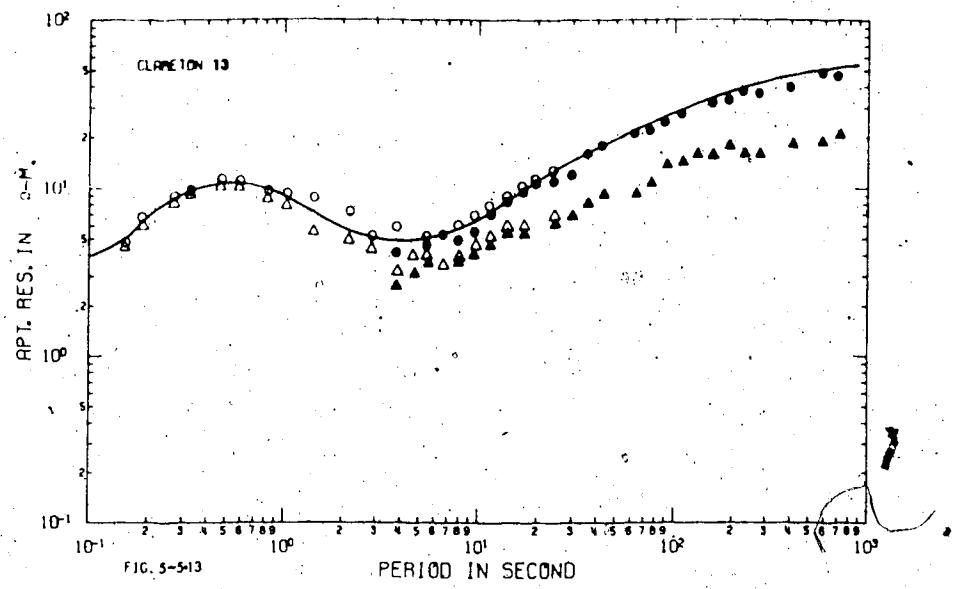


FIG. 5-5-6





107



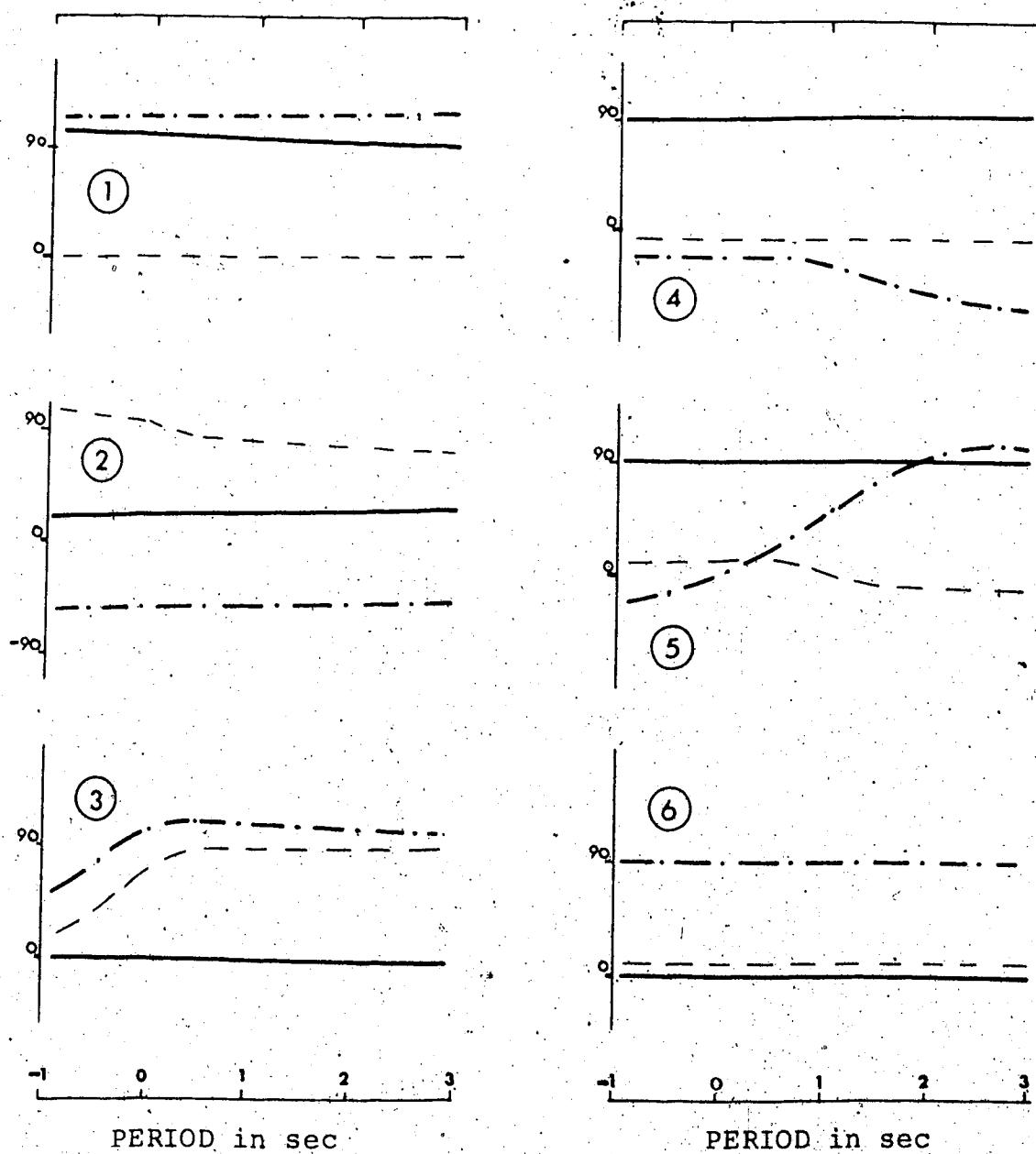


Fig.(5-6.1). Directions of maximum apparent resistivity  $\theta_o$  (—), maximum admittance  $\theta_{ze}$  (---), and  $\theta_{zh}$  (- - -) vs period. The numbers in the circles refer to the site number. (Directions in degree measuring from the North)

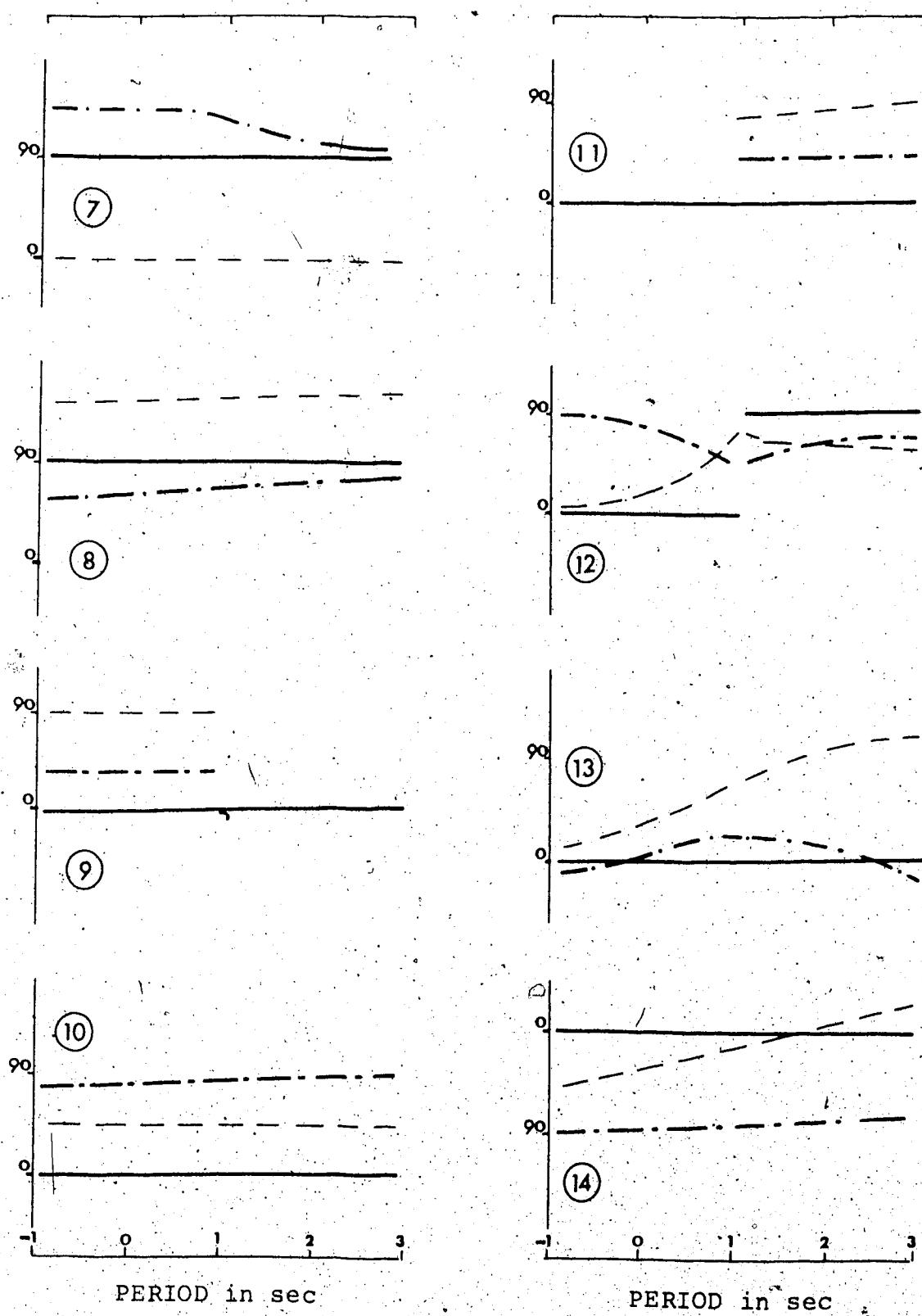


Fig.(5-6.2). (Same as figure 5-6.1).

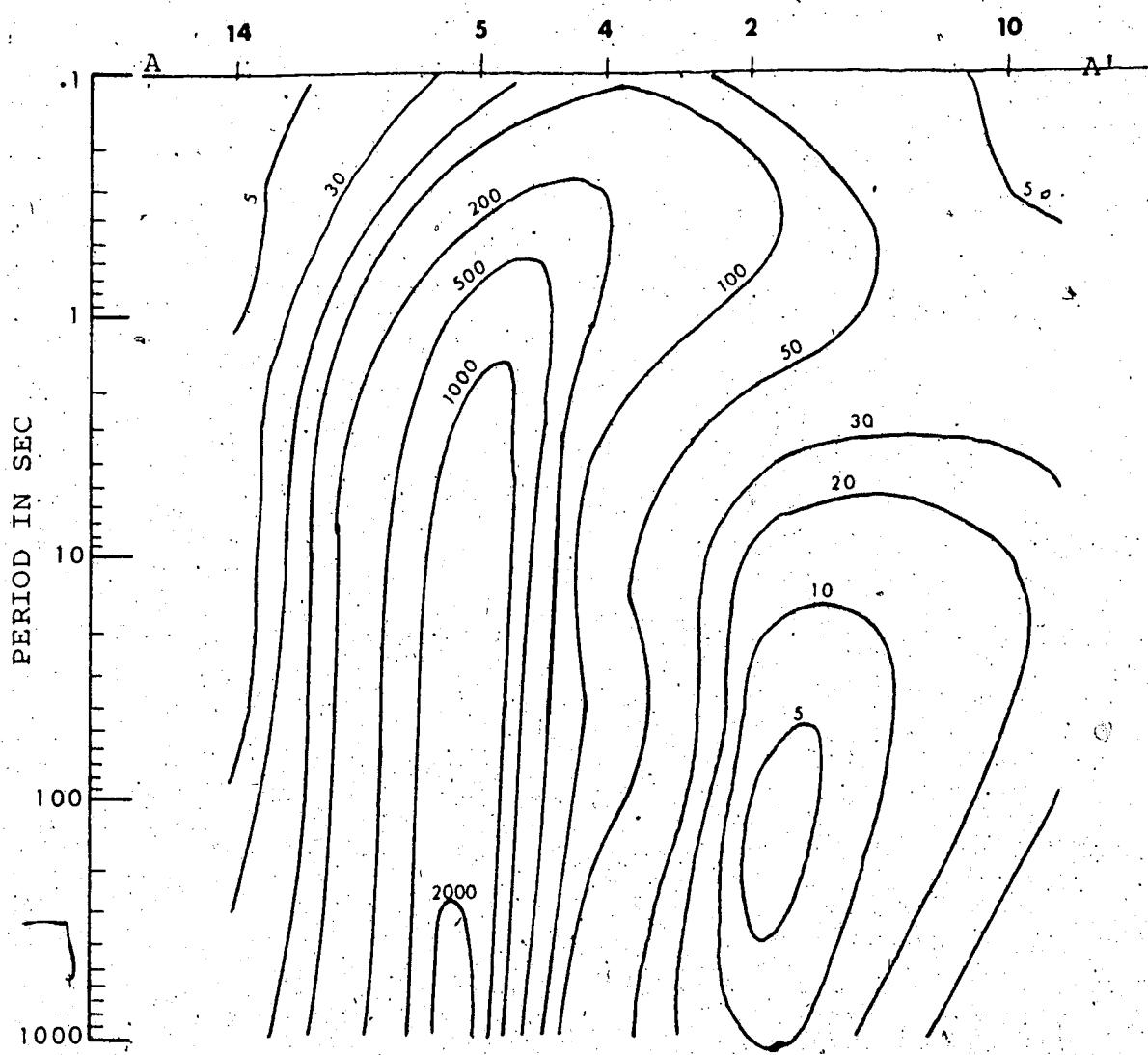


Fig.(5-7.1). Contour map of maximum apparent resistivity  
(in  $\Omega\text{-M}$ ) vs period along A-A' traverse.

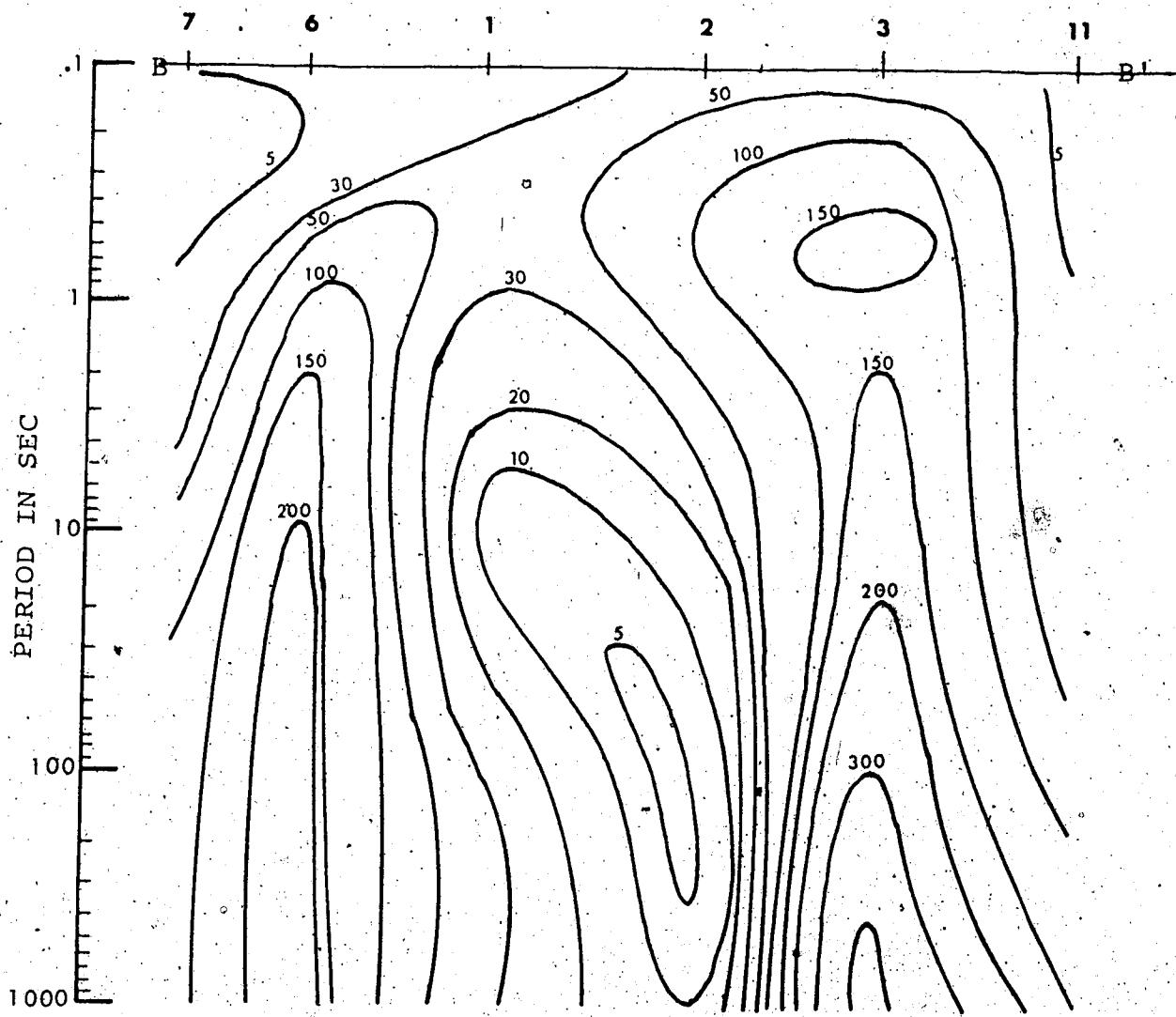


Fig. (5-7.2). Contour map of maximum apparent resistivity  
(in  $\Omega\text{-M}$ ) vs period along B-B' traverse.

of E and H respectively and are also shown in Figs.

(5-6.1, 5-6.2) for each station.

Pseudo-sections for the two traverses A-A' and B-B' indicated in Fig. (5-1) are shown in Figs. (5-7.1) and (5-7.2), where apparent resistivity is plotted as a function of period. These are converted to MT resistivity cross sections in Figs. (5-8) and (5-9) and will be discussed in the next section.

#### V-C. Interpretation and Discussion

##### (1) Discussions

The results, i.e. the anisotropic character of the apparent resistivity and the dimensional indicators, show that in the area studied, the conductivity structure is 3-dimensional. In the central region of the Hills, at sites 1-5, the two principal apparent resistivities diverge throughout most of the spectral range. The skew factor is large and also  $\theta_0$  does not coincide with either  $\theta_{ze}$  or  $\theta_{zh}$  as they would for a two dimensional earth. In the area external to the Hills, sites 7-14, the apparent resistivity curves coincide for short periods and diverge for longer periods. This is interpreted to imply that in this region the earth is homogeneous, isotropic, and layered near the surface and more complicated at depth. While the dimensional

indicators are in disagreement with this interpretation it must be considered that these indicators are much more sensitive to the minor contributions from 3-dimensional structure. On the east, west, and south sides around the Hills where sites 9-14 are located, the apparent resistivities indicate that below the conductive near surface layers, there is a relatively resistive layer, which in turn is underlain by a conductive third layer. In the area north of the Hills where sites 7 and 8 are located, it seems that the conductive third layer is not present. At the southern corner of the uplift at site 12, the principal axis of the apparent resistivity changes direction by  $90^\circ$  at  $t \sim 10$  sec as shown in Figs. (5-5.12) and (5-6.2); such a change only happens at this site.

## (2) Earth Model

Taking site 2 as an example and applying the one-dimensional approach, an apparent resistivity curve as shown in Fig. (5-5.2) is used to construct a four layer earth with the parameters as given in Table (5-2).

The first two layers would be acceptable due to the isotropic behavior of the apparent resistivities in the first decade of the period. The third layer has a very high conductivity and is located at approximately

3-10 KM, corresponding to a period ranging of about 1-500 sec. In this range the apparent resistivities appear anisotropic. Thus this anomaly may be placed at a distance of 3 KM from the measuring point but in an arbitrary direction in the half space. However, as shown in Fig. (5-7.2), for the same period range  $\theta_o$  is closer to  $\theta_{ze}$  rather than to  $\theta_{zh}$ . According to the results of the two-dimensional model calculations,  $\theta_o$  coinciding with  $\theta_{ze}$  is consistent with the measuring site being on the conductive side of the strike of a resistivity anomaly. Thus, this low resistivity anomaly is placed below the measuring site. This result is consistent with that of the group of stations in the anomalous region.

The model parameters of layered earth for each site are calculated and listed in Table (5-2). The maximum apparent resistivity  $\rho_M$  corresponds to the direction in which most of the power is concentrated and thus  $\rho_M$  is considered to be more reliable for curve fitting. The solid lines in Fig. (5-5) represent the model results.

Pseudo-depth, defined by  $\bar{z}(\omega) = \delta(\omega)$  are calculated using the skin depth theory for each one-dimensional profile. The pseudo-depth represents the depth reached by the field with its amplitude reduced by a factor  $1/e$  for a given frequency in a given

layered model. Contours of apparent resistivity and admittance vs pseudo-depth are referred to as MT resistivity cross sections. These are shown for the two given traverses in Figs. (5-8) and (5-9).

While the MT resistivity cross sections are not adequate for detailed interpretation, they do give a qualitative picture which is useful in the total interpretation.

Considering all the results, a final 3-dimensional resistivity section can be constructed. Figs. (5-10) show this construction for two traverses AA' and BB'. Figs. (5-11) show a plane view of the same resistivity distribution at horizontal planes at depths of 1-KM, and 5-KM, and 20-KM, respectively.

### (3) Summary and Comments

The cyclic operation developed in Chapter IV has proven very effective in extending the spectral range of data that can be utilized with a high confidence level. This has enabled a more detailed interpretation of the field data from the 1973 field season than had been heretofore considered possible.

In the area surrounding the Black Hills uplift, low resistivity with some layering is found in relatively shallow depths underlain by material of high resistivity. Within the central region was found a

Table 5-2. Parameters of One-dimensional Model

(R = Resistivity, H = Thickness)

<u>Site</u>	<u>R (<math>\Omega\text{-M}</math>)</u>	<u>H (KM)</u>	<u>Site</u>	<u>R (<math>\Omega\text{-M}</math>)</u>	<u>H (KM)</u>
Mahant (1)	30-50	.3	Newell (8)	4	1.5
	500	3.0		2000-3500	5.
	5	2.0		200	10.
	2000	30.		100	
	50				
Custer (2)	30-50	.3	Wall (9)	2	.1
	1000-3000	3.		35	.6
	5	7.		5	1.
	50			3000	30.
				30	
Kahta (3)	30-50	.3	Chey. R. (10)	4	.1
	1000-2000	8.		40-50	.1
	50	4.		200	4.
	50-400			5	1.
				50	
Moon (4)	30-50	.3	Walker (11)	1	.06
	2000-4000	3.		30-50	1.2
	30	6.		5	1.2
	1500	20.		1500	10.
	50			30	
Smith (5)	30-50	.5	Ardmore (12)	2	.1
	1000-3000	1.5		150	3.
	5000-10000	20.		4	1.
	1500			1000	10.
				30	
Belle Four. (6)	10	.3	Clareton (13)	2	.1
	2000-3000	10.		40	1.2
	100	3.		1	.5
	200			1000	10.
				100	
Colony (7)	4	.5	Morecraft (14)	2	.1
	2000-3000	20.		35	.5
	100			5	1.2
				1000-2000	10.
				100	

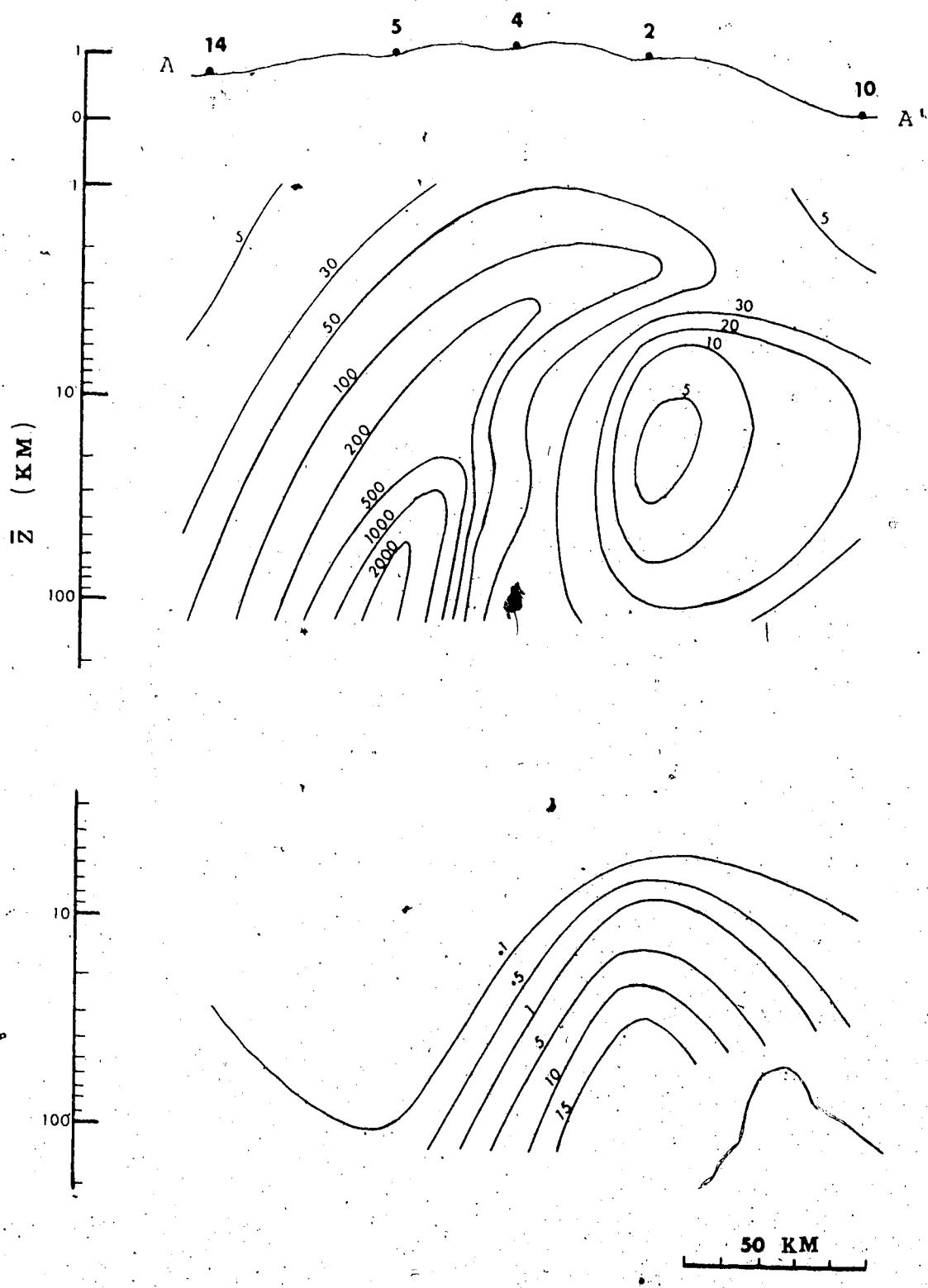


Fig. (5-8). Contours of maximum apparent resistivity (upper, in  $\Omega\text{-M}$ ) and maximum admittance (lower, in arbitrary units) vs  $\bar{z}$  (pseudo-depth) along A-A' traverse.

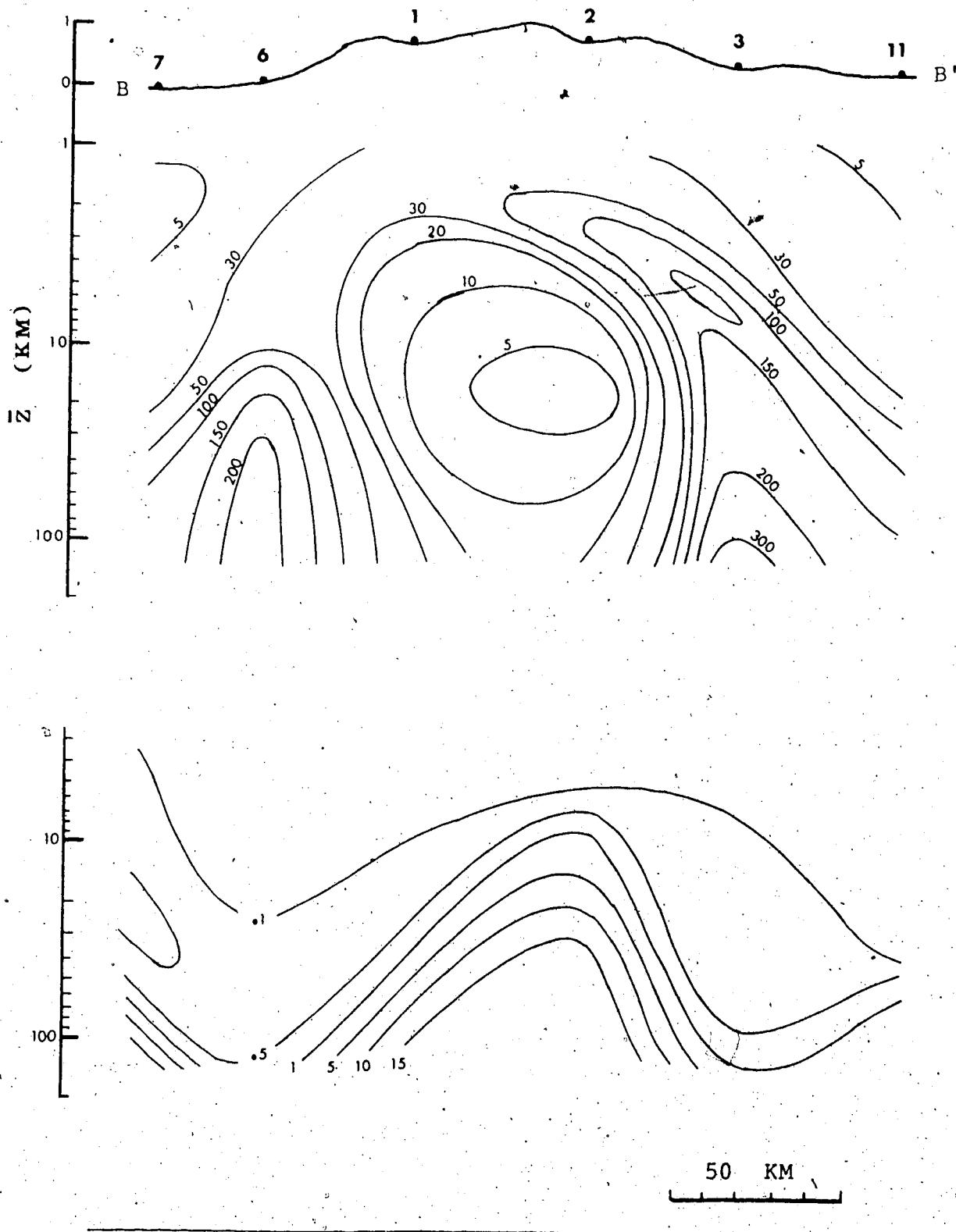


FIG. (5-9). Contours of maximum apparent resistivity (upper, in  $\Omega \cdot \text{M}$ ) and maximum admittance (lower, in arbitrary units) vs  $\bar{z}$  (pseudo-depth) along BB' traverse.

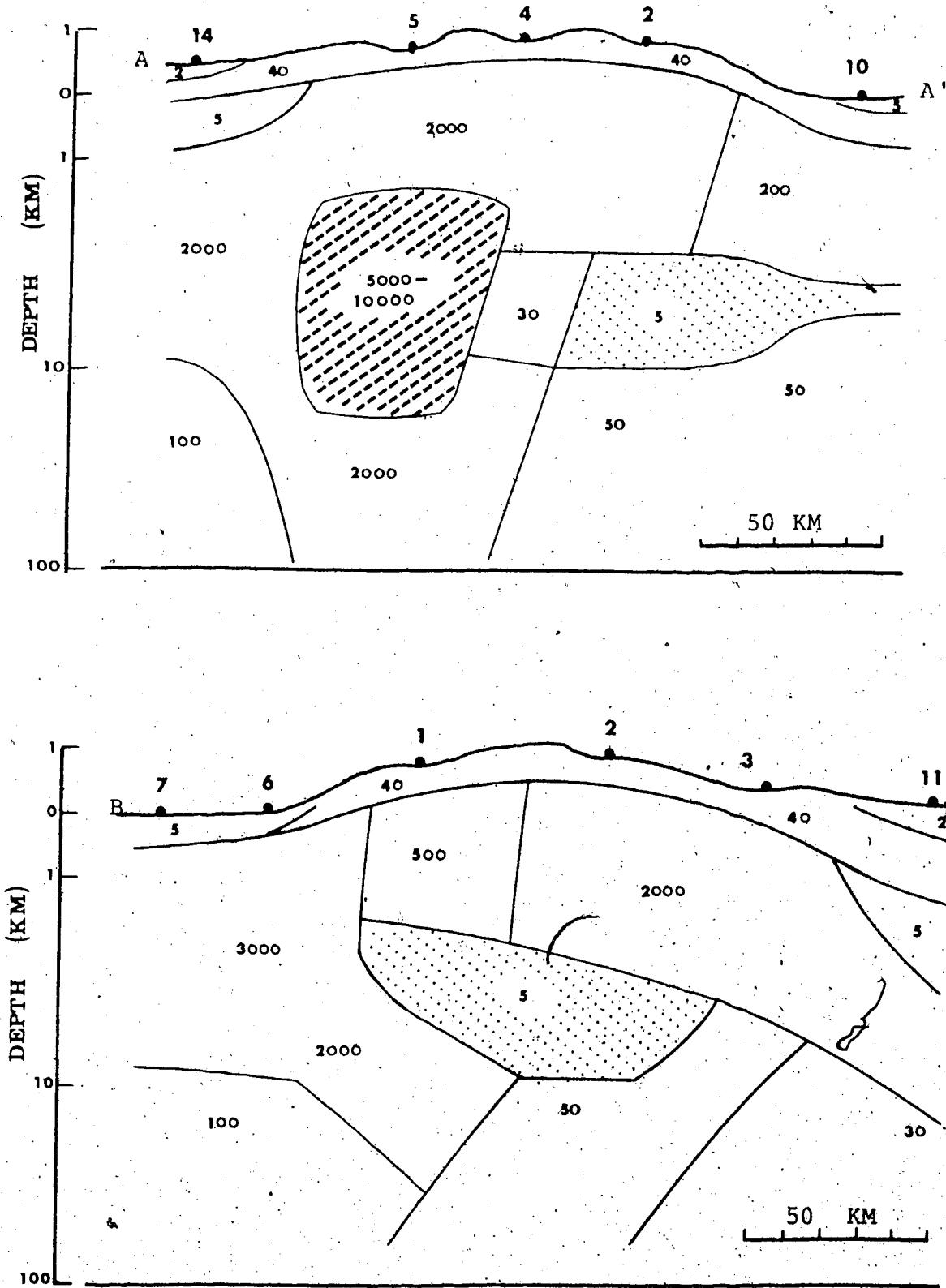


Fig. (5-10). Resistivity ( $\Omega\text{-M}$ ) distribution along A-A'  
(upper) and B-B' (lower) traverses.

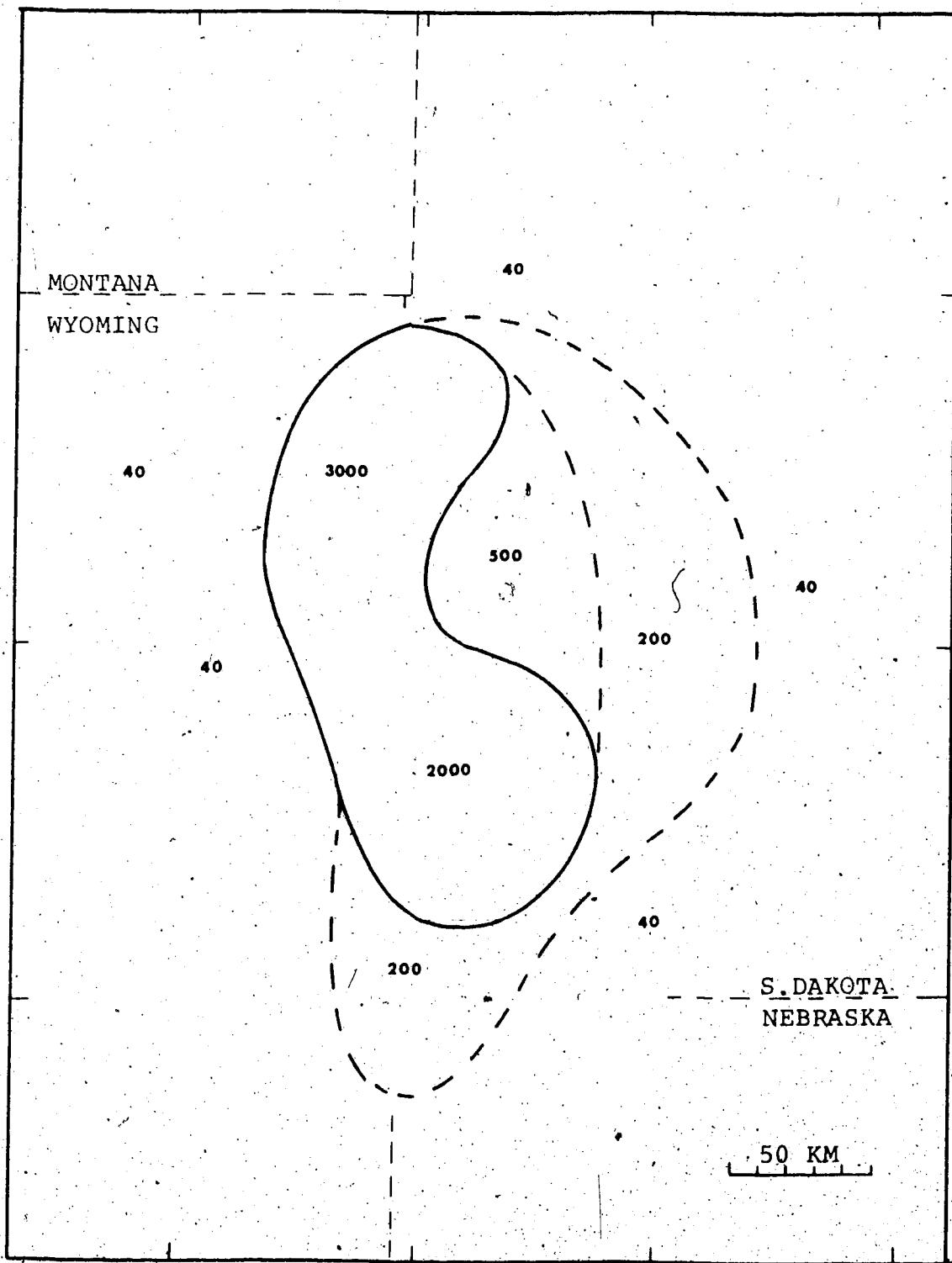


Fig. (5-11.1). Resistivity ( $\Omega\text{-M}$ ) distribution in a horizontal plane at a depth of 1-KM from the surface.

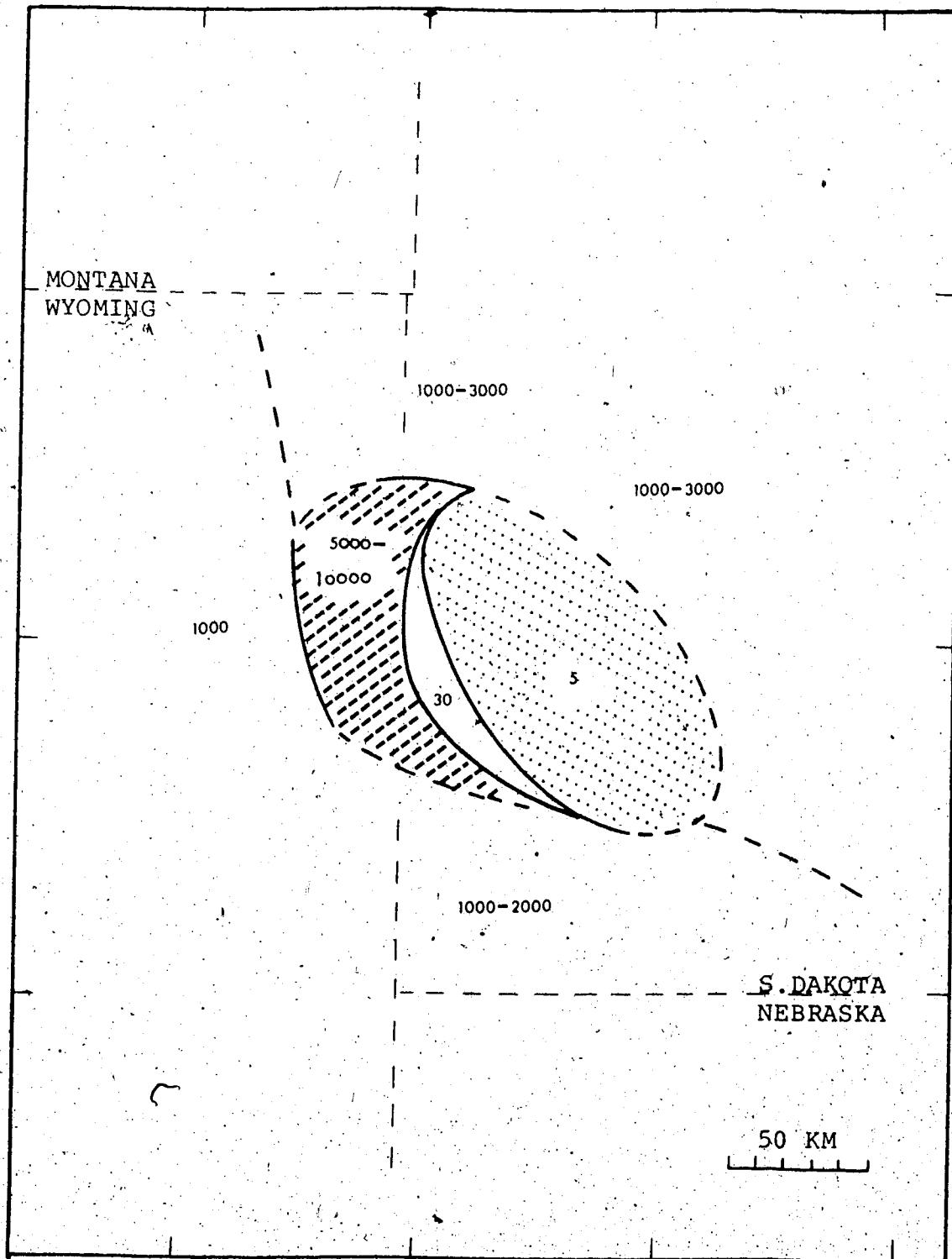


Fig.(5-11.2). Resistivity ( $\Omega\text{-M}$ ) distribution in a horizontal plane at a depth of 5-KM from the surface.

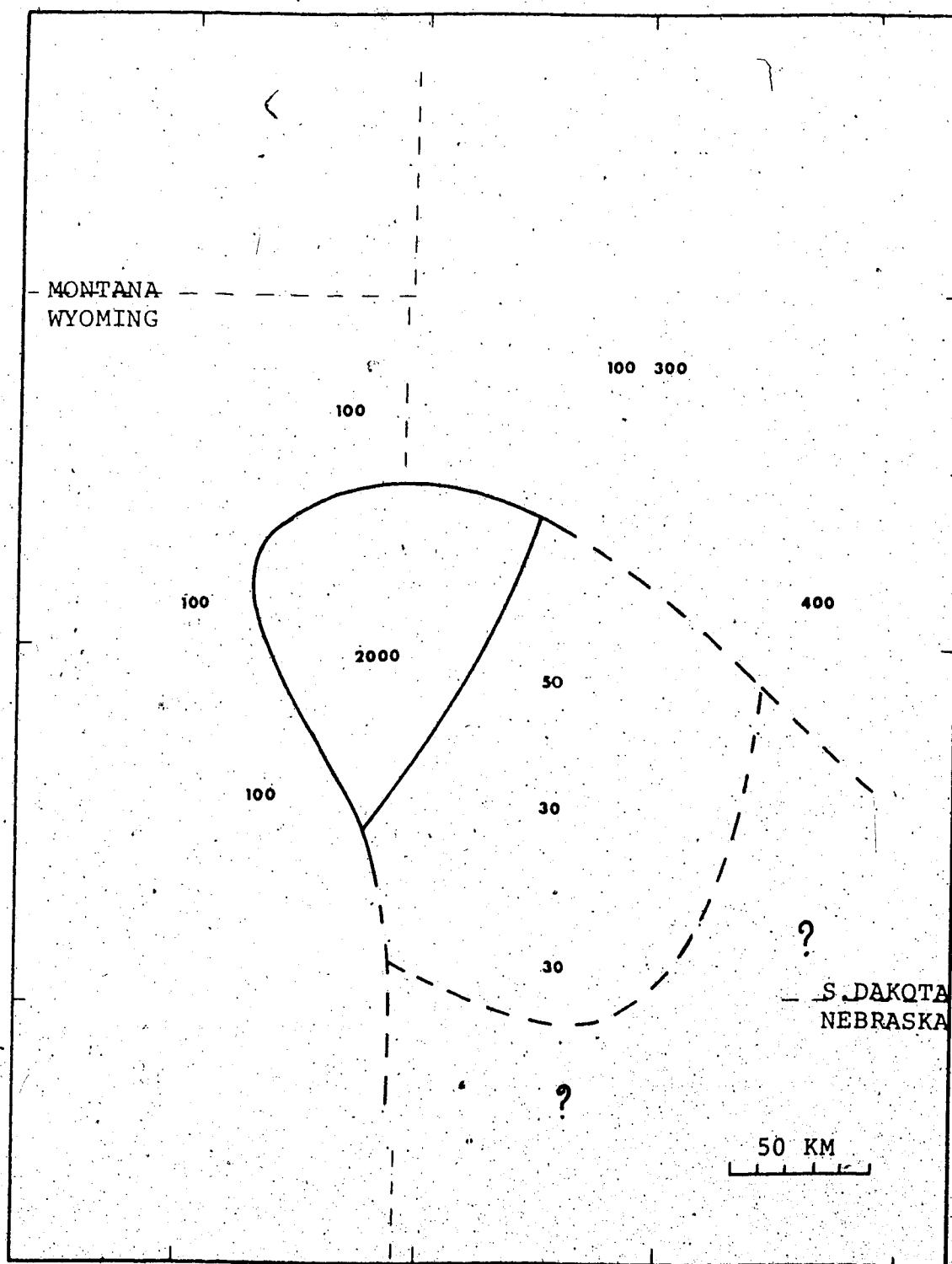


Fig. (5-11.3) Resistivity ( $\Omega\text{-M}$ ) distribution in a horizontal plane at a depth of 20-KM from the surface.

thinner layer of low resistivity, a second layer with high resistivity and the closed low resistivity anomaly previously reported by Rankin and Reddy (1973), which is now interpreted to lie at a relatively shallow depth. While the low resistivity of the surface material is expected and usually attributed to the moisture and salt content of the recent sediments, the anomaly in the middle of highly resistive crystalline rocks is more difficult to explain. Since highly conducting material shields the deeper structure, the values assigned to the region below the anomaly are somewhat tentative but it appears that the low resistivity persists downward throughout the entire column of the crust. It is tempting to interpret this low resistive anomaly in terms of a thermal anomaly although other suggestions have been put forward. One other such suggestion by Camfield, Gough and Porath (1971) attributes their elongated induction anomaly to the existence of graphite schists in the basement. The possibility that the anomaly is produced by a distribution of material with an abnormally high value of  $\mu$  cannot be overlooked and it is proposed to compute models embodying this concept.

The extremely low values of apparent resistivity shown at stations 6, 7 and 8 are associated with peaks in the power density spectrum of the H field at

approximately 6.25 Hz. This is near the Schumann resonance peak reported at 8.6 Hz. This latter phenomenon is associated with waves propagating within the earth-ionosphere cavity due to lightning strokes. If the peak described here is of the same nature as the Schumann resonance, the plane wave theory is not applicable and the results in this region of the spectrum may need modification. Work on this aspect of the problem is to be carried out.

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## APPENDIX A

### MODELLING CONSIDERATION

Numerical modelling results have been published for numerous one and two dimension cases which have already been referred to. In this appendix several examples are presented schematically and discussed.

#### I. One-Dimensional Model

The typical feature of the one-dimensional solutions for a layered earth is the simple frequency dependence which uniquely reveals information on conductivity and thickness of the layers through the apparent resistivity and phase relations. In Fig. (A-1a), a pair of two-layer models and their apparent resistivity curves are shown. For sufficiently high frequencies, the corresponding skin depth is sufficiently small, so that the lower layer is not sensed and the apparent resistivity is asymptotic to  $\rho_1$ . For sufficiently low frequencies, the corresponding skin depth is large and the upper layer has little effect, so that the apparent resistivity approaches  $\rho_2$ . The variation with frequency is smooth. A greater thickness of the upper layer, as in model-B compared to model-A, causes the apparent resistivity curve to be

displaced as shown. It is clear that the transition from  $\rho_1$  to  $\rho_2$  is a function of the thickness of the upper layer.

Fig. (A-1b) shows the result of a highly conductive and a highly resistive intervening layer. In both cases the apparent resistivity curves approach the same high and low frequency limits, while for intermediate frequencies the effect of the intervening layer is seen. The apparent resistivity never reaches the true values of  $\rho_2$  or  $\bar{\rho}_2$  as long as layer 2 is not very thick, but the effect of this layer will always be present as long as it is not extremely thin. The extension to a multi-layer model can be easily visualized.

## II. Two-Dimensional Model

Fig. (A-2) shows a solution for a vertical fault. Some notable features are as follows:

- a) The apparent resistivity is asymptotic to the appropriate resistivity value at large distances from the fault. It varies smoothly with position along a traverse crossing the fault in the case of  $E_{||}$  (E-field parallel to the strike), but varies discontinuously for the  $H_{||}$  case (E-field perpendicular to the strike), as shown in Fig. (A-2).

b) The vertical component of the magnetic field  $H_z$  exists only for  $E_{||}$ , with the maximum value appearing near the fault, decreasing to zero in both directions. The horizontal H-field perpendicular to the strike  $H_y$  also varies significantly near the fault for the  $E_{||}$  case. Most of these effects can be understood by examining the current flow as shown in the figure. For  $H_{||}$  and  $E_{\perp}$  the current flows across the interface and since the current is continuous, no  $H_z$  is induced. Since  $\sigma_1 \neq \sigma_2$  and  $E_y = \sigma J_y$ ,  $E_y$  is discontinuous across the fault with the condition  $E_1 \rho_1 = E_2 \rho_2$ . The apparent resistivity  $\rho_1$ , which depends on  $E_y^2$ , changes abruptly across the fault and approaches the appropriate value at large distances on either side of the interface.

Fig. (A-3) shows a pseudo-section for a fault. The horizontal scale is the distance along a traverse as used in Fig. (A-2), and the vertical scale is frequency (corresponding to depth) with the highest frequency at the top. The related data values are plotted beneath each site location for each frequency, and the values are contoured.

Fig. (A-4) shows the effect of an overburden above the fault. The result of the overburden is to smooth and attenuate the anomalous effects as shown; the degree of smoothing increases with the thickness

and conductivity of the overburden.

Fig. (A-5) shows a conductive thin dike buried in the vertical position, and Fig. (A-6) shows a resistive dike. 'Thin' describes the condition where the dike is small compared to a wave length at a given frequency. The effect can also be understood by examining the current flow situation. For a conductive dike, the thin sheet-like conductive material immersed in a resistive basement does not affect the currents flowing perpendicular to the sheet; but it does play an essential role when the currents flow parallel to the strike. The resulting anomalous patterns for  $\rho_{||}$  and  $H_z$  is shown in Fig. (A-5). For the resistive dike, the situation is the converse; currents which flow across the dike are affected. The  $\rho_{\perp}$  patterns are shown in Fig. (A-6).

Fig. (A-7) shows a two-dimensional model of a sloping contrast and the anomalous effects at a typical frequency. Fig. (A-8) shows the anomalous effects for various frequencies at one location; the values are plotted as a function of frequency.

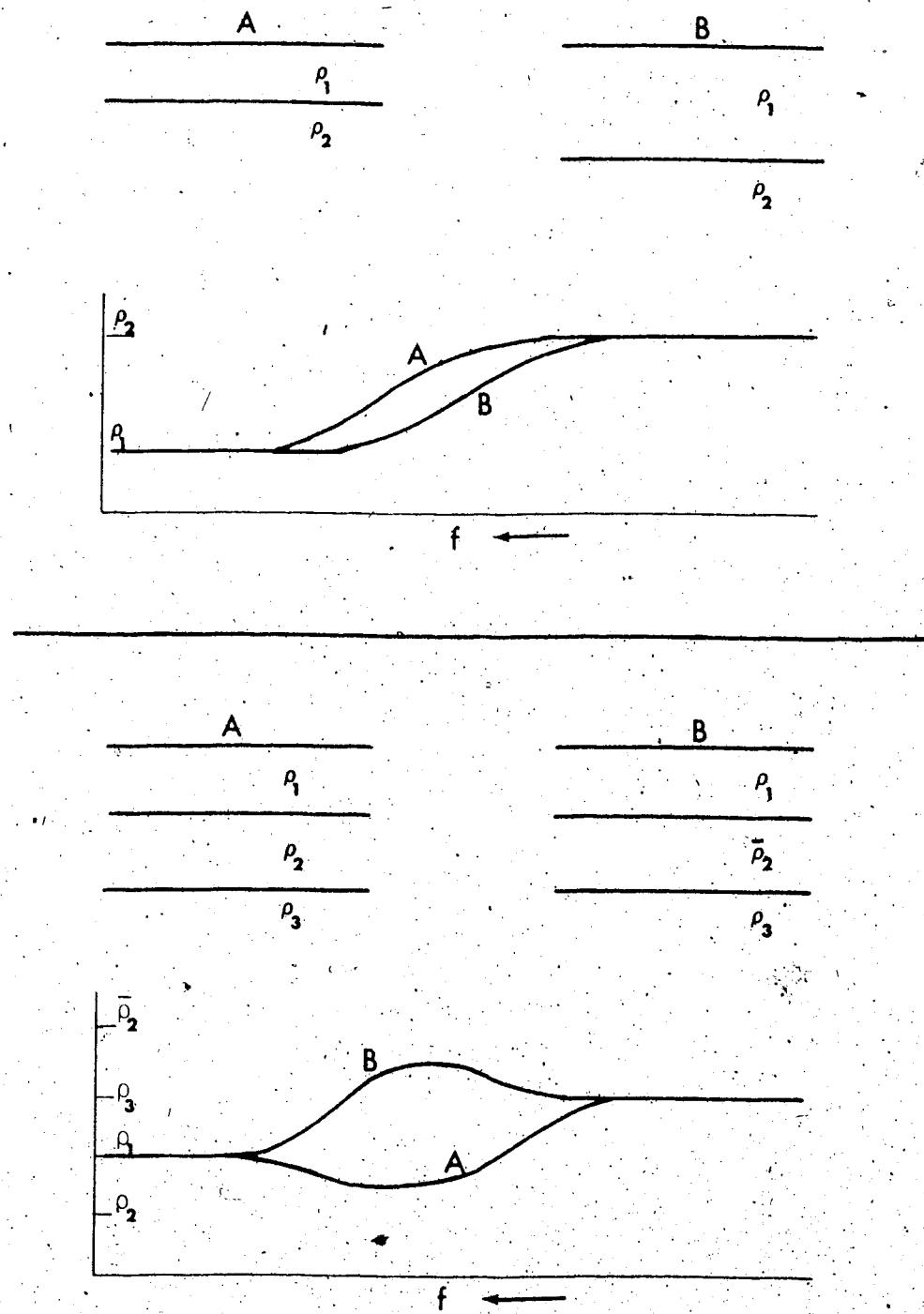


Fig. (A-1). a) Schematic response curves for two-layer models (upper), b) three-layer models (lower).

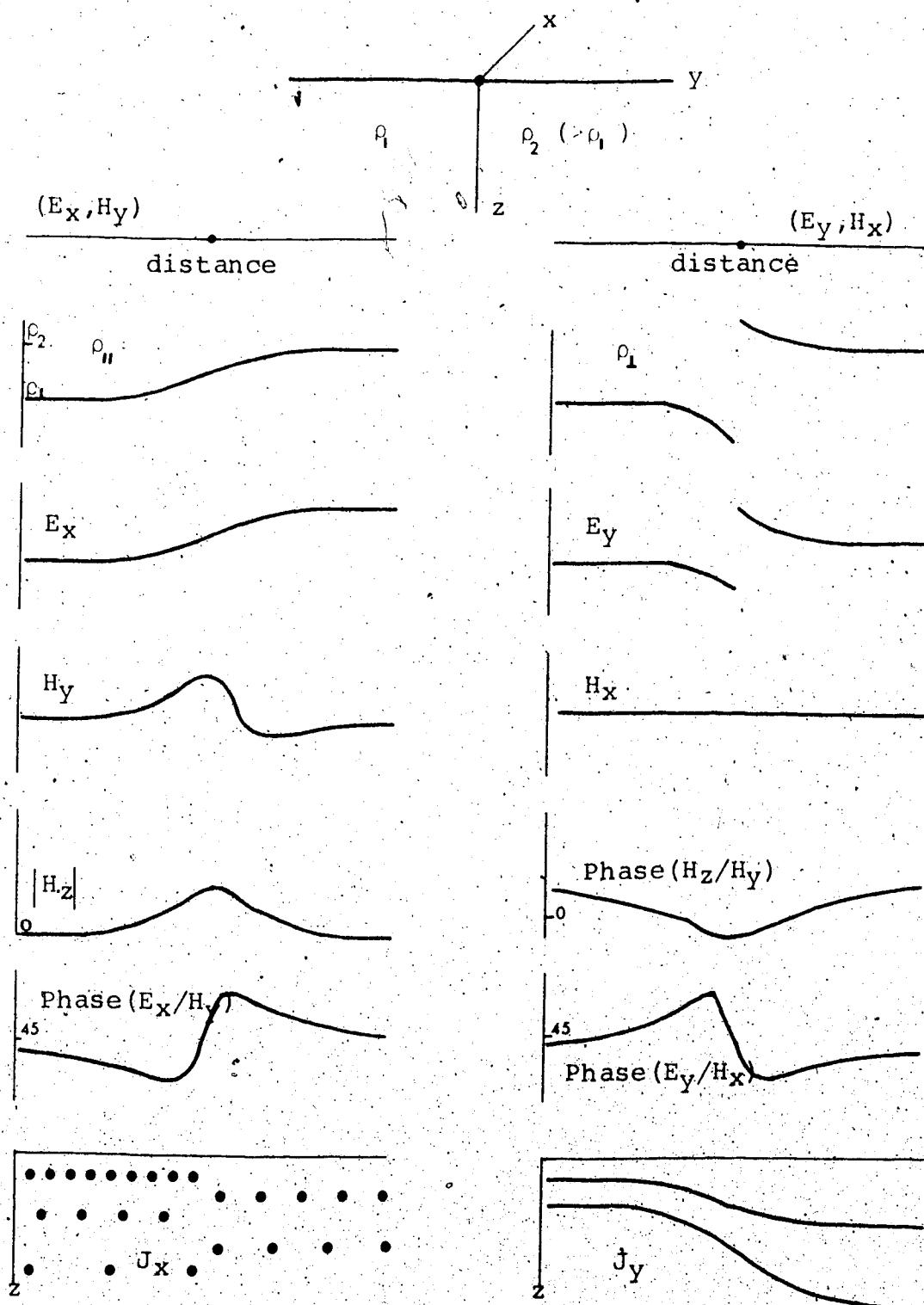


Fig. (A-2). Schematic pseudo-sections for a vertical fault.

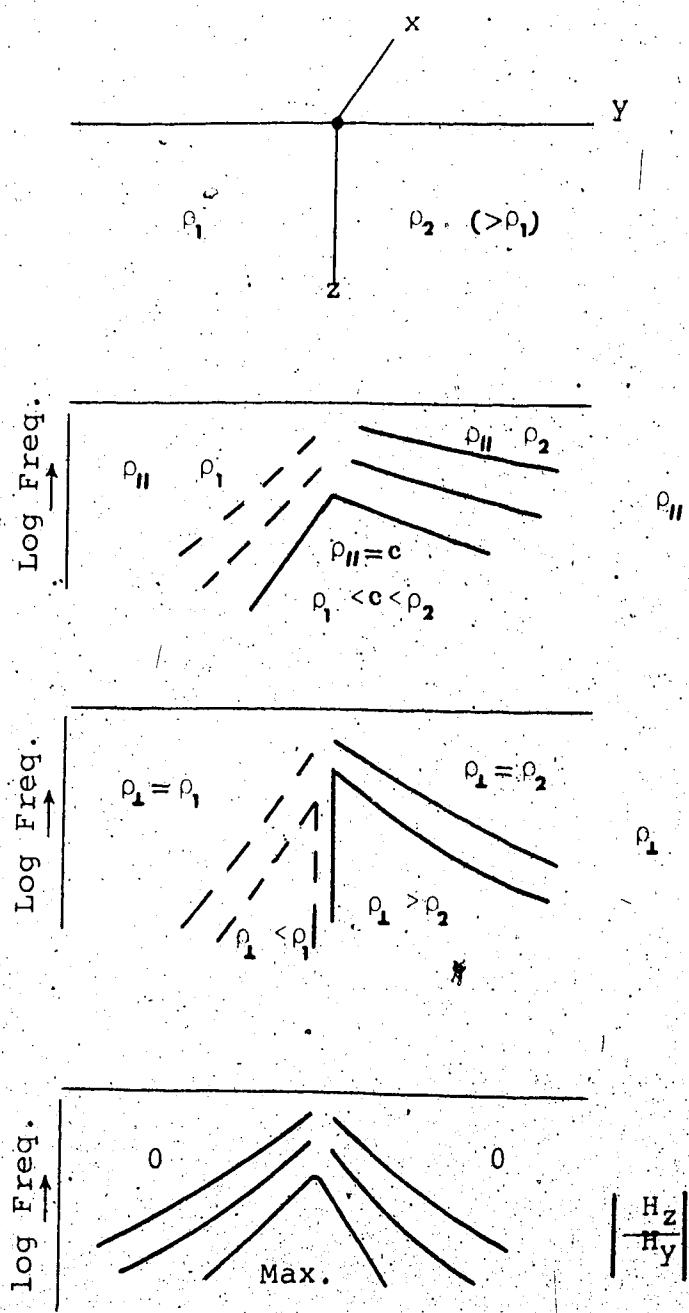
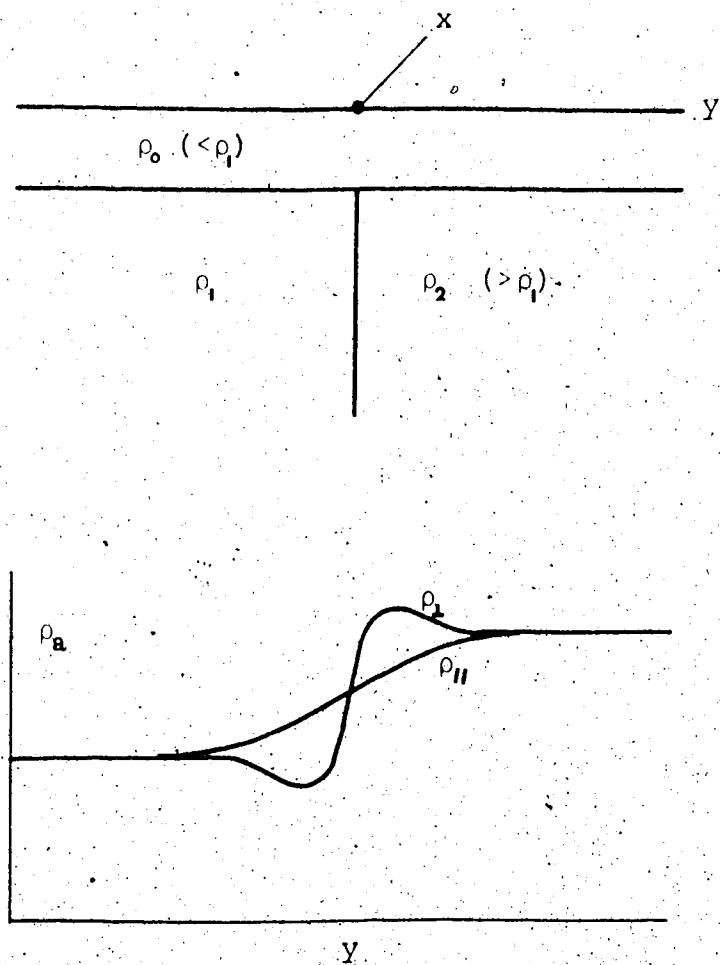


Fig. (A-3). Schematic pseudo-sections for a vertical fault.



Fig(A-4) Effect of overburden on a vertical fault.

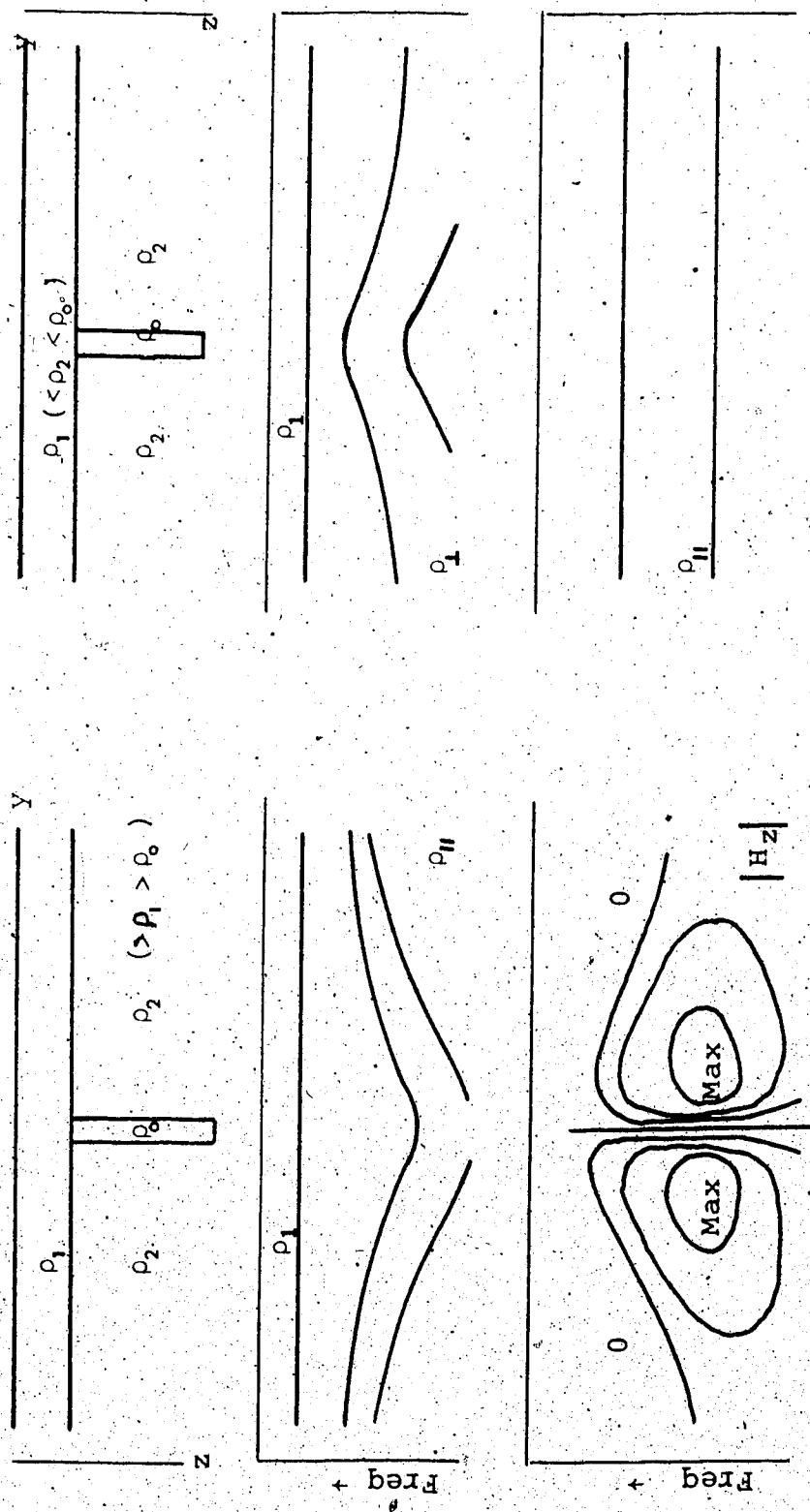


Fig. (A-5). Schematic pseudo-section for a conductive dike.  
Fig. (A-6). Schematic pseudo-section for a buried resistive dike.

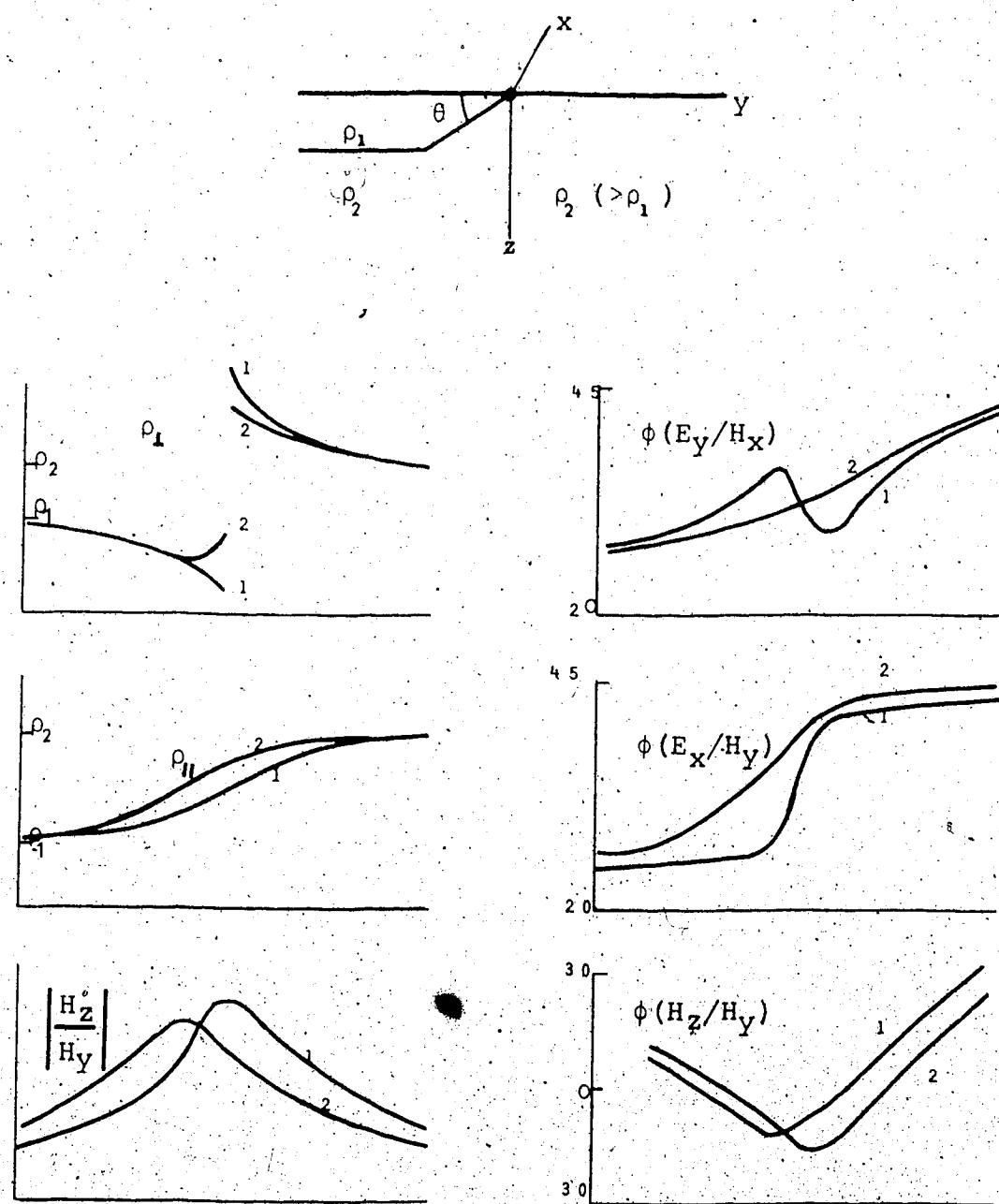


Fig. (A-7). Schematic pseudo-sections for a sloping contact.  $1 \rightarrow \theta = 90^\circ$ ,  $2 \rightarrow \theta = 45^\circ$ .

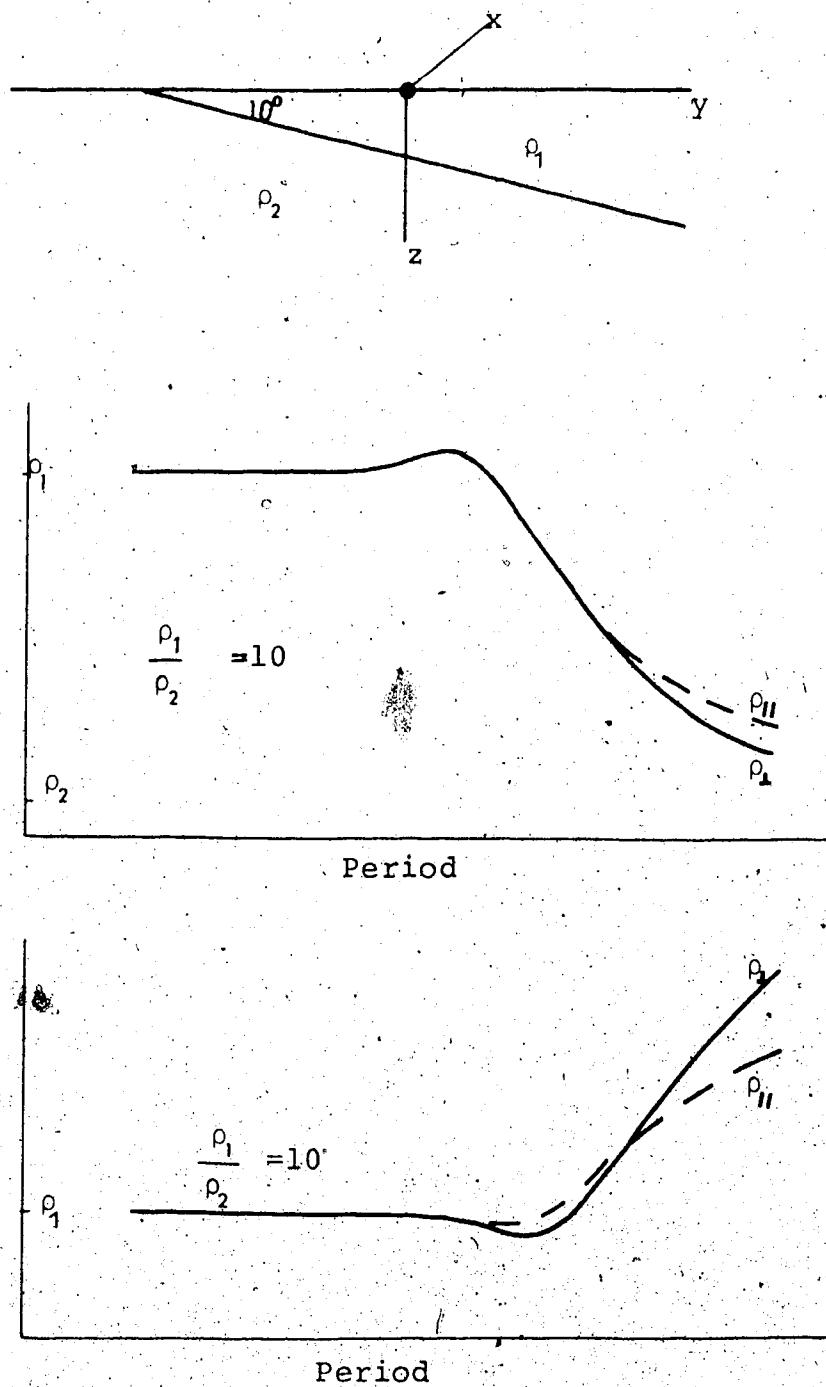


Fig.(A-8). Schematic response curves for a sloping contact.

## APPENDIX B

### ANALYSIS RESULTS

P1 --- (sec).

R1 --- Minimum apparent resistivity ( $\Omega\text{-M}$ ) for  $\theta=\theta_0$ .

PH1 --- Phase angle of R12.

R21 --- Minimum apparent resistivity ( $\Omega\text{-M}$ ) for  $\theta=\theta_0$ .

PH21 --- Phase angle of R21.

RE1 --- Maximum admittance for  $\theta=\theta_{ze}$ .

RH1 --- Maximum value of  $\alpha$  for  $\theta=\theta_{zh}$ .

ANG ---  $\theta_0$ , from the geographical North.

ANGE ---  $\theta_{ze}$ .

ANGH ---  $\theta_{zh}$ .

SKW --- skew factor.

BTA --- (Eqn 2-54b).

RAN ---  $\beta_1$  (Eqn 2-55a).

AIS ---  $\beta_2$  (Eqn 2-55b).

PRD	B12	PH12	R21	PH21	PE1	SH1	ANG	ANSH	SKW	BIA	RAN	AIS	
5.76	23.47	-102.	9.58	39.	12.67	1.53	105.	150.	110.	0.6	0.43	0.41	
5.76	20.41	-126.	11.76	38.	13.02	2.11	100.	165.	135.	0.77	0.02	0.58	
5.87	39.03	-128.	23.45	137.	13.78	1.92	60.	165.	135.	0.00	0.25	0.35	
6.13	66.62	-140.	0.75	79.	162.	1.16	60.	165.	140.	0.43	0.11	0.02	
6.41	77.53	-124.	0.39	179.	9.84	2.16	60.	175.	175.	0.43	0.26	0.48	
6.41	62.57	-125.	0.26	-121.	11.03	1.44	70.	170.	155.	0.26	0.14	0.04	
6.43	72.62	-125.	0.44	102.	19.85	1.87	65.	170.	65.	0.27	0.18	0.05	
6.43	68.82	-131.	0.38	-111.	16.23	1.77	70.	165.	35.	0.40	0.25	0.11	
6.43	82.21	-144.	0.37	-109.	8.97	1.25	70.	165.	35.	0.44	0.32	0.18	
6.43	91.36	-145.	0.46	84.45	14.35	0.37	65.	165.	45.	0.39	0.29	0.11	
6.43	33.80	-143.	0.60	78.28	14.35	5.32	60.	165.	60.	0.26	0.14	0.00	
6.43	75.71	-145.	0.64	34.87	14.35	3.57	65.	165.	75.	0.38	0.21	0.06	
6.43	71.43	-147.	0.59	25.51	154.	2.19	65.	160.	110.	0.38	0.17	0.08	
6.43	61.24	-153.	0.93	53.87	154.	1.69	65.	150.	125.	0.39	0.13	0.07	
6.43	33.27	-154.	0.91	118.	1.62	0.49	65.	160.	110.	0.47	0.15	0.02	
6.43	22.32	-156.	0.98	48.26	156.	0.98	65.	165.	90.	0.47	0.15	0.02	
6.43	3.61	-156.	1.01	40.04	156.	1.23	65.	165.	95.	0.49	0.09	0.02	
6.43	6.70	-154.	0.93	126.	1.23	0.74	65.	155.	90.	0.51	0.09	0.03	
6.43	1.21	-154.	1.03	1.03	1.29	0.49	65.	155.	100.	0.55	0.12	0.03	
6.43	4.85	-154.	1.24	1.26	1.26	0.32	65.	150.	105.	0.55	0.16	0.04	
6.43	33.38	-148.	1.34	1.34	1.29	0.50	65.	150.	100.	0.59	0.18	0.05	
6.43	21.55	-148.	1.34	1.34	1.29	0.56	65.	150.	100.	0.59	0.18	0.05	
6.43	2.36	-148.	1.34	1.34	1.29	0.50	65.	150.	100.	0.59	0.18	0.05	
6.43	21.80	-139.	1.70	1.70	1.17	0.23	65.	155.	65.	0.66	0.26	0.17	
6.43	24.77	-136.	2.32	2.32	1.12	0.11	70.	135.	75.	0.19	0.10	0.08	
6.43	27.28	-133.	2.46	2.46	0.88	0.14	70.	170.	50.	0.62	0.28	0.15	
6.43	28.20	-133.	4.32	4.32	1.17	0.72	75.	165.	35.	0.62	0.28	0.16	
6.43	29.00	-137.	0.60	1.17.	0.18	0.83	75.	175.	30.	0.53	0.31	0.13	
6.43	7.07	-142.	0.60	1.22.	0.23	0.84	65.	175.	10.	0.44	0.17	0.08	
6.43	14.49	-135.	0.52	-130.	2.33	0.32	65.	175.	10.	0.44	0.17	0.03	
6.43	14.15	-131.	0.45	-125.	2.32	1.00	65.	175.	5.	0.46	0.19	0.09	
6.43	31.40	-128.	0.74	126.	2.16	1.23	65.	170.	170.	0.40	0.20	0.03	
6.43	12.12	-128.	0.76	-128.	1.96	2.49	2.03	170.	5.	0.34	0.42	0.15	
6.43	18.78	-138.	0.62	-132.	1.73	2.96	80.	150.	10.	0.26	0.54	0.21	
6.43	59.52	-138.	2.01	14.4.	1.57	2.56	85.	155.	5.	0.50	0.25	0.11	
6.43	17.00	-138.	0.55	-131.	1.03	1.70	70.	175.	60.	0.56	0.37	0.19	
6.43	1.18	-131.	0.45	-125.	2.32	1.00	65.	175.	5.	0.46	0.19	0.09	
6.43	32.19	-118.	0.43	109.	1.61	1.61	80.	145.	170.	0.62	0.17	0.03	
6.43	27.94	-108.	0.33	-118.	1.52	1.00	80.	150.	5.	0.48	0.17	0.10	
6.43	-10.10	-108.	0.48	-113.	0.93	3.83	80.	150.	30.	0.58	0.16	0.01	
6.43	35.95	-118.	0.66	0.66	0.76	0.37	80.	165.	165.	0.52	0.16	0.03	
6.43	49.28	-122.	0.52	1.16	2.02	0.16	1.25	155.	0.	0.51	0.21	0.12	
6.43	45.55	-118.	1.05	-122.	0.26	0.86	80.	165.	155.	0.58	0.19	0.15	
6.43	63.47	-119.	3.21	3.18	1.19	1.71	80.	165.	120.	0.53	0.13	0.11	
6.43	66.75	-113.	2.62	-123.	0.28	0.67	85.	170.	90.	0.54	0.07	0.05	
6.43	53.68	-135.	2.66	-136.	0.16	1.00	85.	160.	0.	0.57	0.08	0.11	
6.43	42.82	-122.	1.16	1.16	0.37	0.56	85.	155.	35.	0.56	0.16	0.03	
6.43	82.41	-151.	1.72	1.72	1.17	0.03	0.62	85.	140.	150.	0.57	0.06	0.11
6.43	26.79	-161.	1.71	1.71	1.51.	0.08	0.77	85.	165.	25.	0.52	0.10	0.11
6.43	64.55	-174.	1.60	1.60	1.53.	0.04	0.49	85.	150.	135.	0.50	0.06	0.02
6.43	52.85	-178.	2.09	-160.	0.02	0.71	85.	160.	30.	0.50	0.08	0.02	
6.43	38.18	-167.	1.89	-167.	0.05	0.46	85.	165.	35.	0.49	0.08	0.04	
6.43	22.62	-156.	1.47	-180.	0.03	0.27	85.	165.	50.	0.50	0.10	0.05	
6.43	-0.08	152.	1.07	-180.	0.03	0.27	85.	160.	110.	0.55	0.05	0.07	
6.43	7.71	141.	0.66	-160.	0.02	0.27	85.	160.	155.	0.55	0.08	0.05	

-52-

FILE





PRD	R12	RH12	Rc	Rd	R21	R31	R41	R51	R61	R71	R81	R91	R101	R111	R121	R131	R141	R151	R161	R171	R181	R191	R201	R211	R221	R231	R241	R251	R261	R271	R281	R291	R301	R311	R321	R331	R341	R351	R361	R371	R381	R391	R401	R411	R421	R431	R441	R451	R461	R471	R481	R491	R501	R511	R521	R531	R541	R551	R561	R571	R581	R591	R601	R611	R621	R631	R641	R651	R661	R671	R681	R691	R701	R711	R721	R731	R741	R751	R761	R771	R781	R791	R801	R811	R821	R831	R841	R851	R861	R871	R881	R891	R901	R911	R921	R931	R941	R951	R961	R971	R981	R991	R1001	R1011	R1021	R1031	R1041	R1051	R1061	R1071	R1081	R1091	R1101	R1111	R1121	R1131	R1141	R1151	R1161	R1171	R1181	R1191	R1201	R1211	R1221	R1231	R1241	R1251	R1261	R1271	R1281	R1291	R1301	R1311	R1321	R1331	R1341	R1351	R1361	R1371	R1381	R1391	R1401	R1411	R1421	R1431	R1441	R1451	R1461	R1471	R1481	R1491	R1501	R1511	R1521	R1531	R1541	R1551	R1561	R1571	R1581	R1591	R1601	R1611	R1621	R1631	R1641	R1651	R1661	R1671	R1681	R1691	R1701	R1711	R1721	R1731	R1741	R1751	R1761	R1771	R1781	R1791	R1801	R1811	R1821	R1831	R1841	R1851	R1861	R1871	R1881	R1891	R1901	R1911	R1921	R1931	R1941	R1951	R1961	R1971	R1981	R1991	R2001	R2011	R2021	R2031	R2041	R2051	R2061	R2071	R2081	R2091	R2101	R2111	R2121	R2131	R2141	R2151	R2161	R2171	R2181	R2191	R2201	R2211	R2221	R2231	R2241	R2251	R2261	R2271	R2281	R2291	R2301	R2311	R2321	R2331	R2341	R2351	R2361	R2371	R2381	R2391	R2401	R2411	R2421	R2431	R2441	R2451	R2461	R2471	R2481	R2491	R2501	R2511	R2521	R2531	R2541	R2551	R2561	R2571	R2581	R2591	R2601	R2611	R2621	R2631	R2641	R2651	R2661	R2671	R2681	R2691	R2701	R2711	R2721	R2731	R2741	R2751	R2761	R2771	R2781	R2791	R2801	R2811	R2821	R2831	R2841	R2851	R2861	R2871	R2881	R2891	R2901	R2911	R2921	R2931	R2941	R2951	R2961	R2971	R2981	R2991	R3001	R3011	R3021	R3031	R3041	R3051	R3061	R3071	R3081	R3091	R3101	R3111	R3121	R3131	R3141	R3151	R3161	R3171	R3181	R3191	R3201	R3211	R3221	R3231	R3241	R3251	R3261	R3271	R3281	R3291	R3301	R3311	R3321	R3331	R3341	R3351	R3361	R3371	R3381	R3391	R3401	R3411	R3421	R3431	R3441	R3451	R3461	R3471	R3481	R3491	R3501	R3511	R3521	R3531	R3541	R3551	R3561	R3571	R3581	R3591	R3601	R3611	R3621	R3631	R3641	R3651	R3661	R3671	R3681	R3691	R3701	R3711	R3721	R3731	R3741	R3751	R3761	R3771	R3781	R3791	R3801	R3811	R3821	R3831	R3841	R3851	R3861	R3871	R3881	R3891	R3901	R3911	R3921	R3931	R3941	R3951	R3961	R3971	R3981	R3991	R4001	R4011	R4021	R4031	R4041	R4051	R4061	R4071	R4081	R4091	R4101	R4111	R4121	R4131	R4141	R4151	R4161	R4171	R4181	R4191	R4201	R4211	R4221	R4231	R4241	R4251	R4261	R4271	R4281	R4291	R4301	R4311	R4321	R4331	R4341	R4351	R4361	R4371	R4381	R4391	R4401	R4411	R4421	R4431	R4441	R4451	R4461	R4471	R4481	R4491	R4501	R4511	R4521	R4531	R4541	R4551	R4561	R4571	R4581	R4591	R4601	R4611	R4621	R4631	R4641	R4651	R4661	R4671	R4681	R4691	R4701	R4711	R4721	R4731	R4741	R4751	R4761	R4771	R4781	R4791	R4801	R4811	R4821	R4831	R4841	R4851	R4861	R4871	R4881	R4891	R4901	R4911	R4921	R4931	R4941	R4951	R4961	R4971	R4981	R4991	R5001	R5011	R5021	R5031	R5041	R5051	R5061	R5071	R5081	R5091	R5010	R5011	R5012	R5013	R5014	R5015	R5016	R5017	R5018	R5019	R5020	R5021	R5022	R5023	R5024	R5025	R5026	R5027	R5028	R5029	R5030	R5031	R5032	R5033	R5034	R5035	R5036	R5037	R5038	R5039	R5040	R5041	R5042	R5043	R5044	R5045	R5046	R5047	R5048	R5049	R5050	R5051	R5052	R5053	R5054	R5055	R5056	R5057	R5058	R5059	R5060	R5061	R5062	R5063	R5064	R5065	R5066	R5067	R5068	R5069	R5070	R5071	R5072	R5073	R5074	R5075	R5076	R5077	R5078	R5079	R5080	R5081	R5082	R5083	R5084	R5085	R5086	R5087	R5088	R5089	R5090	R5091	R5092	R5093	R5094	R5095	R5096	R5097	R5098	R5099	R50100	R50101	R50102	R50103	R50104	R50105	R50106	R50107	R50108	R50109	R50110	R50111	R50112	R50113	R50114	R50115	R50116	R50117	R50118	R50119	R50120	R50121	R50122	R50123	R50124	R50125	R50126	R50127	R50128	R50129	R50130	R50131	R50132	R50133	R50134	R50135	R50136	R50137	R50138	R50139	R50140	R50141	R50142	R50143	R50144	R50145	R50146	R50147	R50148	R50149	R50150	R50151	R50152	R50153	R50154	R50155	R50156	R50157	R50158	R50159	R50160	R50161	R50162	R50163	R50164	R50165	R50166	R50167	R50168	R50169	R50170	R50171	R50172	R50173	R50174	R50175	R50176	R50177	R50178	R50179	R50180	R50181	R50182	R50183	R50184	R50185	R50186	R50187	R50188	R50189	R50190	R50191	R50192	R50193	R50194	R50195	R50196	R50197	R50198	R50199	R50200	R50201	R50202	R50203	R50204	R50205	R50206	R50207	R50208	R50209	R50210	R50211	R50212	R50213	R50214	R50215	R50216	R50217	R50218	R50219	R50220	R50221	R50222	R50223	R50224	R50225	R50226	R50227	R50228	R50229	R50230	R50231	R50232	R50233	R50234	R50235	R50236	R50237	R50238	R50239	R50240	R50241	R50242	R50243	R50244	R50245	R50246	R50247	R50248	R50249	R50250	R50251	R50252	R50253	R50254	R50255	R50256	R50257	R50258	R50259	R50260	R50261	R50262	R50263	R50264	R50265	R50266	R50267	R50268	R50269	R50270	R50271	R50272	R50273	R50274	R50275	R50276	R50277	R50278	R50279	R50280	R50281	R50282	R50283	R50284	R50285	R50286	R50287	R50288	R50289	R50290	R50291	R50292	R50293	R50294	R50295	R50296	R50297	R50298	R50299	R50300	R50301	R50302	R50303	R50304	R50305	R50306	R50307	R50308	R50309	R50310	R50311	R50312	R50313	R50314	R50315	R50316	R50317	R50318	R50319	R50320	R50321	R50322	R50323	R50324	R50325	R50326	R50327	R50328	R50329	R50330	R50331	R50332	R50333	R50334	R50335	R50336	R50337	R50338	R50339	R50340	R50341	R50342	R50343	R50344	R50345	R50346	R50347	R50348	R50349	R50350	R50351	R50352	R50353	R50354	R50355	R50356	R50357	R50358	R50359	R50360	R50361	R50362	R50363	R50364	R50365	R50366	R50367	R50368	R50369	R50370	R50371	R50372	R50373	R50374	R50375	R50376	R50377	R50378	R50379	R50380	R50381	R50382	R50383	R50384	R50385	R50386	R50387	R50388	R50389	R50390	R50391	R50392	R50393	R50394	R50395	R50396	R50397	R50398	R50399	R50400	R50401	R50402	R50403	R50404	R50405	R50406	R50407	R50408	R50409	R50410	R50411	R50412	R50413	R50414	R50415	R50416	R50417	R50418	R50419	R50420	R50421	R50422	R50423	R50424	R50425	R50426	R50427	R50428	R50429	R50430	R50431	R50432	R50433	R50434	R50435	R50436	R50437	R50438	R50439	R50440	R50441	R50442	R50443	R50444	R50445	R50446	R50447	R50448	R50449	R50450	R50451	R50452	R50453	R50454	R50455	R50456	R50457	R50458	R50459	R50460	R50461	R50462	R50463	R50464	R50465	R50466	R50467	R50468	R50469	R50470	R50471	R50472	R50473	R50474	R50475	R50476	R50477	R50478	R50479	R50480	R50481	R50482	R50483	R50484	R50485	R50486	R50487	R50488	R50489	R50490	R50491	R50492	R50493	R50494	R50495	R50496	R50497	R50498	R50499	R50500	R50501	R50502	R50503	R50504	R50505	R50506	R50507	R50508	R50509	R50510	R50511	R50512	R50513	R50514	R50515	R50516	R50517	R50518	R50519	R50520	R50521	R50522	R50523	R50524	R50525	R50526	R50527	R50528	R50529	R50530	R50531	R50532	R50533	R50534	R50535	R50536	R50537	R50538	R50539	R50540	R50541	R50542	R50543	R50544	R50545	R50546	R50547	R50548	R50549	R50550	R50551	R50552	R50553	R50554	R50555	R50556	R50557	R50558	R50559	R50560	R50561	R50562	R50563	R50564	R50565	R50566	R50567	R50568	R50569	R50570	R50571	R50572	R50573	R50574	R50575	R50576	R50577	R50578	R50579	R50580	R50581	R50582	R50583	R50584	R50585	R50586	R50587	R50588	R50589	R50590	R50591	R50592	R50593	R50594	R50595	R50596	R50597	R50598	R50599	R50600	R50601	R50602	R50603	R50604	R50605	R50606	R50607	R50608	R50609	R50610	R50611	R50612	R50613	R50614	R50615	R50616	R50617	R50618	R50619	R50620	R50621	R50622	R50623	R50624	R50625	R50626	R50627	R50628	R50629	R50630	R50631	R50632	R50633	R50634	R50635	R50636	R50637	R50638	R50639	R50640	R50641	R50642	R50643	R50644	R50645	R50646	R50647	R50648	R50649	R50650	R50651	R50652	R50653	R50654	R50655	R50656	R50657	R50658	R50659	R50660	R50661	R50662	R50663	R50664	R50665	R50666	R50667	R50668	R50669	R50670	R50671	R50672	R50673	R50674	R50675	R50676	R50677	R50678	R50679	R50680	R50681	R50682	R50683	R50684	R50685	R50686	R50687	R50688	R50689	R50690	R50691	R50692	R50693	R50694	R50695	R50696	R50697	R50698	R50699	R50700	R50



PRD	R12	PH12	PH21	PE1	PH1	AMG	ANGZ	ALGH	SKW	BTA	RAN	AIS
1337.75	193.97	44.	41.18	76.	1.47	0.58	165.	50.	.75.	0.53	0.20	0.08
875.76	199.13	37.	40.75	73.	0.99	0.64	165.	40.	90.	0.29	0.16	0.20
668.87	211.82	33.	47.32	69.	0.76	0.64	165.	20.	85.	0.27	0.08	0.22
546.13	224.03	31.	69.34	73.	0.67	0.60	165.	10.	95.	0.04	0.08	0.31
463.41	229.70	31.	77.73	71.	0.62	0.62	165.	175.	95.	0.05	0.08	0.28
357.53	223.90	27.	90.46	65.	0.64	0.67	165.	10.	75.	0.25	0.01	0.40
291.92	251.38	28.	143.44	66.	0.52	0.72	170.	10.	90.	0.40	0.08	0.02
229.42	232.74	42.	188.57	58.	0.82	0.74	170.	5.	75.	0.98	0.07	0.03
189.19	309.98	52.	217.62	45.	0.42	1.01	170.	175.	90.	3.15	0.16	0.14
168.44	264.33	45.	179.27	48.	0.39	0.98	165.	170.	80.	2.81	0.10	0.07
133.80	210.03	37.	159.26	54.	0.39	0.95	165.	0.	70.	1.36	0.13	0.06
106.29	190.30	43.	165.35	56.	0.35	0.96	170.	0.	75.	1.94	0.15	0.08
88.43	202.09	40.	180.37	55.	0.30	0.94	170.	0.	75.	1.63	0.13	0.06
67.06	218.69	42.	174.32	55.	0.28	1.03	170.	0.	80.	1.74	0.09	0.06
53.27	226.53	46.	204.78	54.	0.26	1.18	170.	175.	85.	3.07	0.05	0.05
42.32	226.77	45.	201.91	54.	0.25	1.17	170.	0.	80.	3.09	0.06	0.06
33.61	232.66	44.	214.92	54.	0.21	1.17	170.	175.	80.	2.48	0.09	0.05
26.70	236.73	45.	236.54	53.	0.17	1.09	170.	175.	85.	3.93	0.09	0.08
21.21	246.80	127.	241.12	135.	0.15	1.05	170.	0.	80.	3.87	0.12	0.09
16.85	266.07	128.	233.61	137.	0.14	1.03	170.	0.	80.	3.60	0.14	0.10
13.18	282.95	130.	260.02	119.	0.14	1.24	170.	0.	80.	4.45	0.13	0.08
10.83	299.03	132.	266.08	140.	0.12	1.21	170.	0.	90.	4.46	0.16	0.13
8.44	300.25	135.	284.51	146.	0.05	0.87	85.	130.	105.	3.02	0.14	0.05
6.71	310.34	138.	276.36	148.	0.06	1.06	85.	135.	110.	4.61	0.14	0.09
5.33	328.03	143.	251.60	149.	0.06	1.22	85.	160.	105.	3.74	0.13	0.11
4.18	275.43	120.	248.25	131.	0.22	1.07	85.	10.	80.	3.57	0.19	0.17
2.73	315.68	119.	172.92	126.	0.22	1.06	85.	10.	80.	4.48	0.16	0.20
20.90	179.65	124.	150.42	130.	0.20	1.06	80.	10.	80.	5.10	0.12	0.13
17.07	162.49	121.	135.45	129.	0.18	1.09	80.	80.	5.	4.13	0.09	0.10
14.98	172.96	121.	181.64	129.	0.18	1.11	80.	80.	5.	2.65	0.11	0.11
11.77	133.65	49.	148.02	53.	0.13	1.04	170.	0.	90.	8.78	0.06	0.11
9.12	147.89	47.	137.89	47.	0.11	1.05	170.	170.	95.	45.67	0.07	0.16
7.17	111.06	33.	103.91	67.	0.07	0.75	170.	170.	95.	0.97	0.16	0.08
5.91	138.48	30.	98.36	52.	0.09	0.69	175.	35.	85.	0.86	0.26	0.13
5.26	165.18	30.	109.83	40.	0.10	1.00	170.	25.	55.	1.50	0.12	0.06
4.18	200.69	42.	126.76	143.	0.17	1.78	85.	55.	15.	2.41	0.15	0.12
3.32	152.61	48.	123.01	154.	0.09	1.63	85.	95.	165.	3.94	0.13	0.06
2.64	136.61	58.	135.59	157.	0.06	2.58	85.	170.	160.	3.62	0.15	0.13
2.10	237.01	152.	116.77	161.	0.07	1.63	90.	0.	130.	1.17	0.19	0.08
1.66	152.98	155.	111.47	167.	0.06	1.56	90.	0.	105.	2.47	0.23	0.14
1.32	152.48	164.	98.86	168.	0.05	0.90	90.	160.	135.	2.60	0.20	0.16
1.05	113.58	167.	79.84	171.	0.03	0.87	90.	45.	95.	1.91	0.18	0.11
0.83	104.64	174.	67.62	172.	0.04	0.94	90.	0.	100.	2.03	0.16	0.08
0.66	80.05	178.	47.97	176.	0.02	0.34	90.	170.	10.	1.75	0.14	0.07
0.53	57.67	178.	40.59	178.	0.02	0.59	90.	25.	65.	2.09	0.06	0.05
0.42	41.14	174.	26.51	177.	0.01	0.34	90.	5.	45.	1.54	0.14	0.05
0.33	22.81	177.	10.87	180.	0.01	0.11	90.	165.	100.	0.90	0.11	0.04
0.26	13.32	179.	7.71	178.	0.02	0.22	90.	80.	120.	1.27	0.06	0.03
0.21	7.05	165.	4.85	166.	0.03	0.16	95.	150.	120.	1.68	0.07	0.04
0.17	6.46	137.	4.80	135.	0.02	0.14	95.	175.	200.	2.37	0.09	0.04
0.13	16.29	124.	2.52	117.	0.01	0.18	95.	165.	150.	1.60	0.03	0.03
0.11	14.78	123.	11.32	118.	0.01	0.33	95.	50.	45.	1.63	0.04	0.03
0.08	21.69	126.	17.51	112.	0.00	0.23	95.	150.	145.	1.50	0.05	0.02
0.07	18.60	130.	17.00	116.	0.01	1.88	95.	150.	150.	1.60	0.07	0.07

PRD	812	2H12	321	PH21	PE1	PH1	ANS	ANT	CHAN	SKW	RAN	AIS
1337-75	436-21	-109.	39.55	110.	50.66	15.19	65.	175.	95.	0.93	0.70	0.18
875.76	1068.60	-26.	611.64	81.	24.99	7.51	162.	170.	75.	1.02	0.70	0.57
669.87	293.94	-21.	165.23	71.	24.32	11.32	172.	175.	80.	1.04	0.99	0.22
546.13	179.36	-120.	52.53	-108.	17.25	11.20	85.	86.	5.	0.73	0.11	0.35
463.41	180.99	-122.	44.19	-148.	14.46	6.45	45.	46.	90.	0.63	0.13	0.56
357.53	193.06	-125.	45.46	-145.	4.08	6.56	85.	85.	60.	0.75	0.19	0.24
291.92	182.06	-133.	47.66	-142.	6.74	4.74	85.	85.	65.	0.75	0.18	0.09
229.42	170.35	-137.	56.56	-144.	7.47	2.95	85.	85.	10.	0.46	0.08	0.26
189.19	168.52	-142.	47.19	-143.	5.86	6.35	85.	85.	100.	0.46	0.02	0.33
168.44	172.43	-146.	54.65	-142.	7.73	10.72	85.	85.	95.	0.38	0.04	0.26
133.80	180.49	-141.	51.41	-140.	2.76	2.59	85.	85.	70.	0.52	0.07	0.32
106.29	157.55	-139.	57.15	-138.	1.57	3.90	85.	85.	95.	0.64	0.09	0.28
84.83	158.52	-140.	64.74	-138.	0.31	0.43	85.	85.	55.	0.71	0.12	0.03
67.06	156.38	-140.	57.48	-138.	0.22	0.55	85.	85.	120.	0.49	0.11	0.03
53.27	160.90	-142.	60.25	-138.	0.19	0.68	85.	85.	140.	0.48	0.06	0.02
42.32	172.85	-143.	64.56	-138.	0.17	0.66	85.	85.	130.	0.62	0.06	0.37
33.61	156.96	-146.	62.76	-140.	0.16	0.48	85.	85.	25.	0.65	0.02	0.40
26.70	138.95	-148.	67.41	-141.	0.19	1.03	85.	85.	165.	0.85	0.05	0.49
21.21	161.91	-149.	66.11	-144.	0.21	1.24	85.	85.	160.	0.79	0.05	0.47
16.85	120.14	-151.	68.56	-144.	0.12	0.65	85.	85.	105.	1.14	0.06	0.35
13.38	122.74	-156.	57.25	-148.	0.26	0.88	85.	85.	100.	0.63	0.07	0.47
10.63	98.86	-160.	64.19	-151.	0.21	0.48	85.	85.	30.	1.28	0.02	0.50
8.44	67.06	-152.	39.76	-152.	0.27	1.62	85.	85.	125.	1.40	0.01	0.44
6.71	63.82	-169.	37.55	-159.	0.09	0.56	85.	85.	155.	1.18	0.04	0.59
5.33	59.92	-171.	33.92	-161.	0.15	1.13	85.	85.	125.	0.89	0.06	0.57
4.180	131.63	-152.	36.29	-139.	2.93	2.93	90.	90.	45.	1.28	0.09	0.28
27.37	100.34	-150.	23.92	-138.	1.11	2.32	90.	90.	30.	1.75	0.12	0.24
20.90	97.67	-153.	26.24	-147.	1.17	2.68	90.	90.	25.	1.25	0.46	0.27
17.07	98.58	-162.	36.67	-153.	1.33	2.75	85.	85.	15.	1.75	0.98	0.07
14.48	88.64	-166.	33.62	-155.	1.55	3.15	85.	85.	20.	1.75	1.22	0.39
11.17	99.75	-166.	25.31	-157.	0.49	3.37	85.	85.	175.	1.45	0.99	0.10
9.12	75.78	-165.	23.05	-155.	0.56	5.82	85.	85.	125.	1.31	0.06	0.30
7.17	60.86	-164.	17.87	-143.	0.62	5.00	80.	80.	145.	1.11	0.06	0.18
5.91	52.60	-168.	24.15	-154.	0.94	3.64	80.	80.	15.	1.35	1.29	0.09
5.26	50.86	-171.	6.79	-166.	0.48	2.36	85.	85.	165.	1.65	0.03	0.46
4.18	47.41	-169.	14.31	-172.	1.17	6.52	85.	85.	100.	1.20	0.05	0.53
3.32	38.05	-171.	13.44	-156.	0.66	3.54	85.	85.	15.	1.75	0.08	0.30
2.64	28.81	-174.	9.32	-169.	0.37	2.48	85.	85.	15.	1.41	0.09	0.35
2.10	22.77	-179.	9.57	-176.	0.50	3.21	85.	85.	145.	1.02	0.02	0.32
1.66	5.53	-174.	15.56	-168.	24.15	2.24	85.	85.	15.	1.35	0.05	0.42
1.32	51.75	-179.	7.09	-176.	1.59	1.76	85.	85.	165.	1.65	0.03	0.44
1.05	8.52	-172.	6.19	-179.	0.56	2.43	85.	85.	110.	1.00	0.04	0.60
0.83	5.92	-176.	4.75	-173.	0.20	1.15	85.	85.	180.	1.77	0.15	0.73
0.71	4.61	-170.	3.20	-173.	0.23	1.25	85.	85.	135.	2.49	0.12	0.09
0.66	4.63	-166.	0.51	-172.	0.14	0.61	85.	85.	140.	1.70	0.17	0.08
0.53	4.68	-172.	2.24	-178.	0.14	0.61	85.	85.	190.	1.22	0.09	0.08
0.42	2.34	-167.	1.53	-176.	0.15	0.46	85.	85.	15.	1.75	2.24	0.10
0.33	1.53	-167.	1.38	-176.	0.07	0.39	85.	85.	160.	2.53	0.02	0.90
0.26	1.08	-158.	0.76	-173.	0.11	0.10	85.	85.	90.	1.25	0.12	0.10
0.21	0.65	-138.	0.38	-143.	0.10	0.23	85.	85.	180.	1.95	0.09	0.59
0.17	0.63	-116.	0.51	-112.	0.12	0.21	85.	85.	20.	1.60	4.18	0.08
0.13	1.42	-100.	0.96	-91.	0.05	0.27	85.	85.	120.	1.63	0.05	0.68
0.11	2.68	-94.	1.78	-94.	0.05	0.54	85.	85.	150.	1.65	0.04	0.67
0.08	4.81	-94.	3.41	-94.	0.04	0.47	85.	85.	130.	1.62	0.05	0.71
0.07	5.68	-106.	3.55	-105.	0.12	1.93	90.	90.	45.	2.5.	0.09	0.01

FILE-

7 colony--76

PRD	R12	PH12	R21	PH21	R21	PH1	R1	PH	ANG	SKW	STA	RAN	AIS								
1337.75	124.64	-132.	-119.72	-162.	2.26	0.22	85.	155.	100.	100.	100.	100.	100.	100.	100.	100.	2.81	0.11	0.55	0.96	
875.76	69.86	-142.	62.64	-168.	1.66	0.21	85.	155.	135.	135.	135.	135.	135.	135.	135.	135.	2.88	0.06	0.83	0.91	
668.87	64.49	-131.	38.54	-163.	1.49	0.23	85.	150.	65.	65.	65.	65.	65.	65.	65.	65.	1.81	0.04	0.29	0.60	
546.13	47.33	-122.	30.85	-171.	1.55	0.24	90.	145.	45.	45.	45.	45.	45.	45.	45.	45.	1.17	0.09	0.26	0.69	
463.41	43.44	-129.	24.92	-170.	1.37	0.32	90.	145.	35.	35.	35.	35.	35.	35.	35.	35.	1.32	0.01	0.23	0.57	
357.53	70.31	-134.	18.37	-162.	1.10	0.28	90.	145.	100.	100.	100.	100.	100.	100.	100.	100.	0.91	0.03	0.12	0.26	
29.92	64.85	-136.	17.29	-160.	0.96	0.30	90.	145.	115.	115.	115.	115.	115.	115.	115.	115.	0.87	0.06	0.10	0.27	
229.42	53.56	-141.	19.09	-150.	0.41	0.24	90.	165.	85.	85.	85.	85.	85.	85.	85.	85.	1.03	0.07	0.07	0.36	
189.19	53.54	-144.	23.46	-145.	0.25	0.18	95.	175.	90.	90.	90.	90.	90.	90.	90.	90.	1.10	0.05	0.05	0.44	
168.44	53.71	-145.	24.16	-144.	0.19	0.15	95.	175.	95.	95.	95.	95.	95.	95.	95.	95.	0.99	0.05	0.04	0.45	
131.80	46.83	-152.	22.27	-144.	0.26	0.21	95.	155.	115.	115.	115.	115.	115.	115.	115.	115.	0.87	0.06	0.03	0.48	
106.29	48.06	-152.	21.54	-143.	0.15	0.17	95.	155.	130.	130.	130.	130.	130.	130.	130.	130.	0.70	0.07	0.03	0.45	
89.43	38.10	-155.	18.39	-145.	0.08	0.10	95.	155.	25.	25.	25.	25.	25.	25.	25.	25.	0.65	0.08	0.02	0.48	
67.06	35.47	-156.	19.20	-148.	0.11	0.13	95.	155.	0.	0.	0.	0.	0.	0.	0.	0.	0.57	0.10	0.02	0.54	
53.27	32.46	-161.	18.63	-150.	0.10	0.17	95.	145.	45.	45.	45.	45.	45.	45.	45.	45.	0.43	0.09	0.01	0.57	
42.32	28.82	-165.	17.00	-151.	0.17	0.17	95.	175.	35.	35.	35.	35.	35.	35.	35.	35.	0.36	0.12	0.02	0.59	
33.61	22.30	-161.	16.48	-152.	0.06	0.10	90.	175.	100.	100.	100.	100.	100.	100.	100.	100.	0.72	0.10	0.02	0.74	
26.70	18.73	-162.	14.89	-155.	0.05	0.09	90.	170.	110.	110.	110.	110.	110.	110.	110.	110.	0.66	0.06	0.01	0.79	
21.21	16.15	-165.	13.19	-157.	0.05	0.11	90.	170.	110.	110.	110.	110.	110.	110.	110.	110.	0.35	0.07	0.01	0.82	
16.85	13.70	-166.	10.27	-157.	0.09	0.19	90.	175.	95.	95.	95.	95.	95.	95.	95.	95.	0.22	0.08	0.01	0.75	
10.63	7.87	-166.	6.60	-161.	0.05	0.12	90.	175.	10.	10.	10.	10.	10.	10.	10.	10.	0.54	0.07	0.01	0.84	
8.64	6.64	-164.	5.52	-162.	0.06	0.17	90.	175.	65.	65.	65.	65.	65.	65.	65.	65.	0.65	0.07	0.01	0.84	
6.71	5.80	-161.	4.60	-156.	0.02	0.21	90.	175.	75.	75.	75.	75.	75.	75.	75.	75.	0.88	0.06	0.01	0.83	
5.33	5.70	-163.	5.29	-156.	0.01	0.17	90.	140.	75.	75.	75.	75.	75.	75.	75.	75.	0.44	0.06	0.01	0.79	
4.1.80	25.16	-159.	23.53	-150.	0.13	0.22	90.	150.	125.	125.	125.	125.	125.	125.	125.	125.	1.67	0.06	0.02	0.94	
27.37	16.14	-131.	15.96	-21.	0.13	0.22	90.	145.	75.	75.	75.	75.	75.	75.	75.	75.	1.19	0.06	0.01	0.99	
20.90	13.16	-158.	12.61	-152.	0.15	0.25	90.	150.	150.	150.	150.	150.	150.	150.	150.	1.58	0.06	0.01	0.96		
11.07	13.60	-161.	9.49	-157.	0.18	0.29	95.	165.	115.	115.	115.	115.	115.	115.	115.	115.	0.40	0.06	0.01	0.70	
14.48	12.00	-164.	7.95	-158.	0.16	0.29	95.	170.	140.	140.	140.	140.	140.	140.	140.	140.	0.40	0.09	0.01	0.66	
11.17	9.56	-163.	6.21	-161.	0.15	0.31	95.	170.	115.	115.	115.	115.	115.	115.	115.	115.	0.43	0.06	0.01	0.65	
9.12	6.36	-159.	6.09	-161.	0.22	0.29	95.	170.	125.	125.	125.	125.	125.	125.	125.	125.	2.26	0.07	0.01	0.99	
7.17	5.90	-117.	4.86	28.	0.19	0.48	0.	65.	40.	40.	40.	40.	40.	40.	40.	40.	0.56	0.06	0.01	0.82	
5.91	4.77	18.	4.09	20.	0.29	0.71	0.	65.	15.	15.	15.	15.	15.	15.	15.	15.	1.58	0.06	0.01	0.86	
5.26	4.21	22.	3.46	18.	0.25	0.62	0.	25.	25.	25.	25.	25.	25.	25.	25.	25.	1.37	0.06	0.01	0.82	
4.18	3.84	-154.	3.13	-150.	0.37	1.08	95.	85.	85.	85.	85.	85.	85.	85.	85.	85.	1.88	0.06	0.01	0.91	
3.32	3.26	3.26	3.22	-153.	0.33	0.07	90.	90.	90.	90.	90.	90.	90.	90.	90.	90.	0.58	0.06	0.01	0.99	
2.64	3.93	29.	2.74	37.	0.03	0.24	0.	70.	70.	70.	70.	70.	70.	70.	70.	70.	0.55	0.09	0.01	0.90	
2.10	2.67	-147.	2.67	-143.	0.06	0.14	90.	120.	105.	105.	105.	105.	105.	105.	105.	105.	2.15	0.06	0.01	1.00	
1.66	2.76	37.	2.69	33.	0.13	0.31	0.	5.	25.	25.	25.	25.	25.	25.	25.	25.	1.10	0.05	0.01	0.97	
1.32	3.04	36.	2.73	35.	0.08	0.12	0.	5.	5.	5.	5.	5.	5.	5.	5.	5.	1.15	0.05	0.01	0.90	
1.05	2.97	-160.	2.60	-142.	0.03	0.07	90.	145.	90.	90.	90.	90.	90.	90.	90.	90.	0.51	0.05	0.01	0.88	
0.83	3.13	36.	2.81	36.	0.02	0.13	90.	145.	45.	45.	45.	45.	45.	45.	45.	45.	1.68	0.04	0.01	0.90	
0.66	3.04	34.	2.76	34.	0.02	0.06	0.	105.	105.	105.	105.	105.	105.	105.	105.	105.	1.65	0.04	0.01	0.91	
0.53	3.03	33.	2.54	33.	0.01	0.03	0.	5.	5.	5.	5.	5.	5.	5.	5.	5.	0.52	0.02	0.01	0.84	
0.42	2.28	30.	2.17	29.	0.01	0.08	0.	25.	25.	25.	25.	25.	25.	25.	25.	25.	1.10	0.03	0.01	0.95	
0.33	1.47	-149.	1.07	-151.	0.01	0.04	90.	125.	5.	5.	5.	5.	5.	5.	5.	5.	1.28	0.01	0.01	0.73	
0.26	1.12	-148.	0.83	-150.	0.00	0.02	90.	145.	5.	5.	5.	5.	5.	5.	5.	5.	0.51	0.02	0.01	0.74	
0.21	0.75	-134.	0.63	-133.	0.01	0.03	90.	145.	115.	115.	115.	115.	115.	115.	115.	115.	0.09	0.04	0.01	0.83	
0.17	0.78	-108.	0.75	-103.	0.00	0.02	90.	145.	145.	145.	145.	145.	145.	145.	145.	145.	0.38	0.04	0.01	0.90	
0.13	1.80	89.	1.37	85.	0.00	0.02	0.	150.	150.	150.	150.	150.	150.	150.	150.	150.	0.52	0.03	0.01	0.98	
0.11	2.59	89.	2.36	87.	0.00	0.04	0.	130.	130.	130.	130.	130.	130.	130.	130.	130.	1.18	0.06	0.01	0.91	
0.06	4.20	87.	3.79	87.	0.01	0.08	0.	150.	150.	150.	150.	150.	150.	150.	150.	150.	0.67	0.02	0.01	0.80	
0.07	4.26	77.	3.46	85.	0.03	0.08	0.	100.	100.	100.	100.	100.	100.	100.	100.	100.	0.53	0.06	0.01	0.81	

	PWD	B12	PH21	R21	R21	RHT	ANG	ANGLE	LAUGH	SKIN	BPA	RAN	AIS	
137-75	27.12	21.	18.06	1.	7.11	1.43	170.	150.	5.	4.10	0.67	1.07	0.67	
175-76	10.14	110.	10.33	-163.	13.32	1.24	75.	130.	0.	2.94	0.70	0.99	0.73	
68-87	90.62	165.	21.62	-165.	13.88	2.38	70.	130.	0.	2.62	0.55	0.88	0.53	
46-61	55.69	16.	28.90	-169.	9.64	3.00	70.	120.	25.	2.62	0.50	0.87	0.52	
13-41	77.99	46.	56.66	52.	10.32	4.06	175.	115.	10.	1.94	0.01	0.61	0.73	
163-53	57.58	67.	30.22	32.	6.23	3.46	0.	105.	175.	1.63	0.01	0.11	0.53	
91-92	69.44	40.	25.61	43.	3.34	2.66	0.	100.	20.	1.10	0.04	0.07	0.37	
229-42	72.92	18.	22.91	60.	0.96	2.73	5.	165.	40.	0.51	0.17	0.05	0.31	
89-19	104.56	17.	51.82	80.	1.02	0.73	0.	115.	130.	1.04	0.13	0.09	0.50	
113-44	68.44	21.	63.26	38.	1.02	1.53	175.	120.	1.51	0.09	0.10	0.56	0.56	
33-80	101.51	19.	55.69	38.	0.54	0.71	175.	135.	115.	1.48	0.01	0.10	0.55	
06-29	93.21	21.	61.26	34.	0.81	0.53	175.	135.	110.	0.01	0.03	0.10	0.66	
84-43	81.12	19.	60.89	36.	0.51	0.68	175.	140.	10.	1.83	0.04	0.09	0.75	
67-06	68.96	17.	46.35	29.	0.21	0.43	175.	130.	10.	1.83	0.04	0.09	0.75	
53-27	65.27	18.	42.32	21.	0.17	0.38	0.	100.	145.	1.88	0.08	0.08	0.64	
42-32	55.91	15.	38.34	25.	0.60	0.67	175.	120.	1.67	0.06	0.05	0.65	0.65	
33-61	65.49	17.	36.70	25.	0.13	0.22	175.	135.	175.	1.57	0.05	0.02	0.81	
26-70	37.92	14.	36.91	22.	0.19	0.50	175.	120.	185.	2.75	0.03	0.03	0.97	
21-25	36.66	16.	31.14	23.	0.19	0.25	175.	125.	145.	1.54	0.07	0.02	0.85	
16-05	28.69	14.	26.32	22.	0.08	0.16	175.	120.	130.	1.98	0.09	0.03	0.92	
13-39	26.61	12.	19.31	17.	0.09	0.10	175.	110.	110.	1.16	0.09	0.02	0.72	
10-63	18.53	10.	13.30	19.	0.12	0.28	175.	120.	50.	0.52	0.05	0.01	0.72	
8-4	18.75	11.	13.35	17.	0.12	0.25	175.	150.	55.	0.49	0.06	0.01	0.71	
7-71	19.52	6.	12.85	14.	0.11	0.21	0.	90.	60.	1.20	0.05	0.03	0.66	
5.33	25.12	8.	13.48	10.	0.06	0.31	0.	115.	15.	0.66	0.11	0.02	0.54	
4.80	59.43	160.	22.64	-172.	0.58	0.38	5.	110.	115.	0.88	0.15	0.14	0.16	
27.37	39.41	-157.	5.44	-168.	0.28	0.61	0.	65.	65.	0.17	0.17	0.03	0.14	
20.90	33.47	-153.	15.81	-162.	0.30	0.44	0.	95.	15.	0.18	0.18	0.03	0.14	
17.07	32.93	-155.	24.51	-159.	0.42	0.47	0.	95.	25.	3.39	0.02	0.08	0.74	
14.48	56.89	2.	45.15	28.	0.43	0.43	0.	90.	100.	30.	1.81	0.25	0.20	0.79
11.17	22.01	-160.	15.43	-174.	0.33	0.48	0.	170.	90.	150.	0.45	0.07	0.01	0.70
9.12	17.80	-164.	8.87	-172.	0.16	0.31	175.	100.	110.	0.41	0.06	0.02	0.50	
7.17	11.80	-170.	4.61	168.	0.46	0.50	0.	60.	60.	5.97	0.02	0.04	0.48	
5.91	8.52	-178.	6.71	175.	0.62	1.53	0.	125.	125.	1.21	0.11	0.12	0.52	
5-26	13.33	18.	6.17	3.	0.60	1.28	95.	140.	10.	4.42	0.05	0.10	0.79	
5.18	6.31	-168.	5.83	-157.	0.30	0.55	0.	100.	100.	2.01	0.10	0.13	0.61	
3.32	7.83	29.	6.14	9.	0.15	0.31	85.	115.	5.	0.27	0.13	0.02	0.93	
2.64	5.80	-174.	5.5	-168.	0.07	0.50	0.	60.	5.	0.36	0.09	0.01	0.93	
2.20	6.21	-170.	3.44	-178.	0.18	0.76	175.	105.	5.	0.97	0.02	0.04	0.92	
1.66	6.22	-163.	3.45	-164.	0.05	0.29	175.	115.	115.	0.77	0.05	0.01	0.88	
0.33	5.31	-164.	4.73	-167.	0.04	0.16	0.	95.	45.	4.58	0.04	0.01	0.64	
1.05	4.16	16.	4.40	29.	0.02	0.15	85.	90.	45.	4.07	0.07	0.01	0.89	
0.83	4.58	17.	4.42	25.	0.02	0.15	85.	110.	25.	0.39	0.08	0.01	0.93	
0.66	4.95	21.	4.35	20.	0.02	0.17	85.	105.	5.	0.53	0.07	0.01	0.94	
0.53	5.16	-158.	5.53	-163.	0.03	0.12	175.	110.	85.	0.77	0.05	0.01	0.88	
0.42	4.82	-160.	4.47	-166.	0.03	0.06	175.	115.	125.	0.45	0.04	0.01	0.87	
0.33	4.19	11.	3.46	12.	0.01	0.04	0.	95.	75.	1.33	0.02	0.00	0.87	
0.26	3.35	5.	2.46	9.	0.01	0.07	95.	100.	5.	1.07	0.08	0.02	0.73	
0.21	2.62	-1.	2.49	-8.	0.00	0.02	85.	105.	5.	0.95	0.07	0.01	0.87	
0.17	1.57	-4.	1.45	-6.	0.00	0.03	85.	130.	105.	0.56	0.07	0.01	0.86	
0.13	0.69	-4.	0.61	-6.	0.00	0.03	90.	100.	60.	0.98	0.09	0.01	0.86	
0.11	0.36	0.	0.44	7.	0.01	0.02	85.	115.	115.	0.25	0.09	0.01	0.87	
0.08	0.14	-141.	0.14	-162.	0.01	0.04	0.	175.	115.	155.	0.32	0.07	0.01	0.87
0.07	0.16	66.	0.16	0.16	0.	0.	0.	0.	0.	0.	0.01	0.01	0.01	0.87





PAD	R12	PH12	R11	PH11	R1	PH1	RH1	ANG.	ANG.	ANG.	ANG.	SKW	BTA	RAN	AIS
1337.75	35.00	-31.	34.83	40.	2.75	0.34	5.	55.	105.	-10.28	0.09	0.64	1.00		
876.76	36.13	-145.	28.53	-155.	2.42	0.23	95.	55.	90.	7.26	0.09	0.64	0.79		
666.87	40.40	-145.	28.45	-148.	3.26	0.76	0.53	55.	140.	7.11	0.12	0.62	0.70		
566.13	38.06	-137.	29.61	-147.	2.93	0.53	95.	55.	145.	7.48	0.10	0.62	0.62		
466.41	37.44	-135.	24.70	-149.	3.00	0.48	95.	55.	75.	4.71	0.13	0.62	0.63		
355.53	37.49	-135.	24.70	-149.	2.05	0.43	95.	50.	75.	4.72	0.20	0.50	0.56		
281.92	31.11	-133.	20.67	-139.	0.75	0.39	95.	40.	65.	5.84	0.15	0.47	0.66		
229.42	30.01	-14.	19.15	-143.	0.21	0.53	95.	105.	55.	5.19	0.08	0.35	0.64		
189.19	45.03	-149.	17.88	-142.	0.72	0.38	100.	45.	115.	2.23	0.10	0.27	0.40		
168.44	47.86	-150.	18.22	-141.	0.70	0.36	95.	45.	100.	2.14	0.10	0.27	0.38		
133.80	35.03	-159.	17.22	-139.	0.70	0.40	95.	55.	25.	2.29	0.09	0.20	0.49		
106.29	35.63	-148.	18.17	-136.	0.27	0.19	95.	65.	15.	2.12	0.10	0.18	0.51		
84.63	36.75	-147.	19.49	-139.	0.32	0.31	95.	50.	50.	2.64	0.10	0.19	0.56		
67.06	32.20	-14.	23.88	-140.	0.18	0.22	95.	60.	35.	4.01	0.10	0.14	0.74		
53.27	33.86	-146.	24.57	-137.	0.17	0.23	95.	50.	50.	2.87	0.10	0.11	0.73		
42.32	32.38	-146.	20.99	-143.	0.09	0.19	95.	75.	30.	2.47	0.12	0.09	0.65		
33.61	31.13	-49.	15.88	-141.	0.13	0.26	95.	110.	175.	1.56	0.13	0.08	0.54		
26.70	30.16	-151.	15.98	-141.	0.05	0.18	95.	55.	55.	1.12	0.11	0.05	0.53		
21.21	26.44	-151.	17.08	-142.	0.05	0.25	95.	80.	60.	1.19	0.13	0.04	0.65		
16.85	25.33	-155.	13.96	-141.	0.10	0.32	95.	30.	73.	0.15	0.04	0.04	0.55		
13.38	22.09	-156.	11.31	-144.	0.10	0.50	95.	75.	45.	0.73	0.18	0.05	0.51		
10.63	19.21	-155.	11.93	-143.	0.08	0.33	90.	95.	95.	0.76	0.18	0.05	0.62		
8.44	18.96	-157.	8.35	-148.	0.08	0.46	90.	160.	85.	0.56	0.18	0.07	0.44		
6.71	15.06	-163.	6.00	-158.	0.15	0.60	95.	105.	55.	0.80	0.14	0.05	0.40		
5.33	12.87	-163.	5.66	-165.	0.10	0.62	95.	90.	115.	0.86	0.17	0.06	0.44		
4.1.80	31.93	26.	41.17	29.	0.30	0.58	105.	60.	125.	4.04	0.09	0.18	0.66		
27.37	25.26	35.	35.72	50.	0.34	0.70	95.	65.	155.	1.51	0.05	0.05	0.62		
20.90	20.92	47.	35.72	44.	0.21	0.22	75.	75.	172.	0.13	0.12	0.12	0.47		
17.07	17.25	45.	6.21	43.	0.31	0.96	0.5.	65.	55.	1.76	0.10	0.11	0.48		
14.48	20.60	26.	7.02	40.	0.89	1.84	90.	110.	170.	0.29	0.11	0.02	0.34		
11.17	11.25	39.	7.61	50.	0.61	0.52	90.	120.	60.	2.09	0.13	0.03	0.66		
9.12	12.53	-130.	1.43	-16.	1.34	2.16	5.	105.	2.16	0.11	0.11	0.11	0.81		
7.17	23.93	-154.	2.72	-160.	1.61	4.80	5.	145.	60.	2.23	0.07	0.01	0.24		
5.91	10.10	-145.	12.51	-120.	1.06	6.16	170.	45.	170.	1.19	0.05	0.05	0.66		
5.26	43.10	-166.	2.49	-117.	2.35	11.32	165.	165.	165.	1.52	0.05	0.05	0.62		
4.18	25.81	-165.	11.32	-158.	0.60	4.08	1.84	1.84	1.84	1.46	0.51	0.40	0.43		
3.12	17.54	-144.	6.82	-154.	0.33	3.22	175.	175.	175.	0.37	0.05	0.05	0.44		
2.64	24.33	-164.	10.34	-156.	0.33	3.63	175.	175.	175.	0.37	0.05	0.05	0.44		
2.10	23.91	-166.	6.94	-159.	0.22	1.37	175.	175.	175.	0.37	0.05	0.05	0.44		
1.66	20.23	-161.	6.19	-166.	0.12	1.31	175.	175.	175.	0.37	0.01	0.01	0.24		
1.05	20.17	-176.	4.67	-161.	0.21	1.67	175.	175.	175.	0.37	0.01	0.01	0.24		
0.83	21.71	-174.	4.44	-163.	0.16	1.52	175.	175.	175.	0.37	0.01	0.01	0.24		
0.66	17.03	-178.	4.12	-159.	0.14	1.10	175.	175.	175.	0.37	0.01	0.01	0.24		
0.53	18.53	-171.	3.55	-166.	0.12	1.31	175.	175.	175.	0.37	0.01	0.01	0.24		
0.42	11.77	-167.	3.40	-167.	0.12	1.35	175.	175.	175.	0.37	0.01	0.01	0.24		
0.33	19.68	-163.	3.31	-160.	0.31	1.67	175.	175.	175.	0.37	0.01	0.01	0.24		
0.26	7.90	-164.	3.25	-160.	0.10	0.93	175.	175.	175.	0.37	0.01	0.01	0.24		
0.21	2.53	-157.	1.72	-172.	0.09	0.44	175.	175.	175.	0.37	0.01	0.01	0.24		
0.17	2.52	-150.	1.51	-171.	0.09	0.56	175.	175.	175.	0.37	0.01	0.01	0.24		
0.13	1.46	-154.	1.42	-177.	0.07	0.38	175.	175.	175.	0.37	0.01	0.01	0.24		
0.11	0.29	-176.	0.53	-179.	0.06	0.31	175.	175.	175.	0.37	0.01	0.01	0.24		
0.08	0.14	-150.	0.11	-170.	0.09	0.33	175.	175.	175.	0.37	0.01	0.01	0.24		
0.07	0.10	-116.	0.10	-102.	0.12	0.52	175.	175.	175.	0.37	0.01	0.01	0.24		

PAD	R112	RH12	R21	RE1	FH1	ANG	ANG2	ANCH	SKIN	BTM	BAN	ATB	ATB
1337.75	54.92	-129.	26.15	-162.	7.28	-1.66	85.	30.	90.	2.18	-0.46	0.65	0.46
675.76	69.99	-136.	22.98	-162.	5.15	-1.34	85.	25.	85.	2.45	-0.37	0.62	0.36
668.87	46.63	-138.	20.38	-158.	3.99	-1.18	85.	25.	80.	2.61	-0.30	0.55	0.34
546.13	45.74	-143.	20.20	-150.	2.52	-0.95	85.	25.	80.	3.16	-0.18	0.45	0.44
463.61	41.37	-148.	18.91	-167.	2.67	-0.78	85.	20.	80.	3.33	-0.15	0.43	0.45
357.53	40.23	-146.	17.69	-147.	2.63	-0.72	85.	20.	75.	2.93	-0.13	0.35	0.44
291.92	37.63	-147.	18.10	-148.	2.05	-0.70	85.	20.	75.	3.24	-0.14	0.35	0.46
229.42	36.15	-150.	18.67	-148.	1.51	-0.57	85.	20.	75.	3.57	-0.13	0.35	0.46
189.19	35.73	-151.	17.56	-148.	1.16	-0.46	85.	20.	75.	3.20	-0.17	0.33	0.49
168.66	35.10	+150.	16.88	-151.	0.93	-0.38	85.	20.	75.	3.09	-0.15	0.32	0.49
133.60	30.40	-154.	15.85	-154.	0.77	-0.38	85.	15.	105.	3.33	-0.13	0.30	0.52
106.29	25.88	-154.	14.22	-152.	0.52	-0.25	85.	15.	75.	3.48	-0.14	0.28	0.55
88.93	21.56	-154.	10.66	-150.	0.45	-0.27	90.	15.	155.	2.57	-0.14	0.22	0.49
67.06	21.79	-156.	9.46	-149.	0.38	-0.24	90.	15.	155.	1.72	-0.13	0.13	0.43
53.27	16.65	-158.	9.68	-158.	0.25	-0.21	90.	10.	100.	1.59	-0.05	0.13	0.52
42.32	16.61	-158.	8.42	-152.	0.20	-0.23	90.	10.	60.	2.02	-0.03	0.13	0.52
33.61	13.33	-160.	7.84	-153.	0.17	-0.22	90.	10.	155.	1.98	-0.03	0.12	0.51
26.70	11.51	-161.	6.20	-150.	0.11	-0.19	90.	10.	160.	2.13	-0.04	0.19	0.59
21.21	11.45	-160.	5.60	-151.	0.09	-0.16	90.	10.	155.	1.50	-0.07	0.07	0.54
16.85	8.73	-160.	5.66	-153.	0.09	-0.17	90.	10.	130.	1.33	-0.02	0.06	0.49
13.38	6.98	-160.	4.67	-150.	0.10	-0.18	90.	10.	120.	1.65	-0.05	0.06	0.45
10.63	5.84	-156.	4.49	-148.	0.08	-0.15	85.	10.	165.	1.59	-0.05	0.05	0.67
8.44	4.76	-151.	4.10	-148.	0.07	-0.12	85.	10.	160.	1.53	-0.05	0.04	0.62
6.71	4.69	-149.	3.97	-142.	0.06	-0.12	90.	10.	165.	1.70	-0.07	0.03	0.64
5.33	4.56	-148.	3.16	-141.	0.05	-0.13	85.	10.	150.	1.10	-0.08	0.02	0.69
41.80	14.97	31.	14.12	32.	0.32	0.34	85.	30.	35.	19.33	-0.05	0.10	0.94
27.37	13.37	24.	7.72	39.	0.27	0.29	80.	10.	30.	1.25	0.07	0.06	0.74
20.90	10.90	22.	6.01	36.	0.21	0.23	90.	10.	5.	1.43	-0.11	0.07	0.61
17.97	9.99	26.	6.12	35.	0.17	0.21	90.	10.	170.	1.40	-1.18	0.10	0.62
16.48	9.05	19.	5.64	35.	0.15	0.21	90.	10.	165.	1.35	-0.98	0.10	0.64
11.17	7.26	23.	2.10	35.	0.09	0.17	90.	10.	135.	1.50	-1.20	0.10	0.70
9.12	6.26	26.	4.20	36.	0.08	0.15	90.	10.	130.	1.55	-1.22	0.08	0.68
7.17	5.26	34.	4.01	34.	0.08	0.15	90.	10.	155.	1.55	-2.29	0.08	0.74
5.91	5.75	41.	4.41	47.	0.14	0.26	90.	10.	170.	1.40	-0.04	0.04	0.74
5.26	4.57	43.	4.33	46.	0.22	0.26	90.	10.	135.	1.45	-0.34	0.07	0.63
4.18	4.32	46.	4.30	43.	0.20	0.25	90.	10.	105.	1.57	-0.57	0.08	0.62
3.32	10.33	-156.	4.42	-148.	0.10	0.30	90.	10.	175.	1.00	-0.31	0.41	0.15
2.64	57.26	-135.	16.49	-139.	0.13	0.41	100.	10.	165.	1.22	-0.03	0.54	0.32
2.10	32.36	56.	7.38	58.	0.09	0.56	90.	10.	155.	1.22	-0.03	0.54	0.32
1.66	24.36	-126.	7.71	-125.	0.10	0.50	90.	10.	175.	1.45	-1.46	0.34	0.85
1.32	33.79	-118.	13.84	-129.	0.07	0.55	90.	10.	135.	1.45	-1.46	0.06	0.32
1.05	18.63	-130.	16.07	-134.	0.03	0.52	90.	10.	170.	2.88	-0.13	0.29	0.58
0.83	18.57	46.	12.94	44.	0.08	0.39	90.	10.	165.	2.80	-0.03	0.06	0.72
0.66	15.15	33.	14.73	39.	0.02	0.24	90.	10.	170.	2.88	-0.05	0.03	0.90
0.53	12.46	36.	11.92	35.	0.01	0.15	90.	10.	155.	3.63	-0.12	0.07	0.84
0.42	11.87	26.	10.58	27.	0.02	0.15	85.	10.	170.	1.72	-0.03	0.03	0.89
0.33	11.86	19.	9.64	20.	0.01	0.10	85.	10.	165.	2.91	-0.09	0.01	0.93
0.26	9.56	11.	7.51	12.	0.01	0.06	90.	10.	170.	1.16	-0.06	0.01	0.81
0.21	7.31	44.	4.34	42.	0.02	0.11	90.	10.	170.	1.05	-0.05	0.01	0.76
0.17	4.23	25.	4.72	6.	0.02	0.14	90.	10.	175.	1.55	-0.55	0.06	0.54
0.13	1.82	7.9	5.	5.	0.01	0.07	90.	10.	165.	1.72	-0.03	0.01	0.64
0.11	0.92	7.	4.	4.	0.01	0.07	90.	10.	165.	2.46	-0.05	0.01	0.74
0.08	0.36	39.	35.	35.	0.01	0.07	90.	10.	170.	2.63	-0.06	0.01	0.64
0.07	0.21	71.	39.	39.	0.01	0.07	90.	10.	170.	2.45	-0.07	0.01	0.77

PED.	RH12	RH12	RH21	RH21	RH1	BE1	ANG	ANG	ANG	ANG	SKW	BRA	BAN	MIS
1337-75	120.43	-141.	11.89	-124.	-1.36	0.46	170.	35.	85.	1.36	0.11	0.43	0.10	0.10
875-76	115.13	-148.	15.06	-146.	0.46	0.48	170.	0.	95.	1.46	0.06	0.38	0.13	0.13
668-87	101.03	-150.	15.19	-147.	0.56	0.43	170.	15.	110.	1.50	0.08	0.37	0.15	0.15
546-13	89.51	-147.	24.06	-149.	0.32	0.35	175.	110.	105.	1.96	0.06	0.35	0.27	0.27
463-41	86.96	-147.	23.18	-158.	0.34	0.40	170.	145.	110.	2.05	0.08	0.41	0.41	0.42
357-53	83.21	-146.	24.05	-158.	0.39	0.49	170.	175.	110.	2.04	0.10	0.39	0.29	0.29
291-92	77.13	-146.	31.19	-156.	0.52	0.59	175.	5.	100.	2.51	0.12	0.34	0.41	0.41
229-42	70.15	-149.	30.24	-164.	0.50	0.56	175.	5.	100.	2.57	0.03	0.29	0.43	0.43
189-19	65.92	-154.	27.81	-162.	0.43	0.54	175.	10.	100.	2.13	0.01	0.26	0.42	0.42
168-84	65.33	-151.	26.19	-145.	0.37	0.53	175.	0.	105.	2.34	0.06	0.26	0.40	0.40
133-80	65.84	-157.	19.08	-138.	0.49	0.53	175.	15.	115.	2.21	0.12	0.35	0.42	0.42
106-29	47.85	-158.	17.25	-152.	0.30	0.51	175.	175.	115.	2.32	0.04	0.33	0.36	0.36
86-43	61.01	-158.	15.10	-151.	0.44	0.53	175.	20.	105.	2.39	0.05	0.34	0.37	0.37
67-06	38.97	-156.	14.03	-155.	0.30	0.52	175.	5.	105.	2.31	0.04	0.32	0.36	0.36
53-27	33.38	-160.	11.95	-148.	0.30	0.51	175.	5.	105.	2.31	0.04	0.32	0.36	0.36
42-32	29.63	-161.	11.15	-148.	0.25	0.54	175.	5.	105.	2.05	0.06	0.26	0.30	0.30
33-61	26.71	-162.	10.30	-146.	0.23	0.52	175.	110.	115.	2.05	0.09	0.27	0.38	0.38
26-70	20.88	-163.	10.80	-150.	0.20	0.47	165.	165.	115.	2.05	0.09	0.27	0.42	0.42
21-21	18.78	-163.	9.59	-152.	0.17	0.45	165.	115.	115.	2.68	0.08	0.28	0.52	0.52
16-85	16.68	-164.	9.65	-150.	0.16	0.41	165.	115.	115.	2.33	0.09	0.27	0.51	0.51
13-38	14.02	-164.	8.53	-153.	0.16	0.38	165.	0.	120.	2.33	0.07	0.19	0.58	0.58
10-61	16.99	-163.	6.60	-153.	0.15	0.35	165.	5.	120.	2.38	0.07	0.16	0.36	0.36
8-44	16.69	-163.	5.54	-149.	0.09	0.23	175.	15.	155.	2.29	0.06	0.13	0.30	0.30
6-71	27.69	21.	20.68	19.	0.10	0.26	90.	90.	180.	2.51	0.11	0.32	0.60	0.60
5-33	32.00	34.	14.52	33.	0.08	0.19	85.	160.	170.	2.70	0.13	0.17	0.75	0.75
41-80	39.61	16.	16.44	23.	0.22	0.61	0.	135.	120.	2.21	0.21	0.21	0.44	0.44
27-37	20.27	23.	11.46	30.	0.21	0.52	0.	135.	125.	1.62	0.09	0.19	0.33	0.33
20-90	21.91	24.	9.25	29.	0.21	0.57	0.	145.	130.	1.65	0.08	0.16	0.38	0.38
17-07	20.12	24.	8.16	35.	0.19	0.50	0.	145.	125.	2.06	0.10	0.21	0.42	0.42
14-48	17.72	24.	6.94	35.	0.18	0.50	0.	150.	130.	1.83	0.05	0.19	0.79	0.79
11-17	16.13	23.	6.94	30.	0.19	0.51	0.	165.	125.	2.30	0.05	0.17	0.41	0.41
9-12	16.54	24.	6.37	34.	0.15	0.31	175.	100.	170.	1.50	0.06	0.16	0.35	0.35
7-17	6.37	7.	14.60	49.	0.23	0.40	175.	165.	140.	1.76	0.14	0.05	0.48	0.48
5-91	6.11	31.	4.84	25.	0.18	0.49	120.	120.	125.	1.93	0.09	0.27	0.41	0.41
5-26	4.82	34.	1.60	35.	0.15	0.35	170.	110.	125.	2.06	0.10	0.21	0.42	0.42
4-18	5.82	35.	4.84	44.	0.41	0.45	170.	155.	130.	0.66	0.25	0.23	0.33	0.33
3-32	10.00	15.	4.84	13.	0.20	0.49	175.	85.	90.	0.73	0.09	0.04	0.75	0.75
2-64	4.46	24.	4.56	35.	0.09	0.18	175.	155.	120.	0.78	0.07	0.01	0.95	0.95
2-10	2.51	18.	1.52	25.	0.08	0.18	175.	120.	135.	1.28	0.06	0.05	0.28	0.28
1-66	3.69	31.	4.84	40.	0.15	0.18	120.	120.	155.	1.91	0.05	0.02	0.48	0.48
1-32	2.71	34.	2.04	42.	0.09	0.17	0.	120.	110.	1.62	0.14	0.09	0.83	0.83
1-05	2.59	-135.	1.60	44.	0.07	0.28	0.	120.	70.	1.71	0.17	0.11	0.56	0.56
0-33	3.16	130.	4.45	11.	0.04	0.21	90.	90.	85.	1.58	0.07	0.04	0.85	0.85
0-66	3.69	34.	3.46	27.	0.05	0.27	0.	135.	90.	0.84	0.07	0.01	0.90	0.90
0-53	3.77	20.	3.46	21.	0.05	0.35	0.	150.	95.	1.91	0.05	0.02	0.99	0.99
0-32	3.75	19.	3.35	20.	0.04	0.16	0.	140.	120.	1.62	0.14	0.09	0.87	0.87
0-33	3.31	10.	2.64	6.	0.02	0.07	0.	150.	90.	1.43	0.02	0.00	0.89	0.89
0-26	2.58	b.	4.31	1.	0.01	0.07	0.	150.	100.	1.93	0.04	0.01	0.90	0.90
0-21	1.77	17.	1.62	170.	0.02	0.16	90.	120.	85.	1.35	0.04	0.01	0.92	0.92
0-17	1.03	175.	0.95	169.	0.01	0.08	90.	125.	65.	1.85	0.01	0.01	0.92	0.92
0-13	1.50	176.	0.30	172.	0.01	0.05	90.	120.	90.	0.56	0.04	0.01	0.93	0.93
0-11	0.22	174.	0.41	177.	0.01	0.06	90.	120.	130.	0.75	0.03	0.00	0.99	0.99
0-08	0.11	-140.	0.39	-147.	0.01	0.04	90.	125.	65.	1.05	0.02	0.00	0.92	0.92
0-07	0.11	-108.	0.37	-122.	0.14	0.49	100.	135.	110.	0.42	0.01	0.01	0.85	0.85